

AN11031

PCB layout and ADC circuit recommendations for LPC1100 and LPC1300

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Application note

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1. Introduction

The LPC1100 is based on the ARM Cortex M0, and the LPC1300 family is based on the ARM Cortex M3 core. Both families include an 8 channel 10-bit successive approximation Analog-to-Digital Converter (ADC) with a maximum conversion rate of 400 kHz.

The ADCs of the LPC1100 and LPC1300 differ from other LPC devices in that they do not have an external V_{ref} ¹. This means that extra precautions have to be taken to minimize noise effects that can affect the conversion results.

The goal of this Application Note is to show how to develop a printed circuit board (PCB) that is optimized for the ADC on the LPC1100 and LPC1300 families.

This application note is intended as a guideline for hardware and/or PCB designers with basic PCB design knowledge, with the intent of suggesting methods to prevent potential electromagnetic interference (EMI) and electromagnetic compatibility (EMC) problems.

Unlike many other design issues, EMC is not an area where it is possible to list a set of rules. In general, EMC compliance cannot be guaranteed by design; it has to be tested. That is why this application note includes a test software routine to compare and quantify the difference between a basic and an optimized design.

[1] V_{ref} = reference voltage source

2. PCB Optimizations of the ADC

The microcontrollers in the LPC1100 and LPC1300 families are all equipped with a single external power supply. The digital (core) and analog parts (ADC/ V_{ref}) are all connected to the same external supply. Without precautions, noise produced by the core could influence the readout of the ADC via the internal V_{ref} .

Inside the devices, some measures are taken to reduce the noise on the ADC unit. For example, having an extra set of bonding wires and a separate isolated internal regulator reduces the noise transfer from the core to the ADC. To shield the ADC lines, decoupling capacitors are added. In [Fig 1](#) a schematic wire interconnection block diagram of the ADC and the core is shown.

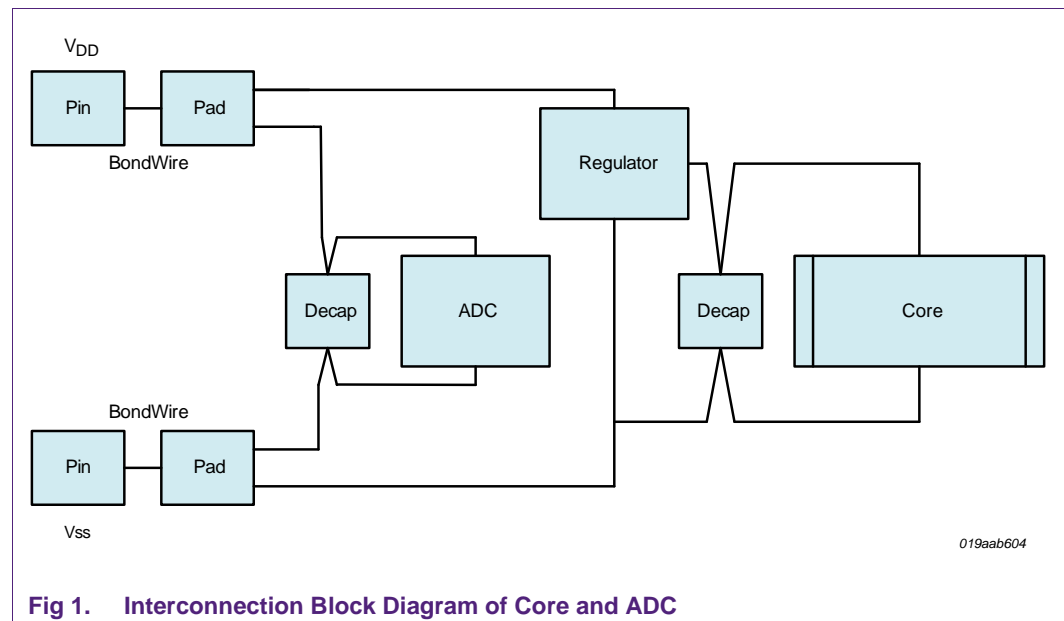


Fig 1. Interconnection Block Diagram of Core and ADC

The physical location of the ADC-block inside the chip is as close as possible to the ADC pins in order to reduce noise.

The LPC1100 and LPC1300 series microcontrollers are equipped with a 10-bit ADC; therefore the resolution of the ADC is $V_{DD} / 1024$. For example this means when V_{DD} equals 3.3V, then resolution is $3.3V / 1024 = \sim 3.2 \text{ mV/bit}$. With these low voltages noise suppression is important. To suppress noise on the V_{DD} of the device, decoupling capacitors on the PCB is very important.

2.1 Example hardware design

The following section provides some details on board layout, grounding and decoupling.

2.1.1 ADC pin out

First step is to make an electrical design. If one does not need all the ADC-channels, the required channels can be selected in a manner so as to reduce crosstalk from the digital domain. See the user manual for the physical pin-out. In the example shown all the channels are wired.

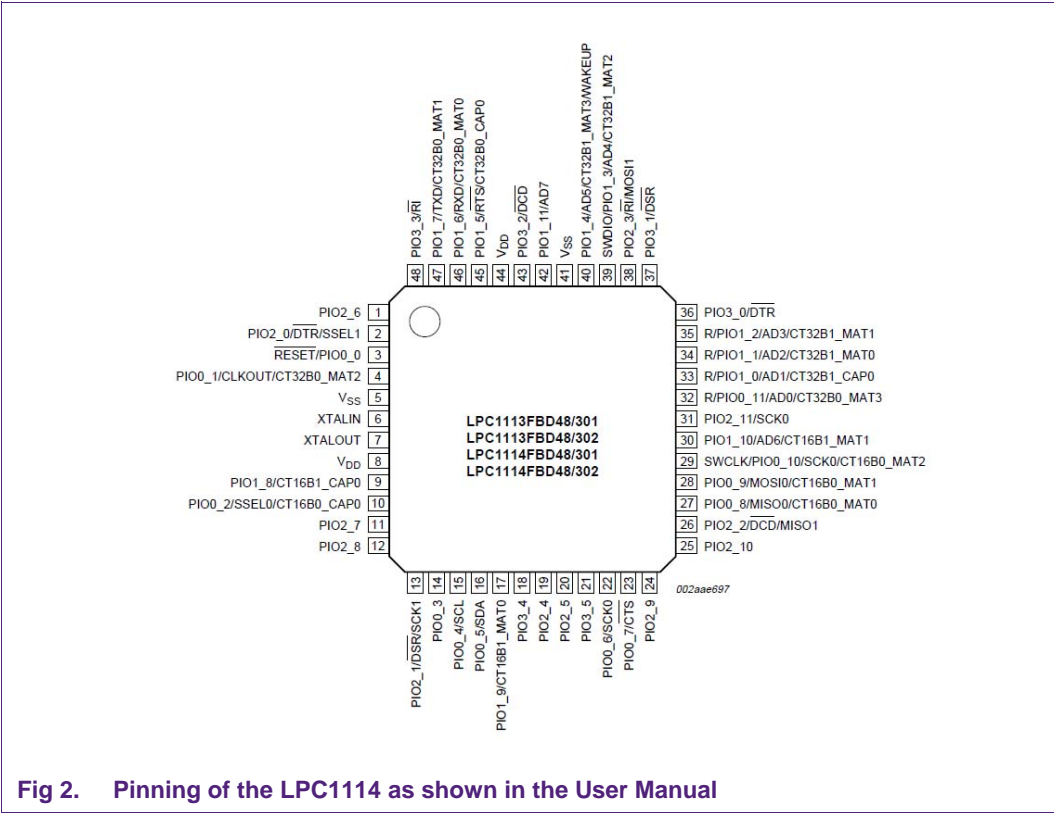


Fig 2. Pinning of the LPC1114 as shown in the User Manual

Observe that all the ADC-channels are situated relatively close to each other (see pins 30, 32 ... 35, 39, 40 and 42). Since some digital signal lines are located in between the analog pin, it is recommended avoid them when possible. Doing so reduces the amount of crosstalk transferred onto the ADC block.



Fig 3. Electrical design with locations where extra precautions have to be taken for shielding (see: dashed lines)

2.1.2 Separate or isolate ground planes

The analog and the digital domains are separated inside the chip; however, they are linked to the same power pins. Therefore, noise on the power and ground lines should be reduced as much as possible. Major noise sources are the Oscillator, CPU core, system PLL, Clock-Tree and General Purpose IO-Ports. A decrease of noise can be achieved by segmenting the ground plane into the digital and analog domains. These planes are physically separated by a small gap and connected only at one point that is a few millimeters in size.

[Fig 4](#) depicts the main (digital) ground plane 'A', the sub ground (analog) planes 'B', 'C' (with ADC-filters), and sub ground plane 'E' (Crystal-circuitry).

By creating an analog sub ground plane with only one entrance 'F'; local ground currents are restricted to travel through this gap. Confining these analog ground currents to this gap prevents other currents entering the analog sub plane and interfering with ADC conversions.

Although the quartz oscillator is not a big source of emissions, it may be susceptible to noise; therefore, special care must be taken when routing the PCB's oscillator section.

One can place a guard ring with capacitors around the quartz oscillator circuit for additional emission isolation. When possible, avoid close proximity between clocks and analog traces. See section 'E' in [Fig 4](#).

The CPU core, system PLL and Clock-Tree are responsible for noise on the power lanes. When using one or more clock dividers, low frequency harmonics are to be expected. The noise contribution of General Purpose IO-pins cannot be generally estimated as it depends on the user pin configuration. If a GPIO is toggled frequently, it may also create additional noise. By keeping digital pin toggling to a minimum, it may be possible to reduce noise emissions.

The noise emissions from GPIO pins are due to their narrow band character. One could expect such emissions in system clock drivers or CAN data lines.

2.1.3 Decouple the supply system

A power supply system consists of one or more power pins and their related ground pins. The LPC1100 and LPC1300 are equipped with two supply pins i.e. two V_{DD} and V_{SS} pins. At least one decoupling capacitor for each power supply system is needed to ensure low impedance over a wide frequency range. In this design two capacitors are placed as close as possible to the power pins of the microcontroller.

In [Fig 4](#) the capacitors are indicated by the letters 'G' and 'H'. For improved decoupling, two vias are used to connect the capacitor to the ground plane.

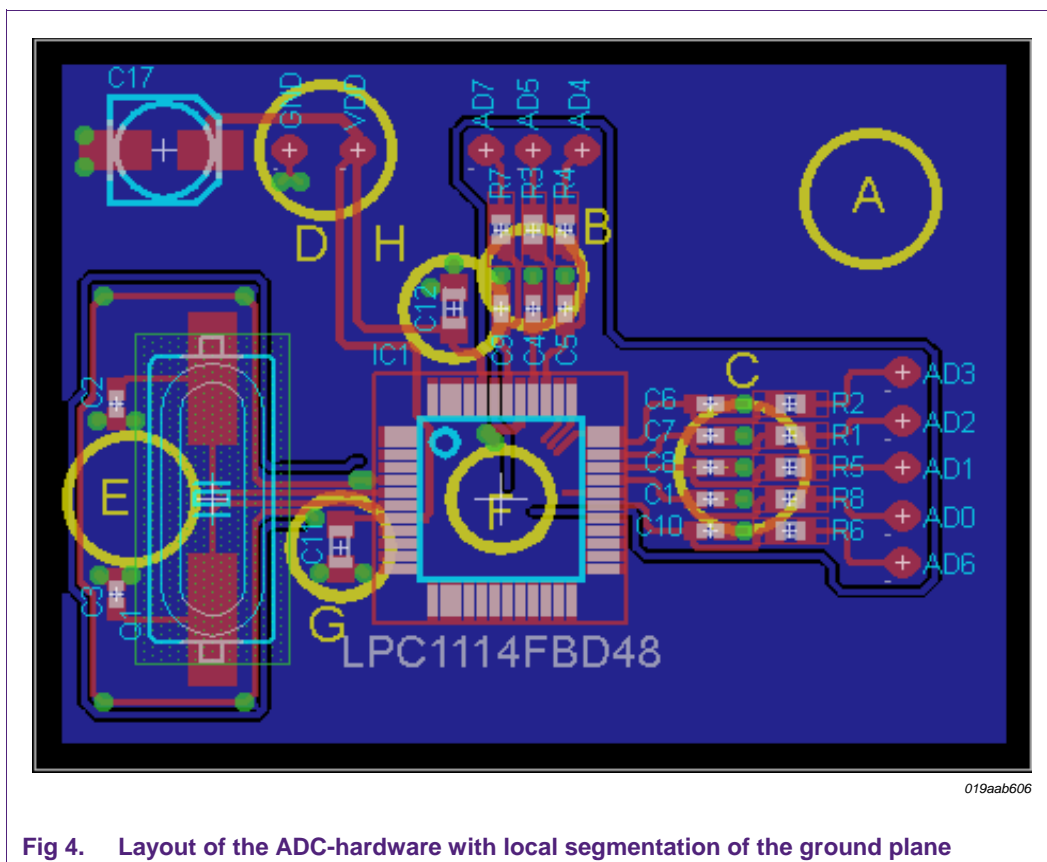


Fig 4. Layout of the ADC-hardware with local segmentation of the ground plane

Any active element inside a microcontroller has direct or indirect connections to the power supply system; thus, any switching inside the microcontroller will cause current to flow. The noise from this current loop is proportional to the area of the loop(s) in which the current flows; therefore, it's necessary to design these loops as small as possible. One such loop is the current loop between the microcontroller and the decoupling capacitors. A power supply has an impedance greater than 0 Ohm. Particularly at higher frequencies the inductive wire impedance of the power supply becomes more significant. Pulsed currents overlay a ripple voltage to the DC power supply that contributes to noise emissions.

By using thicker traces to the microcontroller's power supply its inductive impedance is lowered, thus effectively reducing noise emissions. Due to the power line emissions, it is suggested to separate power lines from signal lines to decrease crosstalk.

Use electrolytic capacitors in the vicinity of the PCB's power source input (shown left side of 'D') and ceramic capacitors next to the power pins of the microcontroller. Input filters connected to the ADC will have a significant impact on the noise immunity of the analog parts. The type of the filters to be used depends on the input signal frequency. The location of the filter's capacitors is also very important. It's recommended to place the filter capacitors onto the same (analog) sub ground plane as shown in [Fig 4](#).

When possible, separate analog traces from digital traces; for example one could use extra grounded traces. Keep digital pins away from analog pins when physically possible. When that is not possible, try to use extra shielding between digital and the analog tracks.

3. Test software for optimizing the hardware of the ADC

Testing the noise immunity of a design is very important. This can be done by using a small software program, which gives a numerical indication of the stability of the ADC conversion results. The test software was developed using the NXP's LPCXpresso IDE:

<http://ics.nxp.com/support/lpcxpresso/>

To determine how relatively stable the ADC design is, 1,000,000 ADC measurements are taken of a known voltage. Of those measurements, the most stable result is within a 1 to 2 LSBs of the expected ADC conversion.

For example if V_{DD} equals 3.3 V and AD7 equals 1.6 V: The readout should be in between 496 and 497 (decimal).

3.1 Gauging the ADC conversion results

Generating a histogram for a series of ADC readings can be done using the following two pass process:

1. Connect the LPCXpresso board to the workstation and apply an external 1.6 V voltage to P1.11 (AD7).
2. "Build all projects" and Debug the "adc" project to execute the software.
3. During the first pass the software performs 1,000,000 measurements in order to calculate the average. This average is used to determine the ADC conversion values' approximate location on a histogram. It essentially "zooms in" into the area of interest.
4. The second pass takes another 1,000,000 measurements and increments the bins that correspond to the individual readings between [avg - 15] to [avg + 15]. This will populate the array with the histogram's data.
5. Export the data to a PC to visualize the results. The application example allows a "copy and paste" of the data into a spreadsheet program to plot a scatter chart.

The histogram gives a graphical representation of distribution of ADC conversions. Running this application could be useful if changes on the electronic design were performed; for example by adding extra filtering, relocating traces, or changing decoupling capacitors.

Ideally, a good set of ADC conversions would result a histogram with a very narrow distribution of samples; meaning that the ADC sampling would have a higher precision.

[Fig 5](#) shows three sets of sample data. ADC data sets D1 and D2 are suffering from noise interference while column D3 shows a set where ADC layout was optimized. Notice that data set D3 has a very narrow distribution set.

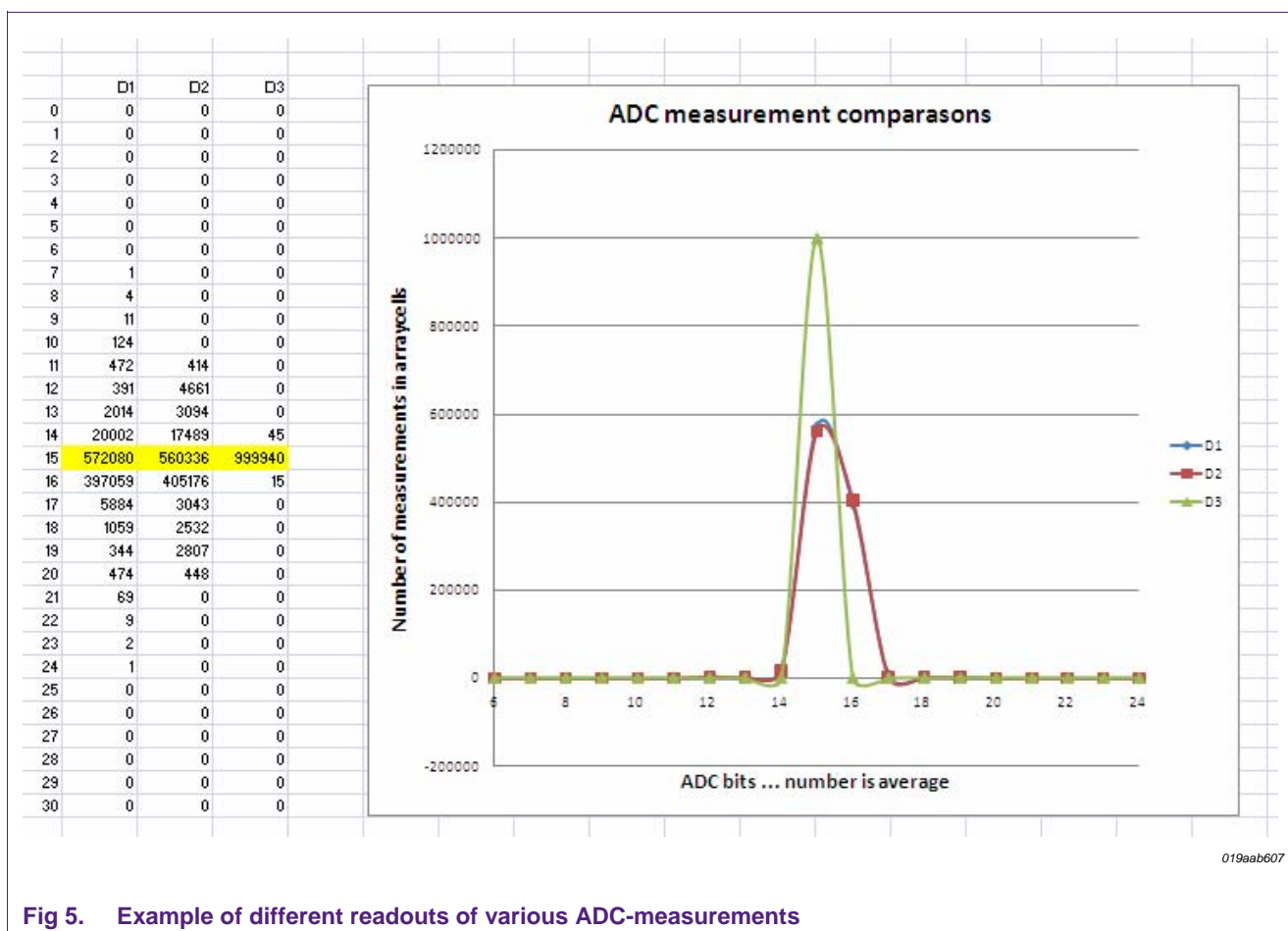


Fig 5. Example of different readouts of various ADC-measurements

3.2 Software example

As mentioned earlier, this sample application takes 1,000,000 measurements and stores histogram distribution into the `cnt_adc_val[31]` array. The center of the histogram is located at `cnt_adc_val[15]`, which corresponds to the histogram's bin at which the software determined the ADC conversion to be equal to the previously calculated average.

For example if the calculated average value from the first iteration of 1,000,000 samples equal 510. Then `cnt_adc_val[15]` will contain a count value of how many times out of the second iteration of 1,000,000 samples the ADC has again measured 510.

By storing the distribution in this manner, it implies that the remainder of the `cnt_adc_val[]` array is used to store the frequency count where the ADC has measured 509 (`cnt_adc_val[14]`), 511 (`cnt_adc_val[16]`), and etc...

This pattern is repeated for the entire array, essentially allowing us to collect a histogram focusing on the "area of interest" (which is nearby the expected average value).

```
// actual calculations -> result 30 counter-values
for (j=0;j<1000000;j++)
{
    ADCRead( 7 );           // ADC conversion
    while ( !ADCIntDone );  // wait till conversion is done
    ADCIntDone = 0;
    temp_ADC7 = ADCValue[7]; // put ADC-value in temp-var
    // value should be > 0.048 V and < 3.248 V
    if ((temp_ADC7 > 15) && (temp_ADC7 < 1023 - 15))
    {
        if (j%100000==0) UARTSend( (uint8_t *)(".", 1 );

        if (temp_ADC7 < (sum_avg - 15)) cnt_adc_val[0]++;
        if (temp_ADC7 > (sum_avg + 15)) cnt_adc_val[30]++;
        if ((temp_ADC7 >= (sum_avg - 15)) && (temp_ADC7 <= (sum_avg + 15)))
        {
            cnt_adc_val[temp_ADC7 - sum_avg + 15]++;
        }
    }
}
}
```

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Fig 6. Software excerpt where the histogram distribution is calculated

```

adc Debug [C/C++ MCU Application] C:\Documents and Settings\Nxp12832\Desktop\AN11031\ADC\LPC1113\AN11031
Starting the ADC
Taking 1,000,000 ADC samples
.....
Calculated sample average: 515
Taking another 1,000,000 samples
.....
Printing histogram data...
Column 1: +/- from the AVG bin
Column 2: # of samples that match the bin
Note that each bin increment corresponds to a LSB of the ADC
(Hint: Copy & Paste into Excel)
-15 0
-14 0
-13 0
-12 0
-11 0
-10 0
-9 0
-8 0
-7 0
-6 0
-5 5
-4 5
-3 30
-2 217
-1 537
0 2565 <-- Bin that contains exactly the AVG
1 7896
2 16336
3 145175
4 672934
5 124166
6 18734
7 6667
8 2357
9 1330
10 339
11 262
12 323
13 89
14 25
15 8
END of Test
```

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Fig 7. Sample console output (using the LPCXpresso debugger)

4. Conclusion

Designing a system using a 10-bit ADC with a single power supply will face issues unless noise cancellation techniques are applied in the PCB design. The PCB's trace layout should follow the guidelines as suggested throughout this application note. This will help to keep noise interface to a minimum, particularly while an ADC conversion is in progress.

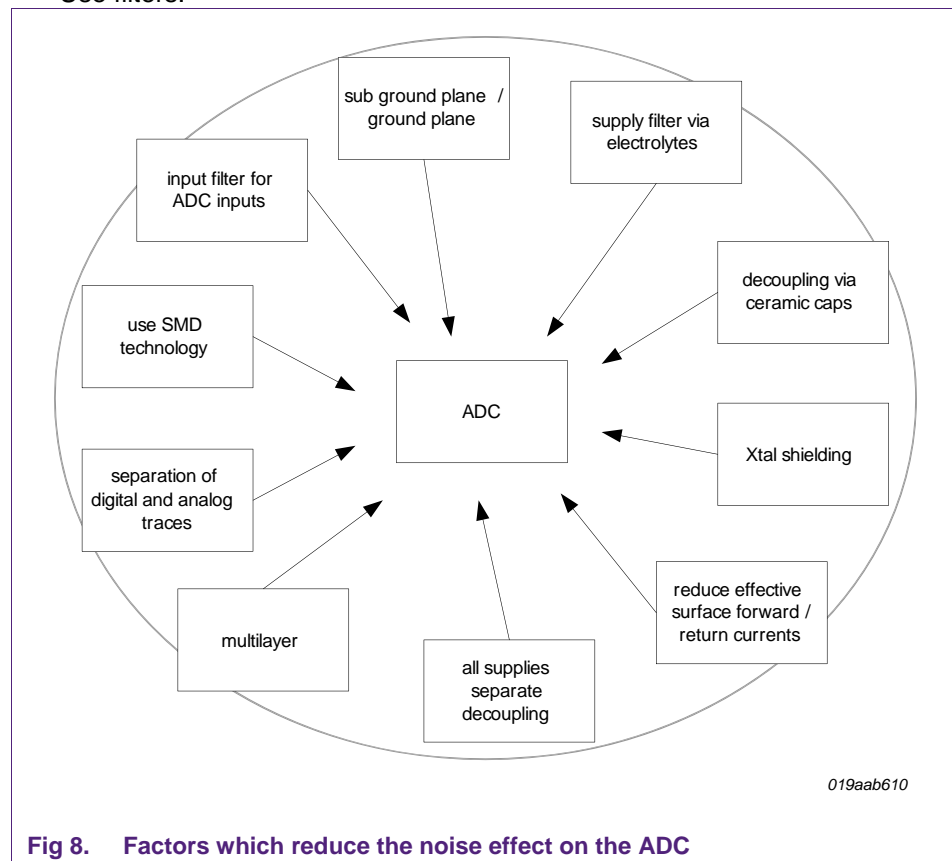
Some of these techniques include separating the ground planes, adding filters, keeping forward and return loops as small as possible, and strategically placing decoupling capacitors as close as possible to the supply pins.

The included application software is a simple and quick method to give a visual check on how noise can affect the ADC input's signal quality. The ideal histogram pattern shows a narrow distribution pattern indicating the ADC's ability to make precise conversions.

5. Attachment: Checklist of PCB design considerations

The following is a list of items to beware of:

- In the schematic show the 'senders and receivers' of noise.
- Look at Emission: How the design may affect the environment.
- Look at Immunity: How the environment may affect the design.
- Use the lowest clock frequency required.
- Split analog and digital parts into isolated segments.
- Use multiple layers, e.g., ground plane(s) and power plane(s) instead of traces.
- Group functional components together.
- Keep the traces from the component to the ground plane as short as possible.
- Shield clock signals.
- Make power lines short and thick.
- Keep signal and return traces above/below each other.
- Place proper decoupling capacitors as close as possible to IC's and planes.
- Shield sensitive signals using ground planes or grounded tracks.
- Use filters.



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