NXP Semiconductors Application Note

PT2000: Three injectors with freewheeling and DCDC

1 Introduction

This application note examines an example of a three cylinder internal combustion engine (ICE) injector drive. The high-voltage is provided by a boost converter managed by the PT2000. It describes a typical hardware topology and related software example to drive three injectors managed in three banks, allowing a full overlap and a DC/DC converter in resonant mode.

The PT2000 is supplied by a battery voltage between 9.0 V and 32 V. An external 5.0 V must be supplied to the VCC5 and the VCCIO pin, to internally supply the I/O buffers. In this example, the V_{CCP} voltage is internally generated to enable the drivers. If V_{BAT} voltage is set higher than 16 V then V_{CCP} has to be forced externally. The boost converter topology is defined to manage DC/DC. A Pi filter prevents circuitry disturbance propagation from the boost regulation area to battery line.

In this configuration, full overlap injection is possible, and to reduce power dissipation and show some capabilities of the PT2000 freewheeling low-sides are used instead of diode.

NXP analog ICs are manufactured using the SMARTMOS process, a combinational BiCMOS manufacturing flow integrating precision analog, power functions, and dense CMOS logic together on a single cost-effective die.

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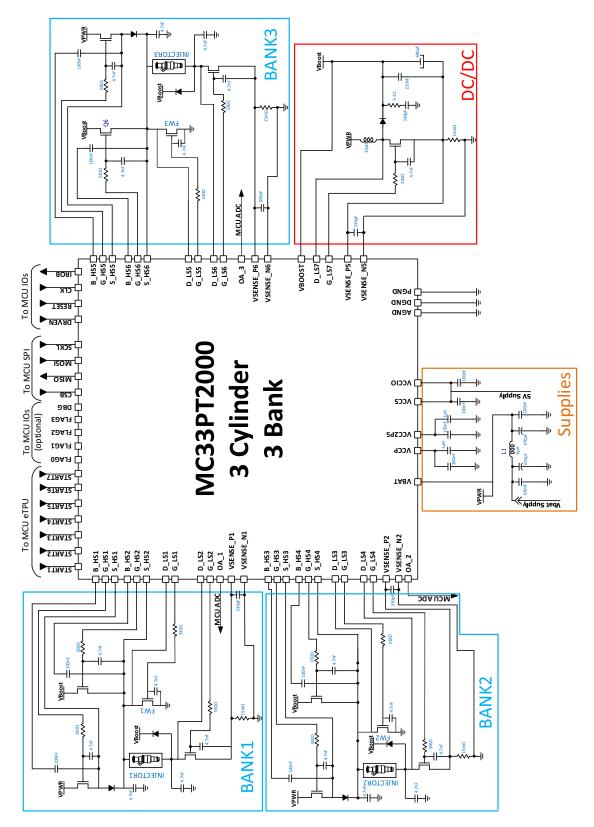


Figure 1. Typical four injector two bank application schematic

3 Application instructions

This topology can be used on the KITPT2000FRDM3C evaluation board. Register settings and microcode downloads can be achieved by using the KL25Z embedded on the KITPT2000FRDM3C. This topology manages three banks of three cylinders. Each bank is individually managed by one microcore, as follows:

- The bank # 1 is managed by the digital microcore Uc0Ch1
- The bank # 2 is managed by the digital microcore Uc1Ch1.
- The bank # 3is managed by the digital microcore Uc0Ch2.

The DCDC is managed by microcore Uc0Ch3. The following is the start-up sequence:

- Apply a battery voltage between 9.0 V and 16 V.
- Download the registers Channel Configuration, Main Configuration, IO Configuration, and then Diagnostic Configuration.
- Download the dedicated microcode in the Logic Channel 1, Logic Channel 2, and Logic Channel 3 Data RAM.
- Set to '1' to the pre-flash enable bit and the en_dual_seq bit in the Flash_enable register of channel 1 (0x100), channel 2 (0x120), and channel 3 (0x140).

The registers configuration and the microcodes are detailed in the following chapters.

Once the DC/DC converter output has reached its nominal voltage, the injector drivers can be actuated with the STARTx pins. Each STARTx pin individually triggers each injector pin rising edge and stops actuation on the falling edge.

- START1 drives INJECTOR 1
- START2 drives INJECTOR 2
- START3 drives INJECTOR 3

Boost regulation is stopped during the injection boost phase.

4 Software requirements

4.1 Current profile management for injection

The current profile is managed to generate an initial high current through the injector. This high current slew rate minimizes the opening delay. This high current is maintained during a given time to ensure injector opening, then is decreased to maintain the injector open, up to the End Of Injection (EOI).

The code dedicated to the injection is loaded into the Code RAM 1. This code is executed independently by the microcores Uc0Ch1 and Uc1Ch1. Each one of the microcores generate a current profile, as described through the injector per the STARTx pin state as illustrated in Figure 2.

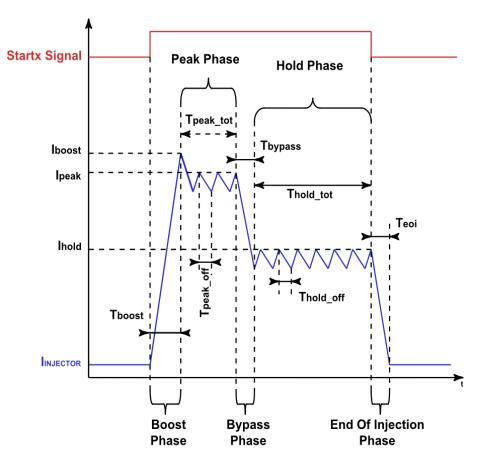


Figure 2. Typical peak and hold current profile

When a rising edge is detected on a STARTx pin, the injection starts with the injection Boost phase. This profile can be stopped at any time by detecting a falling edge on the STARTx pin. In this case, the End Of Injection phase is executed.

During the Boost phase, the corresponding low-side driver is simultaneously switched on with the high-side switch connected to the V_{BOOST} voltage. If the boost current target I_{BOOST} is reached, the high-side driver is switched off and the current recirculates for a fixed time (t_{PEAK} OFF) through the diode connected to ground, then the Peak phase starts.

During the Peak phase, the high-side switch connected to V_{BAT} voltage is turned on. If the I_{PEAK} current target is met, the high-side driver is switched off and the current recirculates through the diode connected ground for the fixed time (t_{PEAK_OFF}). The high-side driver is then switched on again. This cycle repeats until the internal counter reaches its terminal value (t_{PEAK_TOT}), then the Bypass phase begins.

During the Bypass phase, all low-side and high-side switches are turned off. The current decays through the injector, the diode connected to ground, and the diode connected to V_{BOOST} for a fixed time (t_{BYPASS}). The Hold phase then starts.

During the Hold phase, the low-side driver is simultaneously switched on with the high-side switch connected to the V_{BAT} voltage. If the hold current target I_{HOLD} is reached, the high-side driver is switched off and the current recirculates through the diode connected to ground for a fixed time (t_{HOLD_OFF}). The high-side driver is switched on again, and the cycle repeats until the STARTx pin goes low or the internal counter reaches its terminal value (t_{HOLD_TOT} (time out)). The End Of Injection is forced if no falling edge is detected on the STARTx pin.

All the current thresholds and timings are accessed in the Data RAM. The typical values are in Table 1., but must be defined according to the injector used and the injection profile expected.

Parameter name	Description	Value
I _{BOOST}	Current threshold in Boost phase	14 A
I _{PEAK}	Current threshold in Peak phase (Depends on injectors type)	6.0 A
I _{HOLD}	Current threshold in Hold phase	3.0 A
^t PEAK_OFF	Fixed time for high-side switch off in Peak phase	10 µs
^t PEAK_TOT	Fixed time for end of Peak phase	500 µs
t _{BYPASS}	Fixed time for Bypass phase	20 µs
t _{HOLD_OFF}	Fixed time for high-side switch off in Hold phase	10 µs
t _{HOLD_TOT}	Fixed time for end of Hold phase (timeout)	10 ms

In the present case, most of the code branches (jump) are managed according to the counters end of count and the current threshold, by the means of the wait table. The wait table rows are affected, as shown in Table 2., and are changed according to the injection phase.

Table 2. Example of wait table definition

Phase	Boost phase	Peak phase	Bypass phase	Hold phase	EOI phase
Row 1	If startx goes low then jumps to EOI phase	If startx goes low, then jumps to EOI phase	If startx goes low, then jumps to EOI phase	If startx goes low, then jumps to EOI phase	-
Row 2	If the injection current reaches I _{BOOST} then jumps to Peak phase	If t _{PEAK_TOT} is reached, then jumps to Bypass phase	If t _{BYPASS} is reached, then jumps to Hold phase	If $t_{\mbox{HOLD}\ \mbox{TOT}}$ is reached, then jumps to EOI phase	-
Row3	-	if t _{PEAK_OFF} is reached, jumps to Peak On phase (sub phase)	-	if t _{HOLD_OFF} is reached, jumps to Hold On phase (sub phase)	-
Row 4	-	if I _{PEAK} is reached, jumps to Hold Off phase (sub phase)	-	if I _{HOLD} is reached, jumps to Hold Off phase (sub phase)	-
Row 5	-	-	-	-	-

A rising edge issues an injection start. A falling edge triggers the end of injection. In case of an overlap between two STARTx pins on the same bank, it is manage by the Smart Start function of the device. In this case, the first STARTx rising edge is considered. The second STARTx pin high-state is consider when the first actuation is finished. The action of the injection corresponding to the second STARTx pin is stopped when the second STARTx pin falling edge occurs.

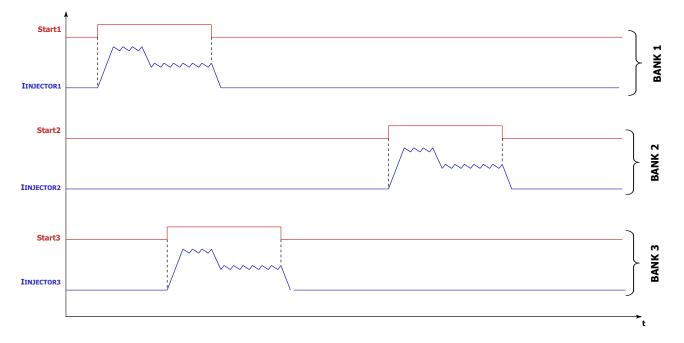


Figure 3. Actuation driven by STARTx pins with possible full overlap

4.1.1 SPI registers setup

The PT2000 registers are setup according their default states, unless defined by the following.

4.1.1.1 Main configuration registers

4.1.1.1.1 CLK configuration registers

A 1.0 MHz CLK has to be supplied from the main MCU to the PT2000. This CLK is then multiplied by a factor to generate the PLL, which is the clock for the microcore memories. In this example, the multiplicator factor chosen is 24, meaning a PLL at 24 MHz. This factor is set by PLL_Config register (1A7h).

Table 3. PLL_Config (1A7h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name		Reserved													PLL_spre ad_disabl e	PLL_fact or
Value	00000000000													0	1	

To run two microcores per channel, the **Clock_Prescaler** register (1A0h) must be setup with a ck_per value of 2 or 3. In this case, the main CLK ("ck") is set to 6.0 MHz. ck = ck_sys/(ck_per +1) = 24 MHz/(3+1) = 6.0 MHz

Table 4. Clock_Prescaler register (1A0h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name					Rese	erved				ck_	per					
Value		-											000	011		

4.1.1.1.2 Offset compensation CLK registers

To improve current accuracy offset compensation is enabled each time the microcore is in the idle state, which means the related start pin is low. To make this work, the CLK offset compensation must be set to a maximum of 500 kHz.

As shown in CLK configuration registers, page 6, the PLL ("ck_sys") is set to 24 MHz. To get a ck_ofscomp to 500 kHz. a prescaler set to 47d + 1d is needed.

Table 5. Ck_ofscmp_Prescaler(1A4h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name				Rese	erved							ck_ofsc	mp_per			
Value				0000	0000							0010	1111			

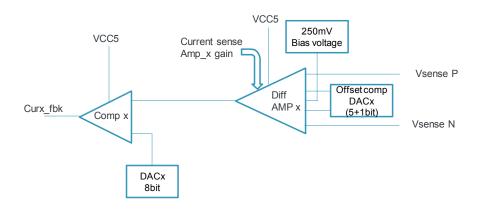


Figure 4. Offset compensation block diagram

As shown in Figure 4, a 5+1-bit DAC is use to compensate the offset. When the "stoc on" instruction runs the 8-bit DAC is automatically set to 253 mV and the 5+1 DAC for offset compensation tries to find the best value to get 250 mV. This is done during idle phase and shuts down as soon as the start pin goes high.

Each new offset compensation starts based on the result of the previous offset compensation run for this current measurement channel. If the offset compensation is stopped from the digital sequencer when the analog offset compensation is not finished, the procedure is aborted, maintaining the last compensation value reached when the procedure was interrupted. A full offset compensation takes 2.0 μ s (500 kHz) x 31 steps = 62 μ s.

4.1.2 Injection bank management register setup

The PT2000 registers are setup according their default states or setup automatically by the PT2000 IDE (i.e. checksum registers, code width), unless defined by the following.

4.1.2.1 IO configuration registers

These registers configure the crossbar switch to give access to output drivers, current sense to each microcore.

4.1.2.1.1 Output driver access

The Uc0Ch1 microcore must have access to the HS1, HS2, LS1, and LS2 pre-drivers.

	-				,											
Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	Acc_u c0_ch1 _ls8	Acc_u c0_ch1 _ls7	Acc_u c0_ch1 _ls6	Acc_u c0_ch1 _ls5	Acc_u c0_ch1 _ls4	Acc_u c0_ch1 _ls3	Acc_u c0_ch1 _ls2	Acc_u c0_ch1 _ls1	Reserv ed	Acc_u c0_ch1 _hs7	Acc_u c0_ch1 _hs6	Acc_u c0_ch1 _hs5	Acc_u c0_ch1 _hs4	Acc_u c0_ch1 _hs3	Acc_u c0_ch1 _hs2	Acc_u c0_ch1 _hs1
Value	0	0	0	0	0	0	1	1	0	0	0	0	0	0	1	1

Table 6. Out_acc_uc0_ch1 register (160h)

In the same way, the Uc1Ch1 must have access to the pre-drivers HS3, HS4, LS3, and LS4.

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Table 7. Out_acc_uc1_ch1 register (161h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	Acc_u c1_ch1 _ls8	Acc_u c1_ch1 _ls7	Acc_u c1_ch1 _ls6	Acc_u c1_ch1 _ls5	Acc_u c1_ch1 _ls4	Acc_u c1_ch1 _ls3	Acc_u c1_ch1 _ls2	Acc_u c1_ch1 _ls1	Reserv ed	Acc_u c1_ch1 _hs7	Acc_u c1_ch1 _hs6	Acc_u c1_ch1 _hs5	Acc_u c1_ch1 _hs4	Acc_u c1_ch1 _hs3	Acc_u c1_ch1 _hs2	Acc_u c1_ch1 _hs1
Value	0	0	0	0	1	1	0	0	0	0	0	0	1	1	0	0

The Uc0Ch1 microcore has default access to the current sense block # 1, and to microcore Uc1Ch1 to the current sense block # 2. The corresponding registers content doesn't need to be changed.

In the same way, the Uc0Ch2 must have access to the pre-drivers HS5, HS6, LS5, and LS6.

Table 8. Out_acc_uc0_ch2 register (162h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	Acc_u c1_ch1 _ls8	Acc_u c1_ch1 _ls7	Acc_u c1_ch1 _ls6	Acc_u c1_ch1 _ls5	Acc_u c1_ch1 _ls4	Acc_u c1_ch1 _ls3	Acc_u c1_ch1 _ls2	Acc_u c1_ch1 _ls1	Reserv ed	Acc_u c1_ch1 _hs7	Acc_u c1_ch1 _hs6	Acc_u c1_ch1 _hs5	Acc_u c1_ch1 _hs4	Acc_u c1_ch1 _hs3	Acc_u c1_ch1 _hs2	Acc_u c1_ch1 _hs1
Value	0	0	1	1	0	0	0	0	0	0	1	1	0	0	0	0

4.1.2.1.2 Current sense access

The following registers give access to current sense block to the specified microcore. The Uc0Ch1 microcore must have access to current sense 1 and Uc1Ch1 to current sense 2.

Table 9. Cur_block_access_part1 register (166h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	acc_u c1_ch 1_curr 6H_6N eg	acc_u c1_ch 1_curr 6L	acc_u c1_ch 1_curr 5H_5N eg	acc_u c1_ch 1_curr 5L	acc_u c1_ch 1_curr 4	acc_u c1_ch 1_curr 3	acc_uc 1_ch1_ curr2	acc_uc 1_ch1_ curr1	acc_uc 0_ch1_ curr6H _6Neg	acc_uc 0_ch1_ curr6L	acc_uc 0_ch1_ curr5H _5Neg	acc_uc 0_ch1_ curr5L	acc_uc 0_ch1_ curr4	acc_uc 0_ch1_ curr3	acc_uc 0_ch1_ curr2	acc_uc 0_ch1_ curr1
Value	0	0	0	0	0	0	1	0	0	0	0	0	0	0	0	1

The Uc0Ch2 microcore must have access to current sense 3.

Table 10. Cur_block_access_part2 register (167h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	acc_u c1_ch 2_curr 6H_6N eg	acc_u c1_ch 2_curr 6L	acc_u c1_ch 2_curr 5H_5N eg	acc_u c1_ch 2_curr 5L	acc_u c1_ch 2_curr 4	acc_u c1_ch 2_curr 3	acc_uc 1_ch2_ curr2	acc_uc 1_ch2_ curr1	acc_uc 0_ch2_ curr6H _6Neg	acc_uc 0_ch2_ curr6L	acc_uc 0_ch2_ curr5H _5Neg	acc_uc 0_ch2_ curr5L	acc_uc 0_ch2_ curr4	acc_uc 0_ch2_ curr3	acc_uc 0_ch2_ curr2	acc_uc 0_ch2_ curr1
Value	0	0	0	0	0	0	0	0	0	0	0	0	0	1	0	0

4.1.2.1.3 Freewheeling access

As mentioned previously, to reduce power dissipation and to show PT2000 capability, a freewheeling low-side is used instead of a diode. One of the features of the PT2000 is to control the FW low-side, automatically depending on the high-side command. Control of the freewheeling is done through microcore using the "stfw" instruction. There is no need to configure the Fw_external_request register (16Ah). High-side on V_{BAT} is the one used as a command, this high-side has to be ON even during the boost phase, to make sure the FW low-side is OFF. The following configuration is used in this example:

Table 11. Freewheeling link register

Freewheeling pre-driver output	Related pre-driver high-side
LS1	HS1
LS3	HS3
LS5	HS5

Table 12. Fw_link (169h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	Reserve d	Flag3_f w_link	Flag2_f w_link	Flag1_f w_link	Flag0_f w_link	Rese	erved	Hs7_f w_link	Reserv ed	Ls7_fw _link	Ls6_fw _link	Ls5_fw _link	Ls4_fw _link	Ls3_fw _link	Ls2_fw _link	Ls1_fw _link
Reset	0	0	0	0	0	0	0	0	0	0	0	1	0	1	0	1

The time between the high-side and FW low-side ("dead time") command is selectable by the SPI, to avoid cross conduction (refer to Table 20. High-side Output Configuration Register).

stfw instruction:

To automatically control the freewheeling low-side, the stfw auto instruction is used in the microcode. Shortcut 1 is considered as the high-side using freewheeling. In this case, HS1 uses LS1 as freewheeling the shortcut definition, for Uc0CH1 is: dfsct **hs1** hs2 ls2; in this case ls1 switches according to the hs1 command.

4.1.2.1.4 OAx settings (optional)

For safety purposes, it is possible to send an image of the current going through the load directly to the MCU through the OAx pins. The following settings map the current sense 1 to OA1 and current sense 2 to OA2.

Table 1. Oa_out1_config (197h)

Γ	Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
	Name		Reserved								oa1_g1		oa_sel1		oa1_	gain	oa1_en
	Value				0	0000000	0				0		000		0	0	1

Table 13. Oa_out2_config (198h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name					Reserved					oa2_g1		oa_sel2		oa2_	_gain	oa2_en
Value				0	0000000	0			0		000		0	0	1	

If this feature is mandatory in the application, special care needs to be taken during the selection of the current for each bank. For example, if it needs to monitor on the OAx pins of each current sense at the same time, the following current sense needs to be selected:

-Bank 1 : current sense 1 mapped to OA1

-Bank 2: current sense 2 mapped to OA2

-Bank3 : current sense 5 mapped to OA3

4.1.2.2 Channel 1-2 configuration registers

The following registers are used to configure each channel used to control injectors.

4.1.2.2.1 Startx pin sensitivity registers

The injectors 1, 2, and 3 are driven according to the logic level of their respective STARTx pin. A high level on STARTx triggers the activation of the corresponding injector. A low level on the STARTx pin automatically stops the actuation, whatever the injection phase. The Uc0Ch1microcore must be enabled by START1 pin, while the Uc1Ch1 microcore must be enabled by the START2 pin, and Uc0Ch2 must be enabled by START3 pin. Consequently, Start_config_reg_part1 register of the channel 1 (103h) must be setup, as shown in Table 14.

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
	start8_	start7_	start6_	start5_	start4_	start3_	start2_	start1_	start8_	start7_	start6_	start5_	start4_	start3_	start2_	start1_
Name	sens_u															
	c1	c0														
Value	0	0	0	0	0	0	1	0	0	0	0	0	0	0	0	1

Table 14. Start_config_reg_part1 registers (103h)

In the same way Start_config_reg_part1 register of the channel 2 (123h) must be setup, as shown in Table 15.

Table 15. Start_config_reg_part1 registers (123h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
	start8_	start7_	start6_	start5_	start4_	start3_	start2_	start1_	start8_	start7_	start6_	start5_	start4_	start3_	start2_	start1_
Name	sens_															
	uc1	uc0														
Value	0	0	0	0	0	0	0	0	0	0	0	0	0	1	0	0

4.1.2.2.2 Entry point registers

Entry point registers define where each microcore starts in each channel. With the new PT2000 IDE, it is possible to specify a label as a start point and the following registers are set automatically. For example the "init0" label is the starting point of Uc0Ch1. If this feature is not used, the following register has to be set manually. In the code example, the entry point of Uc0Ch1 is 0 as the first line executed, and is the first Code RAM line of the channel 1.

Table 16. Uc0_entry_point registers (10Ah) of channel 1

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name			Rese	erved						е	ntry_poin	t_addres	S			
Reset			000	000							00000	00000				

The code entry point of Uc1Ch1 is the 31h Code RAM line of the channel 1.

Table 17. Uc1_entry_point registers (10Bh) of channel 1

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name			Rese	erved						е	ntry_poin	nt_addres	S			
Reset			000	000							00001	10001				

The same configuration must be done for Channel2.

Table 18. Uc0_entry_point registers (12Ah) of channel 2

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name			Rese	erved						е	ntry_poin	t_addres	S			
Reset			000	000							00000	00000				

The code entry point of Uc1Ch1 is the 2Fh Code RAM line of the channel 1.

Table 19. Uc1_entry_point registers (12Bh) of channel 2

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name			Rese	erved						е	ntry_poir	t_addres	s			
Reset			000	000							00001	10001				

4.1.2.3 Diagnosis configuration registers

The high-side and low-side drivers must be directly controlled by the microcores. Consequently, the output_routing fields of the high-side and low-side driver's output configuration register must be set to the value 15. Dead time bits need to be set to avoid cross conduction between high-side and low-side freewheeling.

t_{FWDLY} = Tck x (Dead_time + 1) = 1/6 MHz x (3 + 1) = 668 ns

Table 20. Hsx_output_config registers (1DAh, 1DDh, 1E0h, 1E3h, 1E6h, 1ECh)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	HSx ovr	HSx Vds ref		reserved		dead_time output_routing							inv			
Value	0	0		000				00011					11111			0

Table 21. Lsx_output_config registers (1C2h, 1C5h, 1C8h, 1CBh, 1CEh, 1D1h, 1D4h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	LSx_o vr					Reserved	l					ou	tput_rout	ing		inv
Value	0				C	0000000	0						11111			0

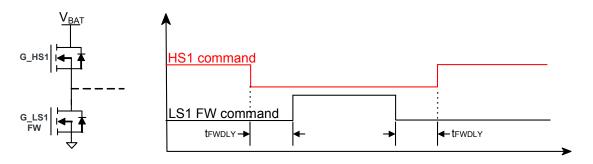


Figure 5. Automatic freewheeling example

4.1.3 Injection banks management algorithm

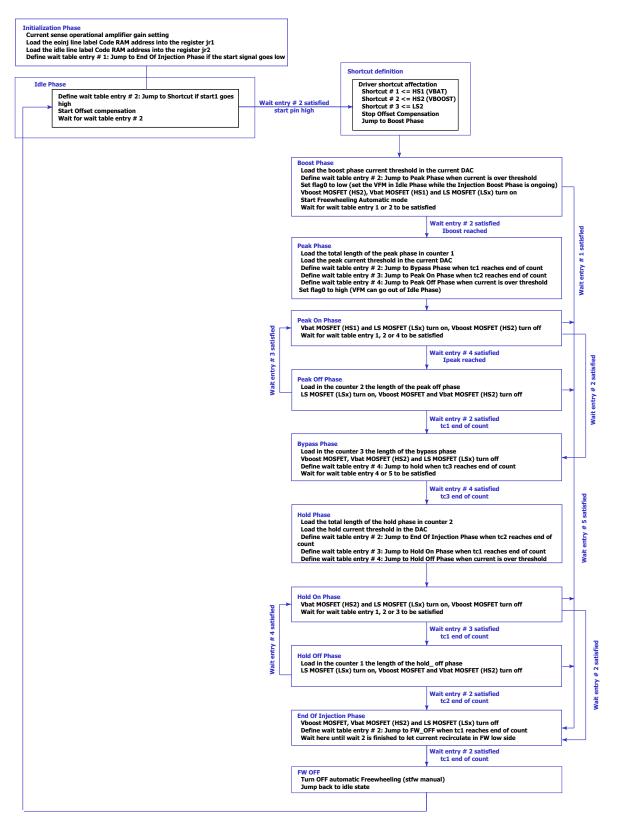


Figure 6. Algorithm example for one injector

Refer to Injection banks management source code

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4.2 DC-DC resonant mode management

In Resonant mode, off switching is triggered by the sense current rising above an upper current threshold and the on switching is triggered by the low-side VDS going below 2.5 V. This mode uses a current control loop within a voltage control loop. The voltage control loop is controlled by microcode, V_{BOOST} high and low threshold are defined in the DRAM.

To use this mode on the KITPT2000FRDM3C, the capacitance C_{RES} needs to be populated, it is not the case by default. The code dedicated to the boost converter regulation loop is loaded into the Code RAM 3. This code is executed independently by the microcore Uc0Ch3.

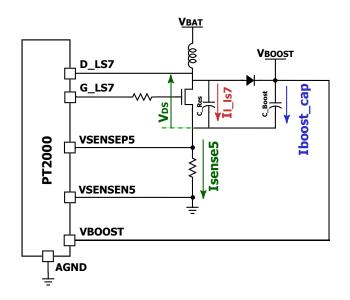


Figure 7. Simplified DC-DC converter topology for resonant mode

At boost startup, the current through the inductor oscillates between a high current threshold and the VDS low-side, this avoid switching loss during the turn OFF. Low-side VDS goes lower than 2.5 V as soon as there is no more current in the inductance, the goal of the C_{RES} is to reduce the switching slew rates to detect when VDS across low-side mosfet goes below 2.5 V with the fast comparator. In case the VDS is not detected (this can be the case if V_{BOOST} - V_{BAT} > V_{BAT}), a timeout is used to turn the low-side ON again.

When this switch is on, the current grows through the sense resistor and the low-side switch. When the switch is open, the current decays through the diode and loads the output capacitor. It increases the voltage until the V_{BOOST} voltage reaches the V_{BOOST} HIGH threshold. This phase uses the asynchronous mode and the current modulation is managed by an independent circuitry enabled by the microcore.

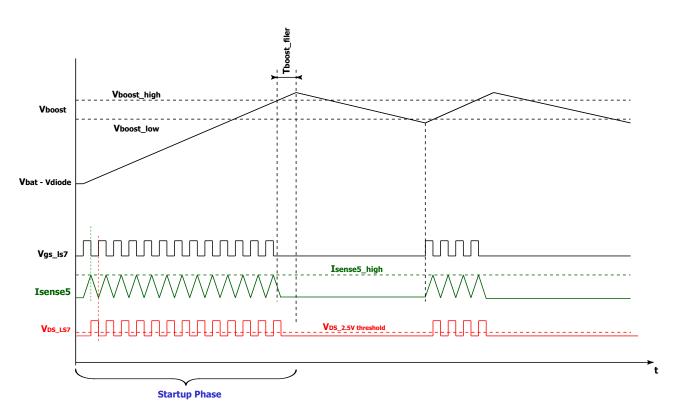


Figure 1. Resonant mode startup sequence

When the V_{BOOST_HIGH} threshold is reached, the synchronous mode is enabled. In this case, the microcore takes the direct control of the low-side switch. The low-side switch is turned off until the boost voltage goes below the V_{BOOST_LOW} threshold.

Each time the V_{BOOST_HIGH} threshold is reached, the V_{BOOST_LOW} threshold is setup, and the synchronous mode is activated after a t_{BOOST_FILTER} filter time required by the voltage comparator circuitry enablement.

Each time the boost voltage falls below the V_{BOOST_LOW} threshold the V_{BOOST_HIGH} threshold is setup, and the asynchronous mode is activated after a t_{BOOST} FILTER filter time.

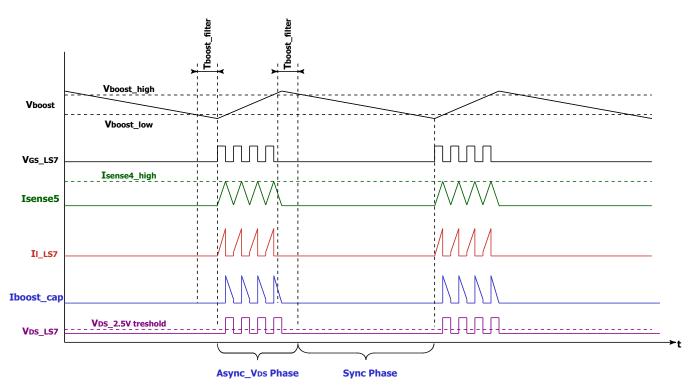


Figure 2. DCDC voltage and current diagram

Parameter name	Description	Value
V _{BOOST_HIGH}	V _{BOOST} voltage high threshold	65.31 V
V _{BOOST_LOW}	V _{BOOST} voltage low threshold	64.69 V
I _{SENSE5_LOW}	Low current threshold (only use in case VDS 2.5 V threshold not reached)	0.07 A
I _{SENSE5_HIGH}	High current threshold	3.95 A

In the present case, most of the code branches (jumps) are managed according to the V_{BOOST} voltage and the flag0 state by means of the wait table. The wait table rows are affected as shown in Table 23. and are changed according to the actuation phase.

Table 23. Example of wait table definition for the DCDC.

Phase	Async_Vds phase (dcdc_on)	Sync phase (dcdc_off)	Idle phase
Row 1	If flag0 is low, then jump to Idle phase	If flag0 is low, then jump to Idle phase	-
Row 2	-	If V _{BOOST} < V _{BOOST_LOW} , then jump to Async_Vds Phase	-
Row3	If V _{BOOST} > V _{BOOST_HIGH} , then jump to Sync phase	-	-
Row 4	-	-	-
Row 5	-	-	-

To avoid regulation disturbances, the boost voltage regulation is stopped by the means of the internal flag 0 when an injection phase starts.

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4.2.1 DC-DC management registers setup

The PT2000 registers are setup according their default states, unless defined by the following.

4.2.1.1 IO configuration registers

4.2.1.1.1 Output driver access

The Uc0Ch3 must have access to the pre-driver LS7 only.

Table 1. Out_acc_uc0_ch3 register (164h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	Acc_u c0_ch3 _ls8	Acc_u c0_ch3 _ls7	Acc_u c0_ch3 _ls6	Acc_u c0_ch3 _ls5	Acc_u c0_ch3 _ls4	Acc_u c0_ch3 _ls3	Acc_u c0_ch3 _ls2	Acc_u c0_ch3 _ls1	Reserv ed	Acc_u c0_ch3 _hs7	Acc_u c0_ch3 _hs6	Acc_u c0_ch3 _hs5		Acc_u c0_ch3 _hs3	Acc_u c0_ch3 _hs2	Acc_u c0_ch3 _hs1
Value	0	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0

Resonant mode ("async_vds) is required to set the following register (182h), to set the V_{DS} LS7 monitoring to use the high speed comparator to detect when V_{DS} is lower than 2.5 V, as fast as possible. Also important is to set the bit vds7_en and set the filter time properly to start the DCDC, even if the VDS_threshold is not reached. This a safety mechanism.

Table 2. Vds7_dcdc_config (182h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	dcdcx	mode	F	Reserve	d	ls7_vds_hig hspeed_en	Cur_dcdc 7_fbk_sel	vds7_ to_en			vo	ls7_dcd	lc_timeo	ut		
Value	0	0		000		1	0	1				00000	00011			

LS7 VDS threshold needs to be set to 2.5 V to make the DCDC resonant mode efficient enough.

Table 3. Vds_threshold_ls_Part 2 (170h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name		Vds t	hr Ls8			Vds th	nr Ls7			Vds th	nr Ls6			Vds tł	nr Ls5	
Reset		00	000		01	101 (2.5V	' thresho	ld)		00	00			00	00	

4.2.1.1.2 Current sense access

The Uc0Ch3 must have access to the current sense feedback # 5L (for the timeout detection, if the 2.5 V VDS threshold is not reached) and 5H.

Table 2. Cur_block_access_3 register (168h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	acc_u c1_ch 3_curr 6H_6N eg	acc_u c1_ch 3_curr 6L	acc_u c1_ch 3_curr 5H_5N eg	acc_u c1_ch 3_curr 5L	acc_u c1_ch 3_curr 4	acc_u c1_ch 3_curr 3	acc_uc 1_ch3_ curr2	acc_uc 1_ch3_ curr1	acc_uc 0_ch3_ curr6H _6Neg	acc_uc 0_ch3_ curr6L	acc_uc 0_ch3_ curr5H _5Neg	acc_uc 0_ch3_ curr5L	acc_uc 0_ch3_ curr4	acc_uc 0_ch3_ curr3	acc_uc 0_ch3_ curr2	acc_uc 0_ch3_ curr1
Value	0	0	0	0	0	0	0	0	0	0	1	1	0	0	0	0

4.2.1.1.3 Boost DAC access

The Uc0Ch3 must have access to the Boost DAC to set the right threshold Vboost_high and Vboost_low.

Table 4. Boost_dac_access (180h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name					Rese	erved					uc1_ch 3 acc	uc0_ch 3 acc	uc1_ch 2 acc	uc0_ch 2 acc	uc1_ch 1 acc	uc0_ch 1 acc
Value					00000	00000					0	1	0	0	0	0

4.2.1.1.4 Boost filter time

The t_{BOOST_FILTER} time is defined in the Boost_filter register (181h). This filter time and type can be adjusted to improve the V_{BOOST} voltage stability.

Table 5. Boost_filter (181h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name		Reserved		filter_ty pe						Boost_f	bk_filter					
Value		000		0						000000	000111					

4.2.1.2 Diagnosis configuration registers

The low-side driver 7 must be directly controlled by the Uc0Ch3 microcore. Consequently, the output_routing fields of its output configuration register must be set to the value 31.

Table 6. LS7_output_config (1D4h)

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name	LSx_ov r					Reserved	1					ou	tput_rout	ing		inv
Reset	0				C	0000000	0						11111			0

4.2.1.3 Channel 3 configuration registers

Same as for Channel 1-2 entry points of each microcore must be selected for Channel 3.

4.2.1.3.1 Entry point registers

Entry point registers are defining where each microcore starts in each channel. With the new PT2000 IDE, it is possible to specify a label as a starting point and those registers are set automatically. For example, "init0" is the starting point of Uc0Ch3. If this feature is not used, the following register has to be set manually.

In the code example entry point of Uc0Ch3 is 0 as the first line executed, is the first Code RAM line of the channel 1.

Table 3. Uc0_entry_point registers (10Ah) of channel 1

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name			Rese	erved						er	ntry_poin	t_addres	SS			
Reset			000	000							00000	00000				

Uc1CH3 is not used in our example, but to avoid an issue if the dual sequencer is selected, the code entry point of Uc1Ch3 is set and a dummy code of at least 3 lines is added to the example. This is unusable if the bit dual microcore (140h) is not set to '1'.

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```
Table 4. Uc1_entry_point registers (10Bh) of channel 1
```

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Name			Rese	erved						е	ntry_poin	nt_addres	S			
Reset			000	000							100	001				

4.2.2 DC-DC management algorithm

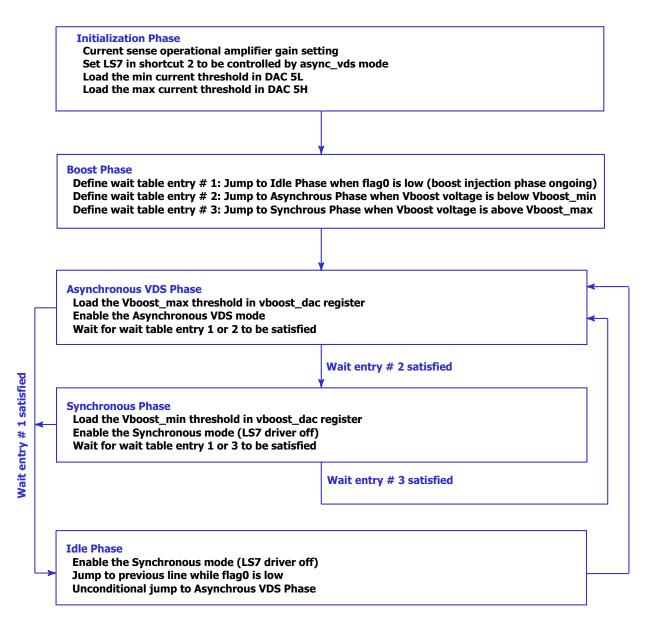


Figure 3. Algorithm example for DCDC

Refer to DC-DC Management Source Code

5 PCB layout recommendations

5.1 Ground connections

The PT2000 exposed pad must be connected to the PCB ground. All the grounds (AGND, DGND, and PGND) must be 'in star' or considering a unique ground layer, such as to minimize the introduction of offset and noise mainly in the signal return lines.

5.2 Sense resistors connection

The sense resistor's layout must be considered with special care, to sense the voltage as close as possible to the resistor terminations.

Balanced series resistance, induced by the layout, between the sense resistor positive termination to the VSENSEPx and the sense resistor negative termination to the VSENSENx pin is recommended. The balance can be achieved by implementing similar line lengths.

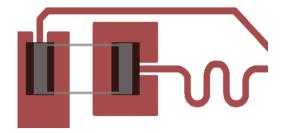


Figure 4. Example of force and sense connection layout

It is highly recommended to place the sense resistor as close as possible to its corresponding low-side MOSFET transistor.

Application source code 6

Download the full project, including microcode, plus register and DRAM settings, at this location http://www.nxp.com/files/analog/doc/app_note/AN5187SW.zip

Injection banks management source code 6.1

*	* * * * * * * * * * * * * * * * * * * *	**************************************	:*************************************	* * * * *
* File Nam	me: MC33PT2000_VFM			*
	Revision: 1.0			*
	: MC33PT2000 examp	le - 1 Bank Channel 1 controls INJ	I and INIA	*
* Descript	CION: MC33F12000	Channel I Concrois ING	I ANU INUZ	*
* REV AU	THOR DATE	DESCRIPTION OF CHANG	E	*
*			-	*
* 1.0 ble	6868 2015/03/25	- initial coding		*
*	* * * * * * * * * * * * * * * * * * * *	*****	*****	*
			***************************************	* * * *
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-		-	the application or use of a	
	-		; neither does it convey a	-
-			of others. NXP products	*
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			or other applications intend	
			n in which the failure of t sonal injury or death may	the *
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* arising	out of, directl	y or indirectly, any	claim of personal injury	or *
			nauthorized use, even if su	ıch *
			garding the design or	*
	ture of the part. rks of NXP Ltd.	NXP and the NXP log	o are registered	*
		* * * * * * * * * * * * * * * * * * * *	*****	* * * * *
	0 111			
	crocore uc0 will c	ONTTOI BANKI		
5	de Vbat = hs1 de Vboost = hs2			
-	e Freewheeling = 1	sl		
* Low side				
* current	sense = curl			
#include	"dram1.def";			
******	* * * * * * * * * * * * * * * * * * *	*****	*****	*****
*		INIT PH		
*******	* * * * * * * * * * * * * * * * * * *	******************	***************************************	* * * * * * * * * * * * * * * * * * * *
* ### Init	tialization phase	###		
init0:			the opamp of the current me	
			abel Code RAM address into t	
			el Code RAM address into the signal goes low, go to eoing	
	sto ls1 off;	IUWI; " II UNE SLAFT	arguar goes row, go to eoinj	J Puase
*******	* * * * * * * * * * * * * * * * * * * *	* * * * * * * * * * * * * * * * * * * *	****	****
*		IDLE PH		*
******	* * * * * * * * * * * * * * * * * * *			*****
* ### Idle	e phase- the uPC l	oops here until start	signal is present ###	
idle0:	<pre>stoc on sssc;</pre>		* Turn ON offset compense	ation

cwer CheckStart start row2; * Define entry table for high start pin * uPC is stuck here for almost the whole idle time WaitLoop: wait row2; WaltLoop: walt row2; CheckStart: joslr inj1_start start1; * Jump to inj1 if start 1 is high jmpr WaitLoop; SHORTCUT DEFINITION * ### Shortcuts definition per the injector to be actuated ### stoc off sssc; * Disable Offset Compensation BOOST PHASE ***** * ### Launch phase enable boost ### * Load the boost phase current threshold in the current DAC * Jump to peak phase when current is over threshold boost0: load Iboost dac_sssc _ofs; cwer peak0 cur1 row2; stf low b0; * set flag0 low to force the DC-DC converter in idle mode stos on on on; * Turn VBAT off, BOOST on, LS on * LS1 is used as the freewheeling of hs1 stfw auto; stf high b1; * Test purpose wait row12; * Wait for one of the previously defined conditions PEAK PHASE * ### Peak phase continue on Vbat ### ldcd rst _ofs keep Keep Tpeak_tot c1; load Ipeak dac_sssc _ofs; cwer bypass0 tc1 row2; cwer peak_on0 tc2 row3; cwer peak_off0 cur1 row4;
* Load the length of the total peak phase in counter 1
* Load the peak current threshold in the current DAC
* Jump to bypass phase when tc1 reaches end of count
* Jump to peak_on when tc2 reaches end of count
* Jump to peak_off when current is over threshold peak0: stf high b0; * set flag0 high to release the DC-DC converter idle mode peak_on0: stos on off on; * Turn VBAT on, BOOST off, LS on wait row124; * Wait for one of the previously defined conditions peak_off0: ldcd rst ofs keep keep Tpeak_off c2; * Load in the counter 2 the length of the peak off phase stos off off on; * Turn VBAT off, BOOST off, LS on wait row123; BYPASS PHASE * ### Bypass phase ###

 bypass0:
 ldcd rst ofs keep keep Tbypass c3;
 * Load in the counter 3 the length of the off_phase phase

 stos off off off;
 * Turn VBAT off, BOOST off, LS off

 cwer hold0 tc3 row4;
 * Jump to hold when tc3 reaches end of count

 wait row14; * Wait for one of the previously defined conditions HOLD PHASE * ### Hold phase on Vbat ### ldcd rst _ofs keep keep Thold_tot c1; * Load the length of the total hold phase in counter 2 hold0:

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hold_off0: ldcd rst _ofs keep keep Thold_off c2; * Load the length of the hold_off phase in counter 1 stos off off on; * Turn VBAT off, BOOST off, LS on wait row123; hold on0: stos on off on; * Turn VBAT on, BOOST off, LS on wait row124; * Wait for one of the previously defined conditions END OF INJECTION PHASE * ### End of injection phase ### eoinj0: stos off off off; * Turn VBAT off, BOOST off, LS off stf high b0; * set flag0 to high to release the DC-DC converter idle mode ldcd rst _ofs keep keep FWtimeOFF c1; * Load the length of the time when FW low side will stay ON cwer FW_OFF tc1 row2; * Jump to eoinj phase when tc1 reaches end of count wait row2; * Wait until FW phase is complete * Turn OFF FW low side, go back to manual mode FW_OFF: stfw manual; jmpf jr2; * Jump back to idle phase * ### End of Channel 1 - uCore0 code ### ***** * This microcore uc1 will control BANK2 * High SIde Vbat = hs3 * High SIde Vboost = hs4 * Low side Freewheeling = 1s3 * Low side = 1s4 * current sense = cur2 ***** INIT PHASE * ### Initialization phase ### stgn gain8.68 sssc; ldjr1 eoinj1; ldjr2 idle1; init1. * Set the gain of the opamp of the current measure block 1 * Load the eoinj line label Code RAM address into the register jr1 * Load the idle line label Code RAM address into the register ir2 ldjr2 idle1; cwef jr1 _start row1; * If the start signal goes low, go to eoinj phase sto ls3 off; + IDLE PHASE * ### Idle phase- the uPC loops here until start signal is present ### idle1: stoc on sssc; * Turn ON offset compensation cwer CheckStart1 start row2; * Define entry table for high start pin WaitLoop1: wait row2; * uPC is stuck here for almost the whole idle time * Jump to inj2 if start 2 is high CheckStart1: joslr inj2 start start2; jmpr WaitLoop1; SHORTCUT DEFINITION ****** * ### Shortcuts definition per the injector to be actuated ### inj2_start: dfsct hs3 hs4 ls4; * Set the 3 shortcuts: VBAT, VBOOST, LS4 * Stop offset compensation stoc off sssc;

```
BOOST PHASE
                                                                                     *
* ### Launch phase enable boost ###
boost1: load Iboost dac_sssc _ofs;
                                         * Load the boost phase current threshold in the current DAC
         cwer peak1 cur2 row2;
                                         * Jump to peak phase when current is over threshold
* set flag0 low to force the DC-DC converter in idle mode
         stf low b0;
                                        * Turn VBAT off, BOOST on, LS on
         stos on on on;
          stfw auto;
                                          * LS1 is used as the freewheeling of hs3
          stf high b2;
          wait row12;
                                         * Wait for one of the previously defined conditions
+
                                        PEAK PHASE
* ### Peak phase continue on Vbat ###
peak1: ldcd rst _ofs keep keep Tpeak_tot c1; * Load the length of the total peak phase in counter 1
         load Ipeak dac_sssc_ofs; * Load the peak current threshold in one if
the peak dac_sssc_ofs; * Load the peak current threshold in one if
the peak dac_sssc_ofs; * Jump to bypass phase when tcl reaches end of count

         cwer bypass1 tc1 row2;
cwer peak_on1 tc2 row3;
                                         * Jump to peak_on when tc2 reaches end of count
                                     * Jump to peak_on when tc2 reaches end of count
* Jump to peak_off when current is over threshold
          cwer peak_off1 cur2 row4;
          stf high b0;
                                          * set flag0 high to release the DC-DC converter idle mode
peak off1: ldcd rst ofs keep keep Tpeak off c2; * Load in the counter 2 the length of the peak off phase
          stos off off on;
                                          * Turn VBAT off, BOOST off, LS on
         wait row123;
peak on1: stos on on on;
                                          * Turn VBAT on, BOOST off, LS on
                                          * Wait for one of the previously defined conditions
         wait row124;
*****
                                        BYPASS PHASE
* ### Bypass phase ###

      bypass1:
      ldcd rst ofs keep keep Tbypass c3;
      * Load in the counter 3 the length of the off_phase phase

      stos off off off;
      * Turn VBAT off, BOOST off, LS off

         stos off off off;
          cwer hold1 tc3 row4;
                                         * Jump to hold when tc3 reaches end of count
          wait row14;
                                          * Wait for one of the previously defined conditions
HOLD PHASE
* ### Hold phase on Vbat ###

      hold1:
      ldcd rst _ofs keep keep Thold_tot c1;
      * Load the length of the total hold phase in counter 2

      load Ihold dac_sssc _ofs;
      * Load the hold current threshold in the DAC

      cwer eoinj1 tc1 row2;
      * Jump to eoinj phase when tc1 reaches end of count

         cwer eoinj1 tc1 row2;
cwer hold_on1 tc2 row3;
                                         * Jump to hold on when tc2 reaches end of count
                                          * Jump to hold_off when current is over threshold
          cwer hold_off1 cur2 row4;
hold_off1: ldcd rst _ofs keep keep Thold_off c2; * Load the length of the hold_off phase in counter 1
          stos off off on;
                                          * Turn VBAT off, BOOST off, LS on
         wait row123;
hold_on1: stos on off on;
                                          * Turn VBAT on, BOOST off, LS on
         wait row124;
                                          * Wait for one of the previously defined conditions
END OF INJECTION PHASE
```

Application source code

* ### End	of injection phase ###
eoinj1:	stos off off off;
	<pre>stf high b0;</pre>
	<pre>ldcd rst _ofs keep keep FWtimeOFF c1;</pre>
	<pre>cwer FW_OFF1 tc1 row2;</pre>
	wait row2;
FW_OFF1:	<pre>stfw manual;</pre>
	jmpf jr2;

- * Turn VBAT off, BOOST off, LS off
- * set flag0 to high to release the DC-DC converter idle mode
- \star Load the length of the time when FW low side will stay ON
- * Jump to eoinj phase when tc1 reaches end of count
- * Wait until FW phase is complete
- * Turn OFF FW low side, go back to manual mode
- * Jump back to idle phase

* ### End of Channel 1 - uCorel code ###

6.2 DC-DC Management Source Code

DCDC mode used in this case is resonant
Some changes on the hardware needs to be done to go to resonant mode or freewheeling

#include "dram3.def";

* ### Initi	alization phase ###	
init0: sl56dac dac5;		* DAC5 is used for DCDC
	<pre>stgn gain19.25 sssc; dfsct undef ls7 undef;</pre>	* Set the gain of the opamp of the current measure block 56 * NEW on the PT2000 only shortcut 2 is sensitve to async vds
instructior		" NEW ON the Fizodo only shortcut z is sensitive to async_vus
2.5V VDS no	<pre>stos keep off keep; load Iboost_L dac_sssc _ofs; ot reached</pre>	* Load Isense5L just used for the timeout detection in case
<pre>load Iboost_H dac56h56n _ofs;</pre>		* Load Isense56_high current threshold in DAC 56H
<pre>stdm null; cwer dcdc_idle _f0 row1; cwer dcdc_on _vb row2; cwer dcdc_off vb row3;</pre>		 * Set the boost voltage DAC access mode * Wait table entry for Vboost under Vboost low threshold condition * Wait table entry for Vboost under Vboost low threshold condition * Wait table entry for Vboost over Vboost high threshold condition
	chronous phase ###	
dcdc_on:	<pre>load Vboost_H dac56h56n _ofs; stdcctl async vds;</pre>	* Load the upper Vboost threshold in vboost_dac register * Enable asynchronous mode
	wait row13;	* Wait for one of the previously defined conditions
* ### Synch	nronous phase ###	
	load Vboost_L dac56h56n _ofs;	* Load the upper Vboost threshold in vboost_dac register
	<pre>stdcctl sync; wait row12;</pre>	* Enable synchronous mode * Wait for one of the previously defined conditions
	walt lowiz,	wate for one of the previously defined conditions
* ### Idle		* Dachle surchuser wede
acac_iare:	<pre>stdcctl sync; stos keep off keep;</pre>	* Enable synchronous mode
	<pre>jocr dcdc_idle _f0;</pre>	* jump to previous line while flag 0 is low
	jmpr dcdc_on;	* force the DC-DC converter on when flag 0 goes high
* ### End of Channel 3 - uCore0 code ###		

7 References

Following are URLs where you can obtain information on NXP products and application solutions:

Document Number and Description		URL	
PT2000	Data Sheet	http://www.nxp.com/files/analog/doc/data_sheet/MC33PT2000.pdf	
KITPT2000FRDM6C	User Guide	http://cache.nxp.com/files/analog/doc/user_guide/KTPT2000FRDM6CUG.pdf	
PT2000SWUG	Programming Guide	http://cache.nxp.com/files/analog/doc/user_guide/PT2000SWUG.pdf	
PT2000IDEUG	Developer Studio	http://cache.nxp.com/files/analog/doc/user_guide/PT2000-IDEUG.pdf	
Support Pages		URL	
NXP.com		http://www.nxp.com	
Product Summary Page		http://www.nxp.com/webapp/sps/site/prod_summary.jsp?code=PT2000	
Analog Home Page		http://www.nxp.com/analog	

8 Revision history

Revision	Date	Description
1.0	9/2015	Initial release
1.0	7/2016	Updated to NXP document form and style

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