

Generic Tone Generation on the StarCore™ SC140/SC1400 Cores

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The use of DSPs in computationally intensive applications, such as media gateways, voice over packet networks (VoIP, VoFR, VoATM), and xDSL modems, has resulted in efficient designs and systems with increasing computational power. Most communication systems require precise tone generation and detection for functions such as dual tone multiple frequency (DTMF) signaling, call progress tones, and FAX requesting tones. The ITU-T Recommendation E.180/Q.35 [1] specifies the technical characteristics of tones used in most telephone services available today. Most off-the-shelf components do not meet the cost, size, and power consumption requirements of many systems. However, all the required signaling tones rely on the same basic principle, which can be efficiently implemented in today's DSPs. The use of DSPs enables many unrelated tone protocols to be synthesized with the same firmware and used across a wide range of tone frequencies.

This application note presents the design, implementation, and testing of a generic tone generator on the Freescale MSC8101 processor. The first member of the Freescale MSC8100 family of DSPs, the MSC8101 processor [2] is based on the StarCore™ SC140 four-ALU DSP core. This device has 512 KB of memory and a Communications Processor Module (CPM), which make it a versatile device for communication applications. The proposed module has two major components: initialization (`fsl_tone_gen_init`) and generation (`fsl_tone_gen`). The first component relies on polynomial approximation of a cosine for calculating initial states and coefficients to be used in the second component. The core of the

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tone generator (second component) relies on a digital oscillator. We tested this approach with a large number of signaling tones (such as DTMF, MF-R1, MF-R2), and it showed a good compromise in terms of performance, memory requirements, and precision. The module requires less than 0.1 MCPS on average, which can be improved with further hand-optimization of the assembly code. We also verified compliance with related ITU-T standards in terms of frequency accuracy, using both FFT magnitude plots of actual data generated from the module and real-time measurements with a spectrum analyzer.

1 Digital Sine Wave Generation

The most common ways to synthesize sine waves are:

- *Table look-up.* Offers the lowest complexity but is not used in the proposed tone generator because it cannot guarantee high precision with reasonable table sizes.
- *Polynomial approximation.* Offers the same high precision for every frequency, but requires the highest computational complexity.
- *Digital oscillator.* Uses intrinsic properties of trigonometric functions, which can be implemented with low memory and computational complexity, and usually exceeds precision requirements.

The proposed tone generator employs both the digital oscillator and polynomial approximation methods.

1.1 Digital Oscillator

The core of the tone generation module is a digital oscillator [3], that is, an IIR filter whose poles are on the unit circle. Our implementation uses a second-order filter, from which single-frequency tones are generated. The phase (θ) of the pole is related to the frequency of the synthesized sine wave:

$$\theta = 2\pi f / f_s$$

The equation for the oscillator is given as follows:

Equation 1

$$x(n) = 2 \cdot \cos\left(\frac{2\pi f}{f_s}\right) \cdot x(n-1) - x(n-2)$$

where $x(n)$ is a sine wave signal, f is the required frequency, and f_s is the sampling frequency.

The amplitude and phase of the sine wave are determined by the initial states $x(-1)$ and $x(-2)$. Depending on initial conditions, either a cosine or a sine signal is generated by a digital oscillator. This algorithm has two main advantages:

- Low data and program memory requirements
- Low computational complexity

A digital oscillator requires the following initializations:

- The coefficient

$$2 \cdot \cos(2\pi f / f_s)$$

- The two initial conditions: $x(-1)$ and $x(-2)$

The following table shows the computation of initial conditions:

COSINE Initialization	SINE Initialization
$x(-2) = \pm amplitude$ $x(-1) = \pm amplitude \times \frac{coefficient}{2} = amplitude \times \cos\left(\frac{2\pi f}{f_s}\right)$	$x(-2) = \pm amplitude \times \sin\frac{2\pi f}{f_s} = amplitude \times \cos\left(\frac{2\pi f}{f_s} - \frac{\pi}{2}\right)$ $x(-1) = 0$

For a pre-defined number of target frequencies, the initialization values can be pre-computed. However, a generic tone generation requires the computation with high precision for any frequency. A polynomial approximation, described in the following section, is employed for this calculation.

1.2 Polynomial Approximation

A polynomial approximation is used to compute the cosine necessary to initialize the digital oscillator. This method uses the Taylor's expansion of a cosine:

Equation 2

$$\cos(x) = \sum_{k=0}^{\infty} a_k x^{2k}, a_k = \frac{(-1)^k}{(2k)!}$$

Replacing x with the following:

$$2\pi f/f_s$$

and defining the normalized frequency as follows yields **Equation 3**:

$$f_n = f/f_s$$

Equation 3

$$\cos(2\pi f_n) = \sum_{k=0}^{\infty} b_k f_n^{2k}, b_k = (2\pi)^{2k} a_k$$

The actual implementation uses a factorized version of this formula with a finite number of elements and normalized coefficients. A sixth-order polynomial provides sufficient precision for 16-bit coefficients; furthermore, the same polynomial can be used to compute either sine or cosine, as follows:

$$\sin(x) = \cos\left(x - \frac{\pi}{2}\right)$$

Thus, the final polynomial approximation is as shown in **Equation 4**:

Equation 4

$$\cos(2\pi f_n) = 2^m (((c_3 f_n^2 + c_2) f_n^2 + c_1) f_n^2 + c_0), c_k = 2^{-m} b_k$$

The corresponding normalized coefficients to be stored in memory are scaled by 2^{-7} (that is, $m = 7$) and listed in **Table 1**.

Table 1. Normalized Coefficients

Coefficients	Fractional Value	16-bit Fixed Point
C ₀	0.0078125	256 (0x0100)
C ₁	-0.1542053	60483 (0xEC43)
C ₂	0.5073242	16624 (0x40F0)
C ₃	-0.6676025	43660 (0xAA8C)

2 Tone Generator Implementation

The proposed tone generator synthesizes tones composed of up to four simultaneous frequencies with arbitrary duration and frequency values. The method of defining tones is easy and flexible. The output signal is decomposed into elementary components. **Figure 1** shows an example of a possible desired tone.

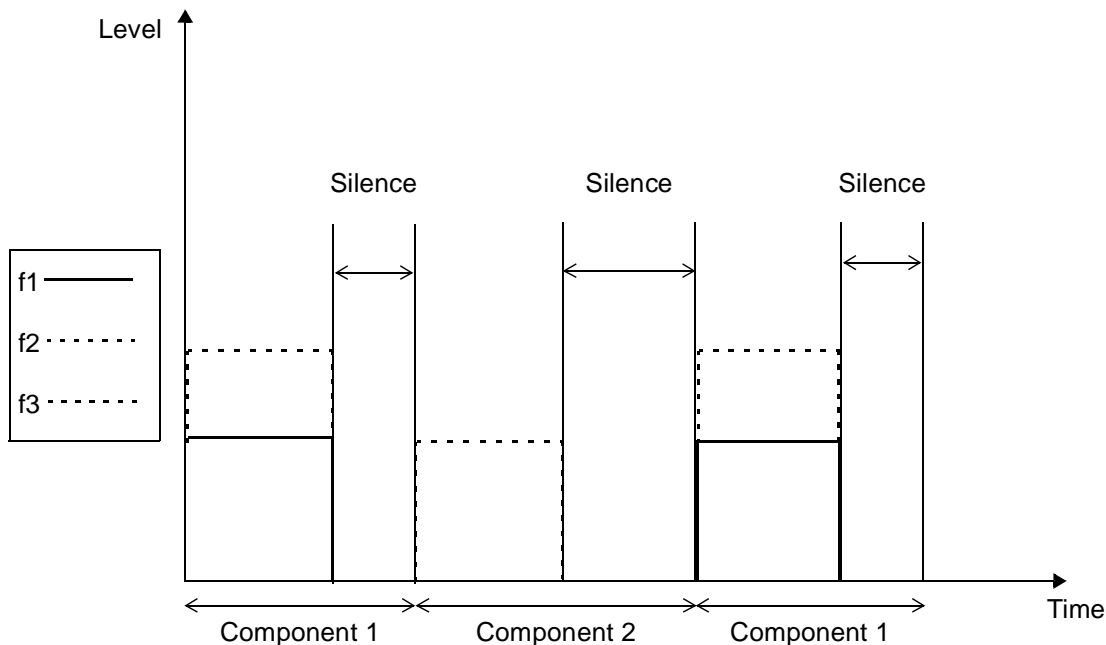


Figure 1. Example of a Desired Tone

2.1 Elementary Components

The elementary tone component is a pair composed of a sound period followed by a silence period (see **Figure 2**). This pair (sound + silence) can be repeated as often as needed to yield a component. In the example shown in **Figure 1**, each component has only one consecutive occurrence of its base pair.

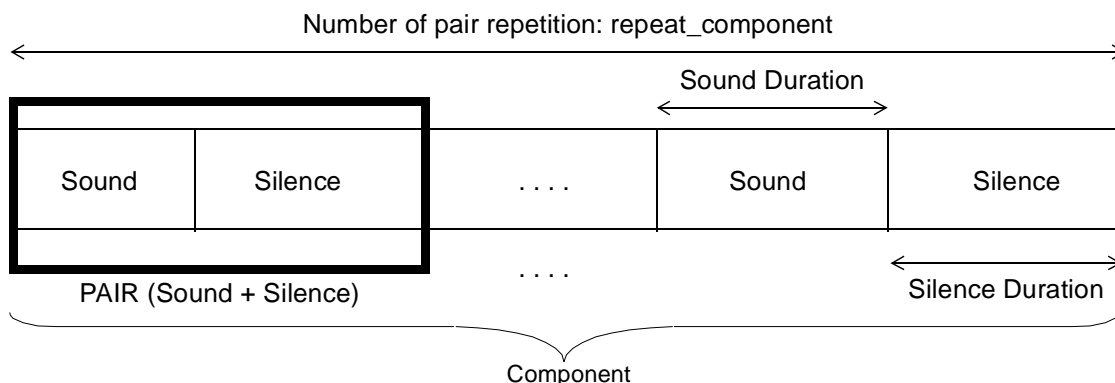


Figure 2. Elementary Tone Component

Each component is defined by the following parameters:

- *Frequencies* (in Hz).
- *Levels* (in dBm0). The signal amplitude per frequency is given as follows:

$$A = 10^{-Level(dBm0)/20} / \sqrt{2}$$

- *Sound duration*. The duration of the sound (in ms).
- *Silence duration*. The duration of the silence period (in ms) after the sound period. The software fills in the output buffer with zeroes to cover that period. A value of zero indicates no silence period.
- *repeat_component*. This value specifies the number of pairs (sound + silence) that must be generated to complete the component. A value of zero indicates that the pair should be repeated indefinitely.
- *Control word*. Defines whether the two sinusoids are to be added or modulated, the initial phase (0 or 180°), and the frequency unit (f or f/3). This flag is stored in 16 bits as follows:

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
—	—	—	—	—	—	P4	P3	P2	P1	U4	U3	U2	U1	C	M

- Bit 0: Add or modulate:
 - 0 The four frequencies are added.
 - 1 The four frequencies are modulated in pairs and then added (that is, $f_1 \times f_2 + f_3 \times f_4$).
- Bit 1: AM modulation type:
 - 0 AM modulation without carrier.
 - 1 AM modulation with carrier.
- Bit 2: Frequency unit 1:
 - 0 Frequency1.
 - 1 Frequency1 divided by 3.
- Bit 3: Frequency unit 2:
 - 0 Frequency2.
 - 1 Frequency2 divided by 3.

- Bit 4: Frequency unit 3:
 - 0 Frequency3.
 - 1 Frequency3 divided by 3.
- Bit 5: Frequency unit 4:
 - 0 Frequency4.
 - 1 Frequency 4 divided by 3.
- Bit 6: Phase 1:
 - 0 Frequency1 initialized with 0° phase.
 - 1 Frequency1 initialized with 180° phase.
- Bit 7: Phase 2:
 - 0 Frequency2 initialized with 0° phase.
 - 1 Frequency2 initialized with 180° phase.

The unit bits produce frequencies with 1/3 Hz. For example, 16 2/3 Hz is generated with 50 Hz when the unit bit is set (16 2/3 Hz = 50/3 Hz).
- Bit 8: Phase 3:
 - Frequency3 initialized with 0° phase.
 - Frequency3 initialized with 180° phase.
- Bit 9: Phase 4:
 - Frequency4 initialized with 0° phase.
 - Frequency4 initialized with 180° phase.

2.2 Tone Cycles

When all components are ready, they must be gathered into the desired tone. A component group is called a cycle, as shown in **Figure 3**. Up to six components can be grouped into one cycle. Finally, the created cycle can be repeated as often as needed to form the signaling tone.

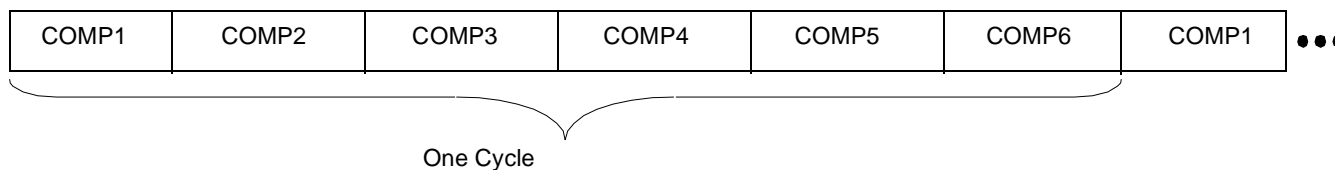


Figure 3. Cycle Components

2.3 Functional Interface

There are many different types of tones, such as single, dual, partial dual, and modulated. The starting times of sinusoids with different frequencies are arbitrary, but they are easily synchronized within the components. The external interface to the software consists of the following C functions (see **Figure 4**):

```
void fsl_tone_gen_init (GEN_CHANNEL *gen_channel, TONE_GENERIC *tone_control);
void fsl_tone_gen (GEN_CHANNEL *gen_channel, SIGNAL *linear_out, unsigned samples_out);
```

- *TONE_GENERIC* is a structure that describes the tone to be generated (tone descriptor).
- *SIGNAL* is a 16-bit signed fractional type.
- *linear_out* is an output buffer.
- *samples_out*, a multiple of eight, indicates how many samples the software must generate in this call.
- *GEN_CHANNEL* is a structure containing the description of the tone to be generated and the actual internal states of the generation process.
- *gen_channel* is a channel-dependent instance of *GEN_CHANNEL* containing all channel data for generation.

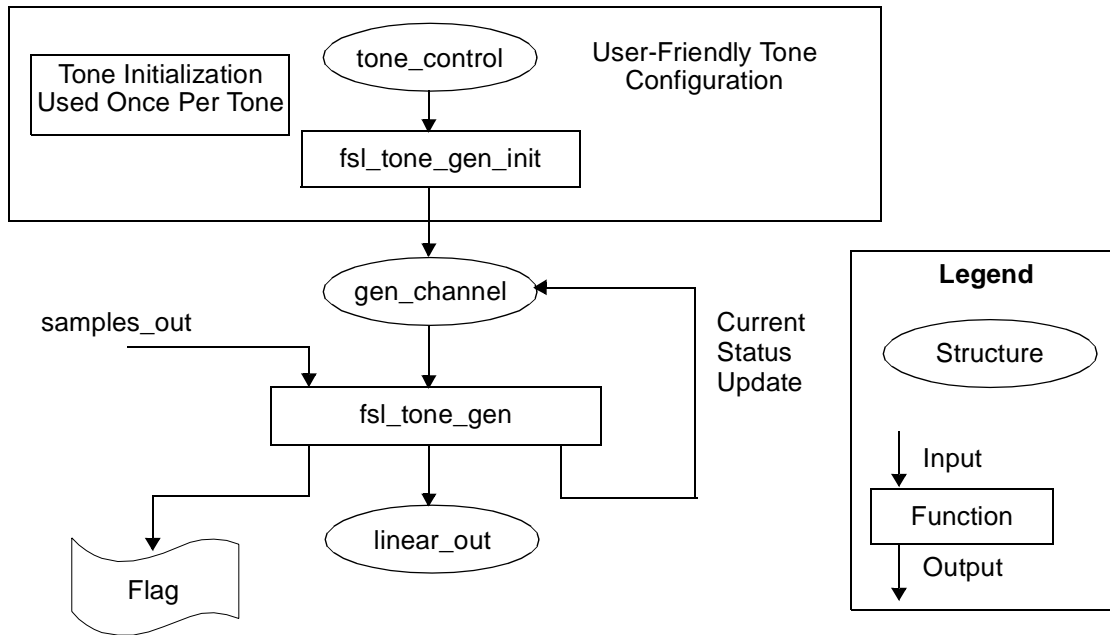


Figure 4. External Interface to Software

fsl_tone_gen returns the following integer flag:

- Bit 0 (LSB): Indicates whether tone generation is complete.
 - 0 Generation is not complete.
 - 1 Generation is complete.
- Bit 1: Indicates whether the content of the entire buffer is silence.
 - 0 Buffer or part of the buffer contains sound.
 - 1 Entire buffer is silence

The main structures are defined as follows:

```

typedef struct {
    UINT32 reserved;
    UINT16 tone_repeat_cycle;
    UINT16 tone_component_count;
    TONE_COMPONENT tone_component[6];
} TONE_GENERIC;
  
```

```
typedef struct {
    UINT32 sound_duration;
    UINT32 silence_duration;
    UINT16 flags;
    UINT16 repeat_component;
    UINT16 frequency[4];
    UINT8 level[4];
} TONE_COMPONENT;
```

where:

- *tone_repeat_cycle*. Number of cycle repetitions needed to generate the full tone.
- *tone_component_count*. Number of components in the cycle.
- *tone_component[6]*. An array that contains the description of every component.

TONE_COMPONENT parameters are described at the beginning of this section. The TONE_GENERIC structure is not persistent, so the `fsl_tone_gen_init` function does not save pointers to any part of this structure. After `fsl_tone_gen_init` returns, the software copies any information that is needed in the channel-dependent structure `gen_channel`. If the M bit is set to 1, the level of the smallest frequency in a given pair ((f1, f2) or (f3, f4)) represents the AM modulation depth in percent; the largest frequency is the carrier.

2.4 Implementation Constraints

The implementation constraints apply to both tone initialization and tone generation.

2.4.1 Tone Initialization

The coefficients for all frequencies are calculated with a cosine polynomial approximation during the initialization routine, which restricts the coefficient precision to 14 bits. We tested the error introduced by this method with MATLAB simulations and found the error to be within the one percent tolerance required by ITU-T standards. We decreased the full-scale level to 0x7FC0 (instead of 0x7FFF) to avoid saturation. The impact of such an “under-level” is negligible after A- or μ -law encoding stages [4]. This 0 dB level can be considered as a full scale one. Half scale tones are often preferred.

The initialization is divided into two parts:

- Creating a GEN_CHANNEL structure in conformance with the user's requirements, computing coefficients and duration for each component and copying any information needed after `fsl_tone_gen_init` returns.
- Updating the channel-dependent structure for every new component using an internal function, which updates the initial samples ($x(-2)$ and $x(-1)$).

Each time new samples are produced, the number of sound and silence samples is updated in `fsl_tone_gen`. When four frequencies are defined, they can either be added or modulated. Modulation is performed in pairs (that is, $f1 \times f2$, $f3 \times f4$) and the resulting modulated signals are added to produce the output signal.

2.4.2 Tone Generation

The SC140 multiple ALU architecture allows simultaneous generation of samples for both frequencies. The performance is optimal with two frequencies because required data can be loaded and updated in a single clock cycle: four fractional moves for the four previous samples and two fractional moves for the two coefficients.

To reduce MCPS, blocks of eight samples are generated. Furthermore, duration times are given in ms (number of samples is $8 \times (\text{duration time})$) so that eight samples are generated as grouped bursts. Each component includes a silence duration after a sound period. If a property of the tone must be changed without prior silence, the silence duration must be set to zero at the end of the previous component.

If a frequency level is 63 (–63 dBm0), the initial filter states are initialized with zeroes so that no samples are generated for this frequency. If a single frequency is generated with the modulation flag set, the result is silence. When a continuous tone must be generated, it is better to repeat a component with a long `sound_duration` because the code runs faster than with a short `sound_duration` repeated very often. Note that the filter state is initialized between two cycles, but not between two occurrences of the same component without silence in between. If only one component is used, it is better to repeat the component, not the tone cycle, to prevent harmonic distortions.

2.5 Status Reporting

The top-level modules are informed of the tone generation status with the return values given by `fs1_tone_gen`. The return flag bits are as follows:

- Bit 0 (LSB): indicates whether tone generation is complete.
 - 1 Complete.
 - 0 Not complete.
- Bit 1: indicates whether the content of the buffer is silence.
 - 1 Entire buffer is silence.
 - 0 Buffer or part of the buffer contains sound.

We recommend that silence be generated with the silence part of a component, not with the sound part initialized with null level (–63 dBm0) frequencies. Otherwise, the return flag does not reflect reality.

The “entire buffer is silence” flag detects whether the program has run the generation routine. If the generated sound is composed of null-level frequencies, the buffer contains only silence, but the returned flag indicates that sound has been generated. This approach saves MCPS in return flag computation (for testing every sample that would not be used).

2.6 Tone Generator Code Structure

Figure 5 shows an overview of the tone generation code structure. The main goal of the tone generation function is to generate the tone when required and update the associated state description structure.

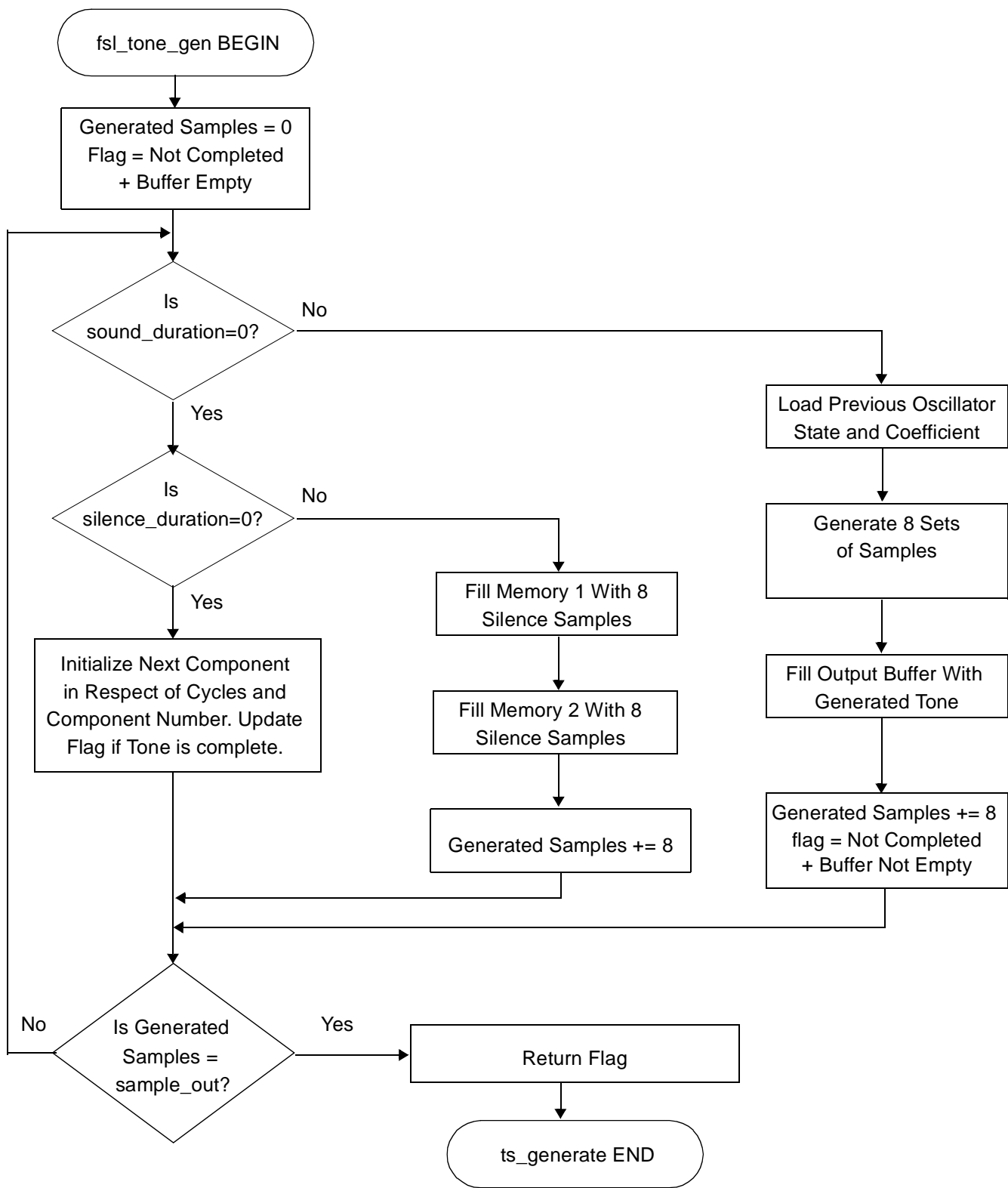


Figure 5. Structure of the Tone Generator Code

Example 1. Tone Generator Pseudo Code

```

Function name:      fsl_tone_gen
Input:             Pointer to output
                  Pointer to buffer containing static variables
                  Number of samples to generate
Output:           Return Flag
BEGIN
    FOR (Number of sample to generate/8)
        IF (sound duration is expired)
            IF (silence duration is not expired)
                Fill memory with 8 silence samples
            ELSE
                Initialize next component with respect to cycles
                and component number.
                Update flag if tone is complete.
            END IF
        ELSE
            FOR (8 couples of samples)
                Load previous oscillator states and coefficients
                Generate individual tone frequencies
                Fill output buffer with generated tone
                Update flag (sound generated)
            END FOR
        END IF
    END FOR
    RETURN (Flag)
END BEGIN

```

3 Using and Testing the Generic Tone Generator

We integrated and tested the generic tone generator on the MSC8101 Application Development System (MSC8101ADS) board [2] to confirm MATLAB simulation results. Many different tones were measured.

3.1 Test Set-up

On the MSC8101ADS, an audio codec is connected to an MSC8101 device through one of its CPM multi-channel controllers (MCCs), which configures the codec, and one of its serial peripheral interfaces (SPIs), which sends audio samples out of the ADS. **Figure 6** shows a diagram of this testing set-up. Alternatively, the ADSFS set-up [5] can be used, allowing generated samples to be written into files using the Ethernet port with a high transfer rate. In this case, the codec is not used.

3.2 Initialization Examples of Multi-Frequency Tones

The examples shown here provide specific configurations of the initialization function to generate several sets of tones composed by one or two frequencies. Because frequency values, power levels, and durations may vary from one case to another, the generic character of the generator is evident. Modifying the code to other types of signaling tones should be a fairly simple task.

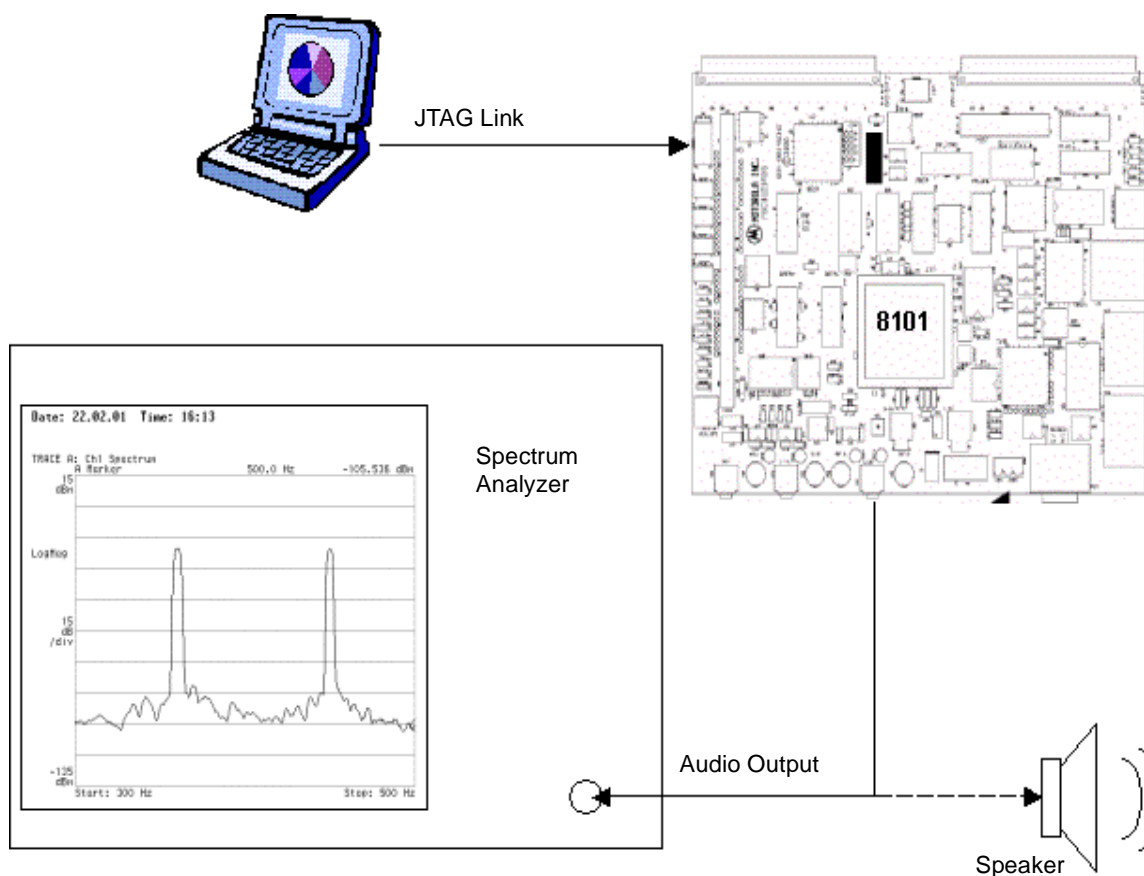


Figure 6. Testing Set-Up

Example 2. DTMF

```

UINT16 DTMF1[] = {697, 770, 852, 941};
UINT16 DTMFh[] = {1209, 1336, 1477, 1633};
void
test_initialize(TONE_GENERIC * testcases)
// testcases must point to a 16-item TONE_GENERIC structure
{
    int i, j;
    for (i = 0; i < 16; i++) {
        // Inialize the input structures for the i+1 key
        testcases[i].tone_repeat_cycle = 1;
        testcases[i].tone_component_count = 1;
        // 1 tone_component description
        testcases[i].tone_component[0].flags = 0x0;
        testcases[i].tone_component[0].repeat_component = 1;
        testcases[i].tone_component[0].sound_duration = 50;
        testcases[i].tone_component[0].silence_duration = 50;
        testcases[i].tone_component[0].level[0] = 10;
        testcases[i].tone_component[0].level[1] = 10;
        testcases[i].tone_component[0].level[2] = 63;
        testcases[i].tone_component[0].level[3] = 63;
    }
}

```

```

// DTMF digits
// Key:  1| 2| 3| 4| 5| 6| 7| 8| 9| 0| *| #| A| B| C| D
// Code: 1| 2| 3| 5| 6| 7| 9|10|11|14|13|15| 4| 8|12|16
for (i = 0; i < 4; i++) {
    for (j = 0; j < 4; j++) {
        testcases[4 * i + j].tone_component[0].frequency[0] = DTMFf[i];
        testcases[4 * i + j].tone_component[0].frequency[1] = DTMFh[j];
        testcases[4 * i + j].tone_component[0].frequency[2] = 0;
        testcases[4 * i + j].tone_component[0].frequency[3] = 0;
    }
}
}

```

Example 3. MF-R1

```

UINT16 R1[] = {700, 900, 1100, 1300, 1500, 1700};
void
test_initialize(TONE_GENERIC * testcases)
// testcases must point to a 15-item TONE_GENERIC structure
{
    int i, j, k = 0;
    for (i = 0; i < 15; i++) {
        // Inialize the input structures
        testcases[i].tone_repeat_cycle = 1;
        testcases[i].tone_component_count = 1;
        // 1 tone_component description
        testcases[i].tone_component[0].flags = 0x0;
        testcases[i].tone_component[0].repeat_component = 1;
        testcases[i].tone_component[0].sound_duration = 50;
        testcases[i].tone_component[0].silence_duration = 50;
        testcases[i].tone_component[0].level[0] = 10;
        testcases[i].tone_component[0].level[1] = 10;
        testcases[i].tone_component[0].level[2] = 63;
        testcases[i].tone_component[0].level[3] = 63;
    }
    // MF-R1 digits
    // Key:  1| 2| 3| 4| 5| 6| 7| 8| 9| 0| ST3P| STP| KP| STP2| ST
    // Code: 1| 2| 3| 4| 5| 6| 7| 8| 9|10| 11| 12| 13| 14| 15
    for (j = 1; j < 6; j++) {
        for (i = 0; i < j; i++) {
            testcases[k].tone_component[0].frequency[0] = R1[i];
            testcases[k].tone_component[0].frequency[1] = R1[j];
            testcases[k].tone_component[0].frequency[2] = 0;
            testcases[k++].tone_component[0].frequency[3] = 0;
        }
    }
}
}

```

Example 4. MF-R2

```

UINT16 R2f[] = {1380, 1500, 1620, 1740, 1860, 1980};
UINT16 R2b[] = {1140, 1020, 900, 780, 660, 540};
void
test_initialize(TONE_GENERIC * testcases)
// testcases must point to a 30-item TONE_GENERIC structure
// First 15 for MF-R2 foward direction
// Last 15 for MR-R2 backward direction
{
    int i, j, k = 0;

```


Example 6. Japan Waiting Tone

```
void test_initialize(TONE_GENERIC * testcases)
{
    // Initalize the input structures
    testcases[0].tone_repeat_cycle = 0;
    testcases[0].tone_component_count = 2;
    // 2 tone_component description
    // COMPONENT 1
    testcases[0].tone_component[0].flags = 0x0;
    testcases[0].tone_component[0].repeat_component = 1;
    testcases[0].tone_component[0].sound_duration = 100;
    testcases[0].tone_component[0].silence_duration = 100;
    testcases[0].tone_component[0].frequency[0] = 440;
    testcases[0].tone_component[0].level[0] = 0;
    // Only one frequency is used - level is initialized to -63db
    testcases[0].tone_component[0].frequency[1] = 0;
    testcases[0].tone_component[0].level[1] = 63;
    testcases[0].tone_component[0].frequency[2] = 0;
    testcases[0].tone_component[0].level[2] = 63;
    testcases[0].tone_component[0].frequency[3] = 0;
    testcases[0].tone_component[0].level[3] = 63;
    // COMPONENT 2
    testcases[0].tone_component[1].flags = 0x0;
    testcases[0].tone_component[1].repeat_component = 1;
    testcases[0].tone_component[1].sound_duration = 100;
    testcases[0].tone_component[1].silence_duration = 1000;
    testcases[0].tone_component[1].frequency[0] = 440;
    testcases[0].tone_component[1].level[0] = 0;
    // Only one frequency is used - level is initialized to -63db
    testcases[0].tone_component[1].frequency[1] = 0;
    testcases[0].tone_component[1].level[1] = 63;
    testcases[0].tone_component[1].frequency[2] = 0;
    testcases[0].tone_component[1].level[2] = 63;
    testcases[0].tone_component[1].frequency[3] = 0;
    testcases[0].tone_component[1].level[3] = 63;
}
```

3.3 Tests

The following MATLAB script displays the FFT of generated tones. A Hanning windowed FFT is used to measure the output frequency. A typical frequency accuracy requirement is about $\pm 1\%$ of the nominal frequency. At an 8 kHz sampling rate, we must use at least 4096 samples (512 ms) to ensure that the error generated by the FFT is not prejudicial.

Example 7. Frequency Response of Generated Tones

```

disp('*** Generic Tone Generation on StarCore SC140 ***')
disp(' '), disp(' ')
REF = '.wav'; TONE_NAME = 'DTMF'; BITS = 16;
SAMPLES = 8 * 50; SILENCE = 8 * 50; SPACE = SAMPLES + SILENCE;
if TONE_NAME == 'DTMF'
    number_of_digits = 16; fid = fopen(['gen_dtmf' REF], 'r');
elseif TONE_NAME == 'MFR1'
    number_of_digits = 15; fid = fopen(['gen_mfr1' REF], 'r');
elseif TONE_NAME == 'MFR2'
    number_of_digits = 30; fid = fopen(['gen_mfr2' REF], 'r');
else
    number_of_digits = 0; disp('Unknown tone.')
end
if number_of_digits > 0
    x = fread(fid, inf, 'short');
    disp(['Playing ' TONE_NAME ' generated tones...'])
    wavplay(x, 8000);
    for n = 1 : number_of_digits
        disp(['Plotting ' TONE_NAME ' tone ' int2str(n) '...'])
        out = x(SPACE + (n - 1) * (SAMPLES + SILENCE + SPACE) + ...
            (1 : SAMPLES)) .* 2 ^ -(BITS - 1);
        abscisse = linspace(0, 4000, SAMPLES / 2);
        fft_out = 20*log10( abs( fft( hanning( SAMPLES) .* out)));
        fft_out = fft_out - max(fft_out);
        [a, b] = sort(-fft_out ( 1 : SAMPLES / 2));
        [freq1, c] = sort(round(abs( abscisse(1, b(1:2)))));
        figure, subplot(211)
        plot(out), grid on
        xlabel('Sample #')
        title([TONE_NAME ' Tone ' int2str(n)])
        subplot(212), plot(abscisse, fft_out(1 : SAMPLES / 2))
        grid on, hold on
        plot(freq1, -a(c(1)), 'bo'), plot(freq1(2), -a(c(2)), 'ro')
        legend(['FFT of Tone ' int2str(n)], ...
            ['F_1 = ' int2str(freq1(1)) ' Hz'], ...
            ['F_2 = ' int2str(freq1(2)) ' Hz'])
        xlabel('Frequency (Hz)'), ylabel('Magnitude (dB)')
    end
end
disp('Test complete.')
fclose(fid);

```


Example 8. Driver for Generating Time Domain Samples Using ADSFS [5]

```

#include "fsl_tone_gen.h"
#define TBUFSZ 4096
#define DTMF
// #define MFR1
// #define MFR2
#ifdef DTMF
#define NUMBER_OF_TONES 16
#endif
#ifdef MFR1
#define NUMBER_OF_TONES 15
#endif
#ifdef MFR2
#define NUMBER_OF_TONES 30
#endif
void test_pgm ()
{
    fract16 samples[TBUFSZ];
    #pragma align samples 8
        char samples2[10];
        Word32 txid; // to write samples in text files
        char file_name[20];
        int channel, len;
        TONE_GENERIC testcases[NUMBER_OF_TONES];
        GEN_CHANNEL gen_channel[1];
        int flag;
        test_initialize(testcases);
        for (channel = 0; channel < NUMBER_OF_TONES; channel++) {
            // Opens a file
            sprintf(file_name, "c:\\gen_out%d.dat", channel);
            txid = open_file (file_name, WRITEONLY);
            if (txid <= 0)
                error (1);
            // Tone generation
            fsl_tone_gen_init (gen_channel, &testcases[channel]);
            flag = fsl_tone_gen (gen_channel, samples, TBUFSZ);
            // Saves samples
            for (len=0; len<TBUFSZ; len++) {
                write_file(txid, (unsigned char *)samples2,
                    sprintf(samples2, "%d\n", samples[len]));
            }
            close_file (txid);
        }
}

```

3.3.1 DTMF Frequency Accuracy Test

The digital oscillator algorithm has reasonable frequency accuracy. The tests listed in **Table 2** validate the use of the digital oscillator algorithm for DTMF generation, showing that it complies with the requirements of the ITU-T Recommendation Q.24 [6] for frequency accuracy. This test uses 4096 samples, generated with a -20 dBm0 level (11 percent of maximum scale). The measured DTMF tone precision is smaller than 0.2 percent.

Table 2. Digital Oscillator Algorithm for DTMF Generation

Frequency (Hz)	Precision (Percent)
697	0.12
770	0.08
852	0.16
941	0.14
1209	0.03
1336	0.06
1477	0.05
1633	0.07

3.3.2 MF-R2 Frequency Accuracy Test

The module capabilities were tested on the most constraining standard for frequency accuracy: MF-R2 signaling defined in the ITU-T Recommendation Q.454 [7]. A deviation of up to 4 Hz is tolerated on generated frequencies. Each frequency is measured with an 8192 point Hanning windowed FFT. At an 8 kHz sampling rate, the resulting accuracy is smaller than 1 Hz. In **Table 3** and **Table 4**, an error of 0 Hz means that the test is unable to detect any frequency deviation.

Table 3. Forward Frequencies Test

	Frequency (Hz)	Measured Frequency (Hz)	Error (Hz)	R2 Test
F0	1380	1379.6	0.4	PASS
F1	1500	1499.7	0.3	PASS
F2	1620	1619.8	0.2	PASS
F3	1740	1740.0	0.0	PASS
F4	1860	1859.1	0.9	PASS
F5	1980	1979.2	0.8	PASS

Table 4. Backward Frequencies Test

	Frequency (Hz)	Measured Frequency (Hz)	Error (Hz)	R2 Test
F0	1140	1139.3	0.7	PASS
F1	1020	1019.2	0.8	PASS
F2	900	900.0	0.0	PASS
F3	780	779.9	0.1	PASS
F4	660	659.7	0.3	PASS
F5	540	539.6	0.4	PASS

All tests of the tone generator passed. Our results indicate that the proposed tone generation module complies with the requirements for frequency accuracy of the ITU-T Recommendation Q.454 for MF-R2.

3.4 Examples of Generated Signaling Tones

Figure 7 and Figure 8 illustrate the spectrum of two typical signaling tones. The figures on the left correspond to the FFT magnitude plots of actual data generated from the proposed module; the figures on the right correspond to measured spectra (using a spectrum analyzer). Observe that the noise floor of the measurements is around 80 dB below the peak values.

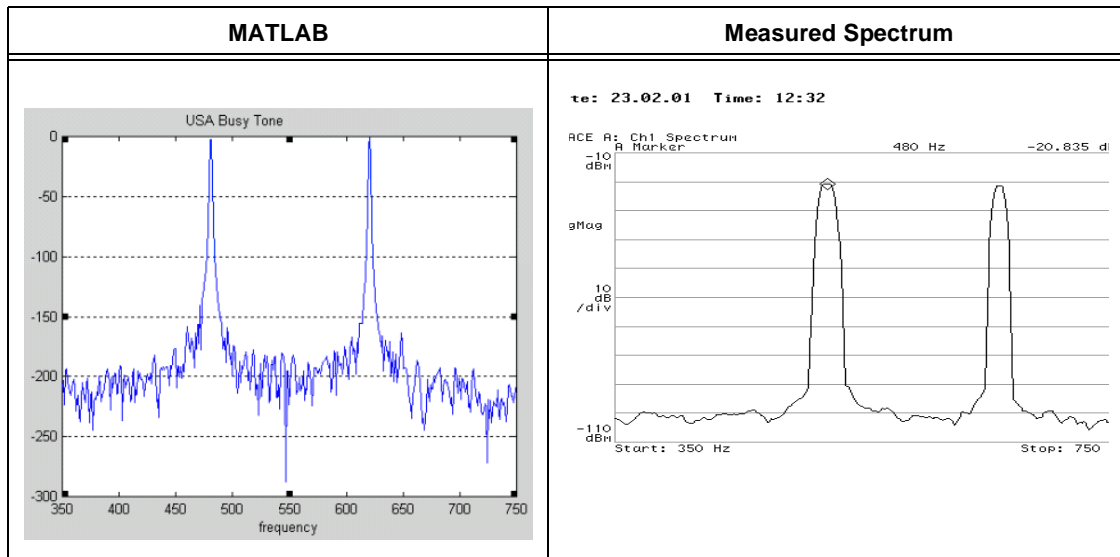


Figure 7. USA Busy Tone: 480 + 620 Hz

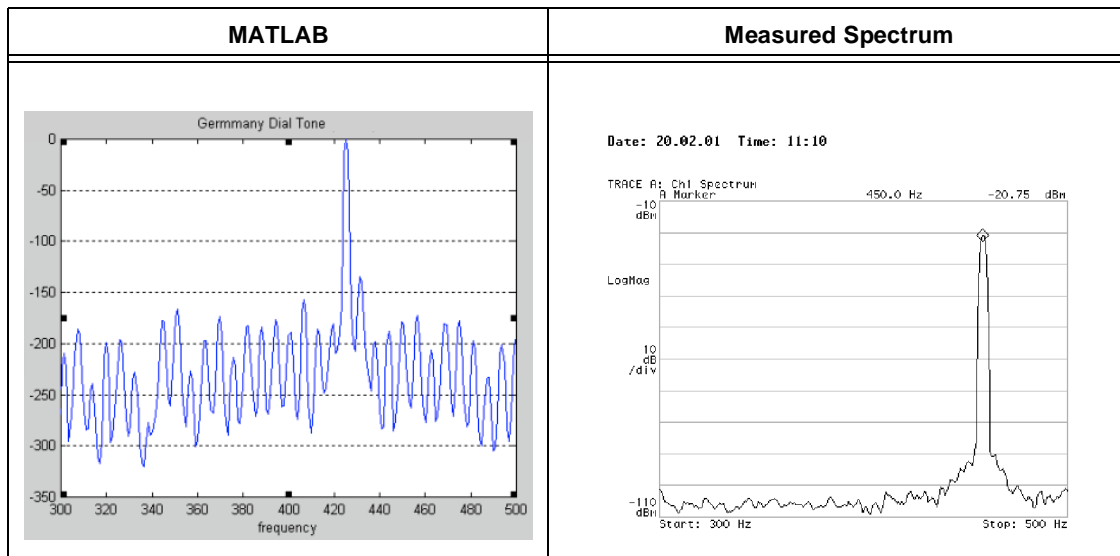


Figure 8. Germany Dial Tone: 425 or 450 Hz

3.5 Performance Measurements

The proposed tone generation module typically requires less than 0.15 MCPS. The worst case would be to change the component every millisecond, but this mode of operation is not used in practice (typical component change intervals are greater than 40 ms).

Table 5. Typical Average MCPS Values

Tone	Initialization (Cycles)	Generation (MCPS)
DTMF	945	0.073
Germany Dial Tone	560	0.106
Japan Waiting Tone	560	0.47

Table 6. Memory Usage

Data / Channel (Bytes)	Table Size (Bytes)	Stack Size (Bytes)	Program Size (Bytes)
272	156	24	2828

4 References

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- [4] *ITU-T Recommendation G.711, Pulse Code Modulation (PCM) of Voice Signals*, 1993.
- [5] *Fast I/O Library for the MSC8101ADS Using Ethernet Communication (AN2260)*, Freescale Semiconductor, Inc., 2004.
- [6] ITU-T Recommendation Q.24, Multi-frequency push bottom signal reception, 11/98.
- [7] ITU-T Recommendation Q.454, The Sending Part of the Multi-Frequency Signalling Equipment.

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