

# 68HC05JB3 68HC705JB3

# SPECIFICATION (General Release)

November 5, 1998

Semiconductor Products Sector

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November 5, 1998 GENERAL RELEASE SPECIFICATION

Section	on	Page
	SECTION 1 GENERAL DESCRIPTION	
1.1 1.2 1.3 1.4 1.4.1 1.4.2 1.4.3 1.4.4 1.4.5 1.4.6 1.4.7 1.4.8	FEATURES MASK OPTIONS MCU STRUCTURE FUNCTIONAL PIN DESCRIPTION  V <sub>DD</sub> and V <sub>SS</sub> OSC1, OSC2 RESET IRQ 3.3V D+ and D- PA0-PA7 PB0-PB2, PB3-PB7 PC0-PC3	1-2 1-4 1-4 1-6 1-6 1-6 1-6 1-6
	SECTION 2 MEMORY	
2.1 2.2 2.3 2.4	I/O AND CONTROL REGISTERS RAM ROM I/O REGISTERS SUMMARY	2-2 2-2
	SECTION 3 CENTRAL PROCESSING UNIT	
3.1 3.2 3.3 3.4 3.5 3.6 3.6.1 3.6.2 3.6.3 3.6.4 3.6.5	Zero Bit (Z-Bit)	3-2 3-2 3-2 3-3 3-3 3-3 3-3
	SECTION 4 INTERRUPTS	
4.1 4.2 4.3 4.4	INTERRUPT VECTORS	4-2 4-4
MC68F REV 1		TOROLA



GENERAL RELEASE SPECIFICATION November 5, 1998

Section	on Page
4.5 4.5.1 4.5.2 4.5.3 4.5.4 4.5.5 4.5.6	HARDWARE INTERRUPTS       4-4         External Interrupt IRQ       4-4         IRQ Control/Status Register (ICSR) - \$0A       4-5         Port A External Interrupts (PA0-PA3, by mask option)       4-6         Timer1 Interrupt (TIMER1)       4-7         USB Interrupt (USB)       4-7         MFT Interrupt (MFT)       4-7         SECTION 5
5.1 5.2 5.3 5.3.1 5.3.2 5.3.3	RESETS         POWER-ON RESET       5-2         EXTERNAL RESET       5-2         INTERNAL RESETS       5-2         Power-On Reset (POR)       5-2         USB Reset       5-3         Computer Operating Properly (COP) Reset       5-3
5.3.4 5.3.5	Low Voltage Reset (LVR)
6.1 6.2 6.3	STOP MODE
7.1 7.1.1 7.1.2 7.1.3 7.1.4 7.1.5 7.2 7.2.1 7.2.2 7.2.3 7.2.4 7.3 7.3.1 7.3.2 7.3.3	PORT-A.       7-1         Port-A Data Register.       7-2         Port-A Data Direction Register       7-2         Port-A Pull-down/up Register       7-2         PA0-PA3 Interrupts.       7-2         PA0-PA7 Optical Interface       7-3         PORT-B.       7-3         Port-B Data Register.       7-3         Port-B Data Direction Register       7-3         Port-B Pull-down/up Register       7-4         PB1, PB2 Slow Transition Output       7-4         PORT-C       7-5         Port-C Data Register       7-5         Port-C Data Direction Register       7-5         Port-C Pull-down/up Register       7-5         Port-C Pull-down/up Register       7-6



November 5, 1998 GENERAL RELEASE SPECIFICATION

Section	on	Page
	SECTION 8 MULTI-FUNCTION TIMER	
8.1	OVERVIEW	8-2
8.2	COMPUTER OPERATING PROPERLY (COP) WATCHDOG	8-2
8.3	MFT REGISTERS	8-2
8.3.1	Timer Counter Register (TCNT) \$09	8-2
8.3.2	Timer Control/Status Register (TCSR) \$08	8-3
8.4	OPERATION DURING STOP MODE	8-4
8.5	COP CONSIDERATION DURING STOP MODE	8-4
	SECTION 9 16-BIT TIMER	
9.1	TIMER REGISTERS (TMRH, TMRL)	9-2
9.2	ALTERNATE COUNTER REGISTERS (ACRH, ACRL)	
9.3	INPUT CAPTURE REGISTERS	
9.4	OUTPUT COMPARE REGISTERS	
9.5	TIMER CONTROL REGISTER (TCR)	9-10
9.6	TIMER STATUS REGISTER (TSR)	9-11
9.7	TIMER OPERATION DURING WAIT MODE	
9.8	TIMER OPERATION DURING STOP MODE	9-12
	SECTION 10 UNIVERSAL SERIAL BUS MODULE	
40.4		40.4
10.1	FEATURES	
10.2	OVERVIEW	
10.2.1 10.2.2		
10.2.2	3 3 3	
10.2.4	·	
10.2.5	·	
	CLOCK REQUIREMENTS	10-10
10.4	HARDWARE DESCRIPTION	
10.4.1		
10.4.2	5 5	
10.4.3		
10.4.4		
10.5	<b>5</b>	10-18
10.5.1		
10.5.2	e i	
10.5.3		
10.5.4	<b>5</b>	
10.5.5	<b>5</b>	
10.5.6	USB Control Register 2 (UCR2)	10-24
MC68F	HC05JB3	MOTOROLA



GENERAL RELEASE SPECIFICATION November 5, 1998

Section	Pa	age
10.5.8 US 10.5.9 US	SB Status Register (USR)	26 26
10.6.1 US	BB End of Transaction Interrupt	27
	d of Packet Interrupt10-2	
	SECTION 11 OPTICAL INTERFACE	
11.1 OVER	RVIEW11	. 4
	CAL INTERFACE ENABLE REGISTER11	
	SECTION 12 INSTRUCTION SET	
12.1 ADDF	RESSING MODES12	2-1
	nerent12	
	mediate12	
	rect12	
	tended12	
	lexed, No Offset12	
	lexed, 8-Bit Offset12	
	lexed, 16-Bit Offset12	
	lative12	
	struction Types12	
	gister/Memory Instructions12	
	ad-Modify-Write Instructions12	
	mp/Branch Instructions12	
	Manipulation Instructions	
	ntrol Instructions	
12.1.15 Ins	struction Set Summary12	2-8
	SECTION 13 ELECTRICAL SPECIFICATIONS	
13.1 MAXII	MUM RATINGS13	3-1
13.2 THER	RMAL CHARACTERISTICS13	3-1
13.3 DC EI	LECTRICAL CHARACTERISTICS13	3-2
13.4 USB [	DC ELECTRICAL CHARACTERISTICS13	3-3
	LOW SPEED SOURCE ELECTRICAL CHARACTERISTICS13	
13.6 CONT	TROL TIMING13	}-5



November 5, 1998 GENERAL RELEASE SPECIFICATION

Section	on F	Page						
	SECTION 14 MECHANICAL SPECIFICATIONS							
14.1	20-PIN PDIP (CASE 738)1	4-1						
	28-PIN PDIP (CASE 710)1							
	20-PIN SOIC (CASE 751D)1							
14.4	28-PIN SOIC (CASE 751F)	4-2						
	APPENDIX A							
	MC68HC705JB3							
A.1	INTRODUCTION	A-1						
A.2	MEMORY	A-1						
A.3	MASK OPTION REGISTER (MOR)							
A.4	BOOTSTRAP MODE	A-3						
A.5	EPROM PROGRAMMING	A-3						
A.5.1	EPROM Program Control Register (PCR)							
A.5.2	Programming Sequence	A-4						
A.6	EPROM PROGRAMMING SPECIFICATIONS	A-5						





GENERAL RELEASE SPECIFICATION November 5, 1998

### **TABLE OF CONTENTS**

Section Page



November 5, 1998 GENERAL RELEASE SPECIFICATION

### **LIST OF FIGURES**

Figure	e Title	Page
1-1	MC68HC05JB3 Block Diagram	1-3
1-2	MC68HC05JB3 Pin Assignments	
1-3	Oscillator Connections	
2-1	MC68HC05JB3 Memory Map	2-1
2-2	MC68HC05JB3 I/O Registers \$0000-\$000F	2-3
2-3	MC68HC05JB3 I/O Registers \$0010-\$001F	
2-4	MC68HC05JB3 I/O Registers \$0020-\$002F	
2-5	MC68HC05JB3 I/O Registers \$0030-\$003F	2-6
2-6	COP Register (COPR)	2-6
3-1	MC68HC05 Programming Model	3-1
4-1	Interrupt Stacking Order	
4-2	Interrupt Flowchart	
4-3	External Interrupt (IRQ) Logic	4-5
4-4	IRQ Control and Status Register (ICSR)	4-5
5-1	Reset Sources	
5-2	COP Watchdog Register (COPR)	5-3
6-1	STOP and WAIT Flowchart	6-2
7-1	PB1 Slow Falling-edge Output	7-5
8-1	Multi-Function Timer Block Diagram	8-1
8-2	Timer Counter Register	
8-3	Timer Control/Status Register (TCSR)	
9-1	Programmable Timer Block Diagram	
9-2	Programmable Timer Counter Block Diagram	
9-3	Programmable Timer Counter Registers (TMRH, TMRL)	9-3
9-4	Alternate Counter Block Diagram	
9-5	Alternate Counter Registers (ACRH, ACRL)	
9-6	Timer Input Capture Block Diagram	
9-7	TCAP Input Signal Conditioning	
9-8	TCAP Input Comparator Output	
9-9	Input Capture Registers (ICRH, ICRL)	
9-10	Timer Output Compare Block Diagram	9-9
9-11	Output Compare Registers (OCRH, OCRL)	
9-12	Timer Control Register (TCR)	
9-13	Timer Status Registers (TSR)	
10-1	USB Block Diagram	
10-2	Supported Transaction Types per Endpoint	
10-3	Supported USB Packet Types	
10-4	Sync Pattern	
10-5	SOP, Sync Signaling and Voltage Levels	
10-6	CRC Block Diagram for Address and Endpoint Fields	
10-7	CRC Block Diagram for Data Packets	
10-8	EOP Transaction Voltage Levels	
10-9	EOP Width Timing	10-8

MC68HC05JB3 REV 1 MOTOROLA



GENERAL RELEASE SPECIFICATION November 5, 1998

### **LIST OF FIGURES**

Figure	e litle	Page
	External Low Speed Device Configuration	
	Regulator Electrical Connections	
	Low Speed Driver Signal Waveforms	
	Differential Input Sensitivity Over Entire Common Mode Range	
	Data Jitter	
	Data Signal Rise and Fall Time	
	NRZI Data Encoding	
	Flow Diagram for NRZI	
	Bit Stuffing	
	Flow Diagram for Bit Stuffing	
	USB Address Register (UADDR)	
	USB Interrupt Register 0 (UIR0)	
	USB Interrupt Register 1(UIR1)	
	USB Control Register 0 (UCR0)	
	USB Control Register 1 (UCR1)	
	USB Control Register 2 (UCR2)	
10-26	USB Status Register (USR)	10-25
	USB Endpoint 0 Data Register (UE0D0-UE0D7)	
	USB Endpoint 1/Endpoint2 Data Registers (UE1D0-UE1D7)	
	OUT Token Data Flow for Receive Endpoint 0	
	SETUP Token Data Flow for Receive Endpoint 0	
	IN Token Data Flow for Transmit Endpoint 0	
	IN Token Data Flow for Transmit Endpoint 1/2	
11-1	A pair of Optical Coupler Interface	
11-2	Optical Interface Comparator	
11-3	Optical Interface Enable Register (TCSR)	11-3
14-1	20-Pin PDIP Mechanical Dimensions	
14-2	28-Pin PDIP Mechanical Dimensions	
14-3	20-Pin SOIC Mechanical Dimensions	
14-4	28-Pin SOIC Mechanical Dimensions	
A-1	MC68HC705JB3 Memory Map	
A-2	EPROM Programming Sequence	A-5



November 5, 1998 GENERAL RELEASE SPECIFICATION

### **LIST OF TABLES**

Table	Title	Page
4-1	Reset/Interrupt Vector Addresses	4-1
7-1	Summary of Port Pin Functions	
8-1	RTI and COP Rates at f <sub>OP</sub> =3.0MHz	8-3
10-1	Supported Packet Identifiers	10-5
10-2	Register Summary	
11-1	Port-A Optical Interface Pairs	
11-2	Optical Interface Reference Voltage Selection	11-3
12-1	Register/Memory Instructions	12-4
12-2	Read-Modify-Write Instructions	12-5
12-3	Jump and Branch Instructions	12-6
12-4	Bit Manipulation Instructions	12-7
12-5	Control Instructions	12-7
12-6	Instruction Set Summary	12-8
12-7	Opcode Map	12-14
13-1	DC Electrical Characteristics	
13-2	USB DC Electrical Characteristics	13-3
13-3	USB Low Speed Source Electrical Characteristics	13-4
13-4	Control Timing	13-5
A-1	EPROM Programming Electrical Characteristics	A-5





GENERAL RELEASE SPECIFICATION November 5, 1998

**LIST OF TABLES** 

Table Title Page



November 5, 1998 GENERAL RELEASE SPECIFICATION

# Freescale Semiconductor, Inc.

# SECTION 1 GENERAL DESCRIPTION

The MC68HC05JB3 is a member of the low-cost, high-performance M68HC05 Family of 8-bit microcontroller units (MCUs). The M68HC05 Family is based on the customer-specified integrated circuit design strategy. All MCUs in the family use the popular M68HC05 central processing unit (CPU) and are available with a variety of subsystems, memory sizes and types, and package types.

The MC68HC05JB3 is specifically designed to be used in applications where a low speed (1.5Mbps) Universal Serial Bus (USB) interface is required.

### 1.1 FEATURES

- Industry standard M68HC05 CPU core
- Memory-mapped input/output (I/O) registers
- 2560 Bytes of user ROM
- 144 Bytes of user RAM (includes 64 byte stack)
- Fully compliant Low Speed USB with 3 Endpoints:
  - 1 Control Endpoint (2×8-byte buffer)
  - 2 Interrupt Endpoints (1×8-byte buffer shared)
- 3.3V dc output for USB pull-up resistors
- 19 Bidirectional I/O pins with the following features:
  - 17 I/Os have software programmable pull-down capability
  - 2 open-drain I/Os have software programmable pull-up, 25mA current sink capability
  - 4 I/Os with external interrupt capability
  - 8 I/Os (in 4 pairs) with programmable optical interface
- Multi-Function Timer (MFT)
- 16-bit Timer with 1 input capture and 1 output compare
- Low Voltage Reset (LVR)
- Computer Operating Properly (COP) Watchdog Reset
- Illegal Address Reset



GENERAL RELEASE SPECIFICATION November 5, 1998

- Power-Saving STOP and WAIT Modes
- Available in 20-pin PDIP, 20-pin SOIC, 28-pin PDIP, and 28-pin SOIC packages

### 1.2 MASK OPTIONS

The following mask options are available:

- External interrupt pins (IRQ, PA0 to PA3):
   [edge-triggered or edge-and-level-triggered]
- Port A, port B, and port C pull-down/pull-up resistors: [connected or disconnected]
- PA0-PA3 external interrupt capability: [enabled or disabled]
- OSC, crystal/ceramic resonator startup delay: [4064 or 224 internal bus cycles]
- Low Voltage Reset (LVR): [enabled or disabled]
- COP function of MFT: [enabled or disabled]

### 1.3 MCU STRUCTURE

Figure 1-1 shows the structure of MC68HC05JB3 MCU.

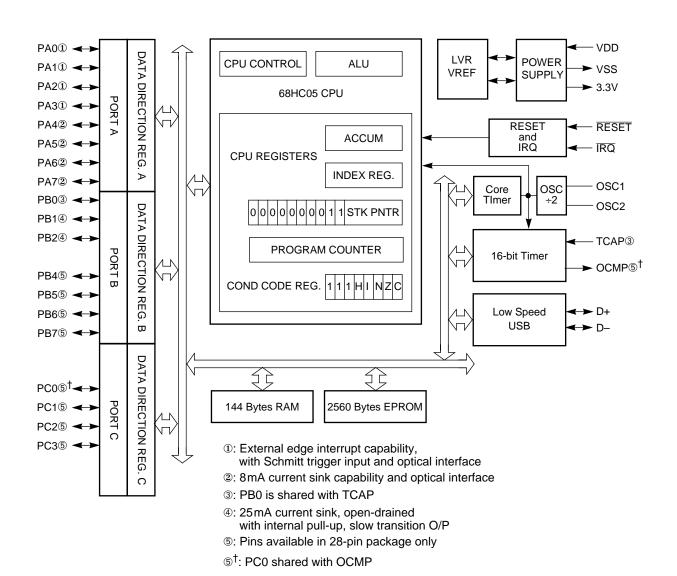


Figure 1-1. MC68HC05JB3 Block Diagram

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GENERAL RELEASE SPECIFICATION November 5, 1998

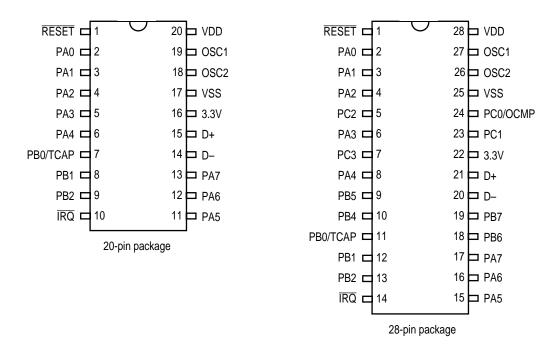


Figure 1-2. MC68HC05JB3 Pin Assignments

### 1.4 FUNCTIONAL PIN DESCRIPTION

The following paragraphs give a description of the general function of each pin assigned in **Figure 1-2**.

### 1.4.1 $V_{DD}$ and $V_{SS}$

Power is supplied to the MCU through  $V_{DD}$  and  $V_{SS}$ .  $V_{DD}$  is the positive supply, and  $V_{SS}$  is ground. The MCU operates from a single power supply.

Very fast signal transitions occur on the MCU pins. The short rise and fall times place very high short-duration current demands on the power supply. To prevent noise problems, special care should be taken to provide good power supply bypassing at the MCU by using bypass capacitors with good high-frequency characteristics that are positioned as close to the MCU as possible. Bypassing requirements vary, depending on how heavily the MCU pins are loaded.

### 1.4.2 OSC1, OSC2

The OSC1 and OSC2 pins are the connections for the on-chip oscillator. The OSC1 and OSC2 pins can accept the following sets of components:

- 1. A crystal as shown in **Figure 1-3**(a)
- 2. A ceramic resonator as shown in Figure 1-3(a)
- 3. An external clock signal as shown in **Figure 1-3**(b)

MOTOROLA 1-4 November 5, 1998 GENERAL RELEASE SPECIFICATION

The frequency,  $f_{OSC}$ , of the oscillator or external clock source is divided by two to produce the internal operating frequency,  $f_{OP}$  If the internal operating frequency is 3MHz, then the external oscillator frequency will be 6MHz. For LS USB 1.5MHz frequency clock can be derived from a divided by 4 circuit. The type of oscillator is selected by a mask option. An internal 2M $\Omega$  resistor may be selected between OSC1 and OSC2 by a mask option (crystal/ceramic resonator mode only).

### **Crystal Oscillator**

The circuit in **Figure 1-3**(a) shows a typical oscillator circuit for an AT-cut, parallel resonant crystal. The crystal manufacturer's recommendations should be followed, as the crystal parameters determine the external component values required to provide maximum stability and reliable start-up. The load capacitance values used in the oscillator circuit design should include all stray capacitances. The crystal and components should be mounted as close as possible to the pins for start-up stabilization and to minimize output distortion. An internal start-up resistor of approximately  $2M\Omega$  is provided between OSC1 and OSC2 for the crystal type oscillator as a mask option.

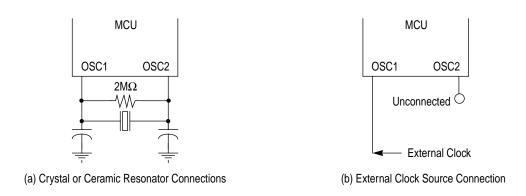


Figure 1-3. Oscillator Connections

### **Ceramic Resonator Oscillator**

In cost-sensitive applications, a ceramic resonator can be used in place of the crystal. The circuit in **Figure 1-3**(a) can be used for a ceramic resonator. The resonator manufacturer's recommendations should be followed, as the resonator parameters determine the external component values required for maximum stability and reliable starting. The load capacitance values used in the oscillator circuit design should include all stray capacitances. The ceramic resonator and components should be mounted as close as possible to the pins for start-up stabilization and to minimize output distortion. An internal start-up resistor of approximately 2 M $\Omega$  is provided between OSC1 and OSC2 for the ceramic resonator type oscillator as a mask option.



GENERAL RELEASE SPECIFICATION November 5, 1998

### **External Clock**

An external clock from another CMOS-compatible device can be connected to the OSC1 input, with the OSC2 input not connected, as shown in **Figure 1-3**(b). This configuration is possible ONLY when the crystal/ceramic resonator mask option is selected.

### 1.4.3 **RESET**

This is an I/O pin. This pin can be used as an input to reset the MCU to a known start-up state by pulling it to the low state. The  $\overline{RESET}$  pin contains a steering diode to discharge any voltage on the pin to  $V_{DD}$ , when the power is removed. An internal pull-up is also connected between this pin and  $V_{DD}$ . The  $\overline{RESET}$  pin contains an internal Schmitt trigger to improve its noise immunity as an input. This pin is an output pin if LVR triggers an internal reset.

### 1.4.4 **IRQ**

This input pin drives the asynchronous IRQ interrupt function of the CPU. The IRQ interrupt function has a mask option to provide either only negative edge-sensitive triggering or both negative edge-sensitive and low level-sensitive triggering. If the option is selected to include level-sensitive triggering, the  $\overline{\text{IRQ}}$  input requires an external resistor to  $V_{DD}$  for "wired-OR" operation, if desired. The  $\overline{\text{IRQ}}$  pin contains an internal Schmitt trigger as part of its input to improve noise immunity.

### NOTE

Each of the PA0 to PA3 I/O pins may be connected as an OR function with the IRQ interrupt function by a mask option. This capability allows keyboard scan applications where the transitions or levels on the I/O pins will behave the same as the  $\overline{\text{IRQ}}$  pin. The edge or level sensitivity selected by a separate mask option for the  $\overline{\text{IRQ}}$  pin also applies to the I/O pins OR'ed to create the IRQ signal.

### 1.4.5 3.3V

This is an output reference voltage nominally set at 3.3V dc.

### 1.4.6 D+ and D-

These two lines carry the USB differential data. For low speed device such as MC68HC05JB3, a 1.5 k $\Omega$  resistor is required to be connected across D– and 3.3V for proper signal termination.

### 1.4.7 PA0-PA7

These eight I/O lines comprise Port A. PA0 to PA7 are push-pull pins with pull-down devices. The state of any pin is software programmable and all Port A lines are configured as inputs during power-on or reset.



November 5, 1998 GENERAL RELEASE SPECIFICATION

PA0 to PA3 has external interrupt function (mask option) with schmitt trigger input circuit, and PA4 to PA7 has 8mA current sink capability.

Port A can also be configured as the optical interface.

### 1.4.8 PB0-PB2, PB3-PB7

These seven I/O lines comprise Port B. The state of any pin is software programmable and is configured as an input during power-on or reset.

PB1 and PB2 are open-drain I/O lines with pull-up devices. PB0 (shared with TCAP) is a push-pull I/O line with pull-down device.

PB1 and PB2 are also slow transition outputs, each has 25mA current sink capability at  $V_{OL}$  =0.5V.

PB4-PB7 I/O lines are push-pull pins with pull-down devices, and are only available in the 28-pin package.

### 1.4.9 PC0-PC3

These four I/O lines comprise Port C. The state of any pin is software programmable and all Port C lines are configured as inputs during power-on or reset.

PC0 to PC3 are push-pull pins with pull-down devices. PC0 is also shared with the OCMP pin from the output compare function of the 16-bit timer.

Port C is only available in the 28-pin package.





GENERAL RELEASE SPECIFICATION November 5, 1998



The MC68HC05JB3 has 8k-bytes of addressable memory, with 64 bytes of I/O, 144 bytes of user RAM, and 2560 bytes of user ROM, as shown in **Figure 2-1**.

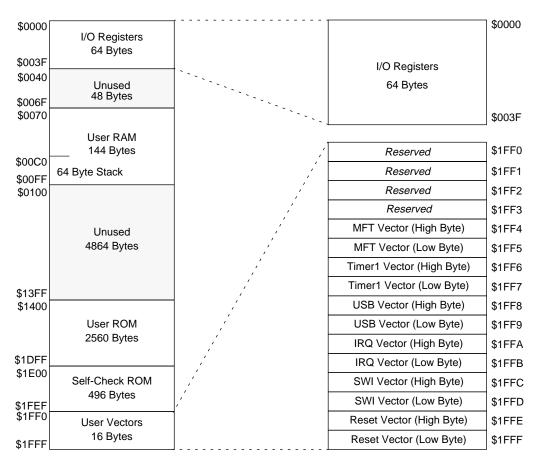


Figure 2-1. MC68HC05JB3 Memory Map



GENERAL RELEASE SPECIFICATION November 5, 1998

### 2.1 I/O AND CONTROL REGISTERS

The I/O and Control Registers reside in locations \$0000 to \$003F. The bit assignments for each register are shown in **Figure 2-2**, **Figure 2-3**, **Figure 2-4**, and **Figure 2-5**. Reading from unused bits will return unknown states, and writing to unused bits will be ignored.

### 2.2 RAM

The user RAM consists of 144 bytes (including the stack) at locations \$0080 to \$012F. The stack begins at address \$00FF and proceeds down to \$00C0. Using the stack area for data storage or temporary work locations requires care to prevent it from being overwritten due to stacking from an interrupt or subroutine call.

### 2.3 ROM

There are a total of 3k-bytes of ROM on chip. This includes 2560 bytes of user ROM with locations \$1400 to \$1DFF for user program storage and 16 bytes for user vectors at locations \$1FF0 to \$1FFF. Also, 496 bytes of Self-check ROM on chip at locations \$1E00 to \$1FEF.



November 5, 1998 GENERAL RELEASE SPECIFICATION

### 2.4 I/O REGISTERS SUMMARY

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
\$0000	Port A Data	R	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0
\$0000	PORTA	W	FA/	FAU	PAS	FA4	PAS	r <sub>AZ</sub>	PAI	PAU
\$0001	Port B Data	R	PB7	PB6	PB5	PB4		PB2	PB1	PB0
Ψ0001	PORTB	W	T D7	1 00	1 00	104		1 02	TDI	1 00
\$0002	Port C Data	R					PC3	PC2	PC1	PC0
Ψ0002	PORTC	W					1 03	1 02	101	1 00
\$0003	Unused	R								
\$0005	Onasea	W								
\$0004	Port A Data Direction	R	DDRA7	DDRA6	DDRA5	DDRA4	DDRA3	DDRA2	DDRA1	DDRA0
φοσοι	DDRA	W	5510.0	<i>DD10</i> 70	<i>DB1010</i>	BBIOTI	<i>DD11</i> 710	DDIVIL	BBIGIT	<i>DD1010</i>
\$0005	Port B Data Direction	R	DDRB7	DDRB6	DDRB5	DDRB4	SLOWE	DDRB2	DDRB1	DDRB0
φοσσσ	DDRB	W	BBINBI	DDNDO	DBRBO	BBRBT	OLOWE	DDND2	DDIND!	DDINDO
\$0006	Port C Data Direction	R	ОСМРО	VROFF			DDRC3	DDRC2	DDRC1	DDRC0
Ψοσσσ	DDRC	W	OOWII O				DDINOG		DDIXO!	DBROO
\$0007	Unused	R								
φοσο.		W								
\$0008	MFT Ctrl/Status	R	TOF	RTIF	TOFE	RTIE	0	0	RT1	RT0
Ψσσσσ	TCSR	W					TOFR	RTIFR		
\$0009	MFT Counter	R	TMR7	TMR6	TMR5	TMR4	TMR3	TMR2	TMR1	TMR0
Ψ0000	TCNT	W								
\$000A	IRQ Control/Status	R	IRQE	0	0	0	IRQF	0	0	IRQPU
	ICSR	W							IRQR	
\$000B	Unused	R								
7.000	0.11.000	W								
\$000C	Unused	R								
<b>40000</b>	3774654	W								
\$000D	Unused	R								
		W								
\$000E	Optical Interface En.	R	TCMPE	VREF2	VREF1	VREF0	OIE3	OIE2	OIE1	OIE0
	OIER	W								
\$000F	Port C Pull-down/up	R								
ļ - J - J - J	PDURC	W					PDRC3	PDRC2	PDRC1	PDRC0

Figure 2-2. MC68HC05JB3 I/O Registers \$0000-\$000F



GENERAL RELEASE SPECIFICATION November 5, 1998

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
<b>CO040</b>	Port A Pull-down/up	R								
\$0010	PDURA	W	PDRA7	PDRA6	PDRA5	PDRA4	PDRA3	PDRA2	PDRA1	PDRA0
\$0011	Port B Pull-down/up	R								
\$0011	PDURB	W	PDRB7	PDRB6	PDRB5	PDRB4		PURB2	PURB1	PDRB0
\$0012	Timer1 Control	R	ICIE	OCIE	TOIE	0	0	0	IEDG	0
\$0012	TCR	W	ICIE	OCIE	IOIE				IEDG	
\$0013	Timer1 Status	R	ICF	OCF	TOF	0	0	0	0	0
\$0013	TSR	W								
\$0014	Input Capture MSB	R	ICH7	ICH6	ICH5	ICH4	ICH3	ICH2	ICH1	ICH0
φυσ14	ICH	W								
\$0015	Input Capture LSB	R	ICL7	ICL6	ICL5	ICL4	ICL3	ICL2	ICL1	ICL0
\$0015	ICL	W								
\$0016	Output Compare MSB	R	OCH7	OCH6	OCH5	OCH4	OCH3	OCH2	OCH1	OCH0
\$0010	OCH	W	OCIT	OCITO	OCHS	00114	OCHS	OCHZ	OCITI	OCITO
\$0017	Output Compare LSB	R	OCL7	OCL6	OCL5	OCL4	OCL3	OCL2	OCL1	OCL0
φυσ17	OCL	W	OCL	OCLO	OCLS	OCL4	OCLS	OCLZ	OCLI	OCLU
\$0018	Timer1 Counter MSB	R	TCNTH7	TCNTH6	TCNTH5	TCNTH4	TCNTH3	TCNTH2	TCNTH1	TCNTH0
ψ0010	TCNTH	W								
\$0019	Timer1 Counter LSB	R	TCNTL7	TCNTL6	TCNTL5	TCNTL4	TCNTL3	TCNTL2	TCNTL1	TCNTL0
ψ0019	TCNTL	W								
\$001A	Alter. Counter MSB	R	ACNTH7	ACNTH6	ACNTH5	ACNTH4	ACNTH3	ACNTH2	ACNTH1	ACNTH0
\$001A	ACNTH	W								
\$001B	Alter. Counter LSB	R	ACNTL7	ACNTL6	ACNTL5	ACNTL4	ACNTL3	ACNTL2	ACNTL1	ACNTL0
Ψ001Β	ACNTL	W								
\$001C	Unused	R								
\$0010	Onusea	W								
\$001D	Unused	R								
\$0010	Unuseu	W								
\$001E	Unused	R								
\$UUTE	Unuseu	W								
\$001F	Unused	R								
φυσιτ	Ulluseu	W								

Figure 2-3. MC68HC05JB3 I/O Registers \$0010-\$001F



November 5, 1998 GENERAL RELEASE SPECIFICATION

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
\$0020	USB Endpoint 0 Data 0	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0020	UD0R0	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
\$0021	USB Endpoint 0 Data 1	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0021	UD0R1	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
\$0022	USB Endpoint 0 Data 2	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0022	UD0R2	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
\$0023	USB Endpoint 0 Data 3	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0023	UD0R3	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
\$0024	USB Endpoint 0 Data 4	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
φυυ24	UD0R4	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
\$0025	USB Endpoint 0 Data 5	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0025	UD0R5	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
\$0026	USB Endpoint 0 Data 6	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0026	UD0R6	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
<b>#0007</b>	USB Endpoint 0 Data 7	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0027	UD0R7	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
Ф0000	USB Endpoint 1 Data 0	R								
\$0028	UD1R0	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
\$0029	USB Endpoint 1 Data 1	R								
\$0029	UD1R1	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
Ф000A	USB Endpoint 1 Data 2	R								
\$002A	UD1R2	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
\$002B	USB Endpoint 1 Data 3	R								
\$002B	UD1R3	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
<b>#0000</b>	USB Endpoint 1 Data 4	R								
\$002C	UD1R4	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
#000D	USB Endpoint 1 Data 5	R								
\$002D	UD1R5	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
Ф000Б	USB Endpoint 1 Data 6	R								
\$002E	UD1R6	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
<b>#000</b>	USB Endpoint 1 Data 7	R								
\$002F	UD1R7	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0

Figure 2-4. MC68HC05JB3 I/O Registers \$0020-\$002F



GENERAL RELEASE SPECIFICATION November 5, 1998

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
¢0000	I Investoral	R								
\$0030	Unused	W								
00001	I lava and	R								
\$0031	Unused	W								
¢0000	l love and	R								
\$0032	Unused	W								
\$0022	Ununad	R								
\$0033	Unused	W								
\$0034	Ununnd	R								
\$0034	Unused	W								
¢0025	Unuond	R								
\$0035	Unused	W								
\$0036	I lava and	R								
\$0030	Unused	W								
\$0037	USB Control 2	R		0	TX1ST	0	ENIADI EO	ENABLE1	STALL2	STALL1
φυυσ <i>ι</i>	UCR2	W		TX1STR			ENABLEZ	ENABLE	STALLZ	STALLT
\$0038	USB Address	R	USBEN	UADD6	UADD5	UADD4 UADD3	UADD2	LIADD1	UADD0	
\$0036	UADR	W	USBEIN	UADD6	UADDS	UADD4	UADD3	UADD2	UADD1	UADDU
\$0039	USB Interrupt 0	R	TXD0F	RXD0F	RSTF	SUSPND	TXD0IE	RXD0IE	0	0
φυυσ9	UIR0	W	0	0	0	SUSFIND	INDUIL	KADUIL	TXD0FR	RXD0FR
\$003A	USB Interrupt 1	R	TXD1F	EOPF	RESUMF	0	TXD1IE	EOPIE	0	0
\$003A	UIR1	W	0	0	0	RESUMFR	IVDIIE	EOFIE	TXD1FR	EOPFR
\$003B	USB Control 0	R	T0SEQ	STALL0	TX0E	RX0E	TDOCIZO	TDOCIZO	TP0SIZ1	TP0SIZ0
\$003B	UCR0	W	10324	STALLU	IAUE	KAUE	TFUSIZS	1703122	1703121	1703120
\$003C	USB Control 1	R	T19E0	ENDADD	TX1E	EDECLIM	TD10172	TD19172	TD10171	TD19170
\$0030	UCR1	W	IISEQ	LINDADD	IXIL	FRESUM	17 13123	TP1SIZ2	1713121	17 13120
\$003D	USB Status	R	RSEQ	SETUP			RPSIZ3	RPSIZ2	RPSIZ1	RPSIZ0
\$0030	USR	W								
\$002F	Doggrand	R								
\$003E	Reserved	W								
\$0025	Pagaryad	R								
\$003F	Reserved	W								

Figure 2-5. MC68HC05JB3 I/O Registers \$0030-\$003F

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
\$1FF0	COP Register	R	0	0	0	0	0	0	0	0
	COPR	W								COPR

Figure 2-6. COP Register (COPR)

# SECTION 3 CENTRAL PROCESSING UNIT

The MC68HC05JB3 has an 8k-bytes memory map. The stack has only 64 bytes. Therefore, the stack pointer has been reduced to only 6 bits and will only decrement down to \$00C0 and then wrap-around to \$00FF. All other instructions and registers behave as described in this chapter.

### 3.1 REGISTERS

The MCU contains five registers which are hard-wired within the CPU and are not part of the memory map. These five registers are shown in **Figure 3-1** and are described in the following paragraphs.

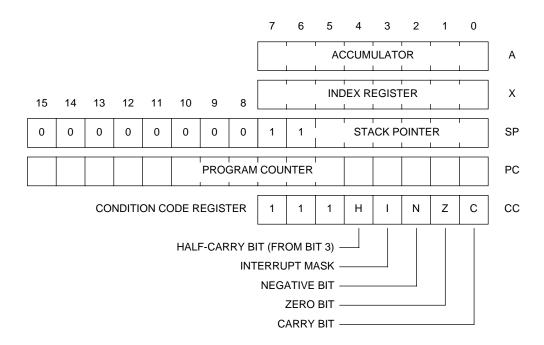


Figure 3-1. MC68HC05 Programming Model



GENERAL RELEASE SPECIFICATION November 5, 1998

### 3.2 ACCUMULATOR (A)

The accumulator is a general purpose 8-bit register as shown in **Figure 3-1**. The CPU uses the accumulator to hold operands and results of arithmetic calculations or non-arithmetic operations. The accumulator is not affected by a reset of the device.

### 3.3 INDEX REGISTER (X)

The index register shown in **Figure 3-1** is an 8-bit register that can perform two functions:

- Indexed addressing
- Temporary storage

In indexed addressing with no offset, the index register contains the low byte of the operand address, and the high byte is assumed to be \$00. In indexed addressing with an 8-bit offset, the CPU finds the operand address by adding the index register content to an 8-bit immediate value. In indexed addressing with a 16-bit offset, the CPU finds the operand address by adding the index register content to a 16-bit immediate value.

The index register can also serve as an auxiliary accumulator for temporary storage. The index register is not affected by a reset of the device.

### 3.4 STACK POINTER (SP)

The stack pointer shown in **Figure 3-1** is a 16-bit register. In MCU devices with memory space less than 64k-bytes the unimplemented upper address lines are ignored. The stack pointer contains the address of the next free location on the stack. During a reset or the reset stack pointer (RSP) instruction, the stack pointer is set to \$00FF. The stack pointer is then decremented as data is pushed onto the stack and incremented as data is pulled off the stack.

When accessing memory, the ten most significant bits are permanently set to 000000011. The six least significant register bits are appended to these ten fixed bits to produce an address within the range of \$00FF to \$00C0. Subroutines and interrupts may use up to 64(\$C0) locations. If 64 locations are exceeded, the stack pointer wraps around and overwrites the previously stored information. A subroutine call occupies two locations on the stack and an interrupt uses five locations.

### 3.5 PROGRAM COUNTER (PC)

The program counter shown in **Figure 3-1** is a 16-bit register. In MCU devices with memory space less than 64k-bytes the unimplemented upper address lines are ignored. The program counter contains the address of the next instruction or operand to be fetched.



November 5, 1998 GENERAL RELEASE SPECIFICATION

Normally, the address in the program counter increments to the next sequential memory location every time an instruction or operand is fetched. Jump, branch, and interrupt operations load the program counter with an address other than that of the next sequential location.

### 3.6 CONDITION CODE REGISTER (CCR)

The CCR shown in **Figure 3-1** is a 5-bit register in which four bits are used to indicate the results of the instruction just executed. The fifth bit is the interrupt mask. These bits can be individually tested by a program, and specific actions can be taken as a result of their states. The condition code register should be thought of as having three additional upper bits that are always ones. Only the interrupt mask is affected by a reset of the device. The following paragraphs explain the functions of the lower five bits of the condition code register.

### 3.6.1 Half Carry Bit (H-Bit)

When the half-carry bit is set, it means that a carry occurred between bits 3 and 4 of the accumulator during the last ADD or ADC (add with carry) operation. The half-carry bit is required for binary-coded decimal (BCD) arithmetic operations.

### 3.6.2 Interrupt Mask (I-Bit)

When the interrupt mask is set, the interrupt and external interrupts are disabled. Interrupts are enabled when the interrupt mask is cleared. When an interrupt occurs, the interrupt mask is automatically set after the CPU registers are saved on the stack, but before the interrupt vector is fetched. If an interrupt request occurs while the interrupt mask is set, the interrupt request is latched. Normally, the interrupt is processed as soon as the interrupt mask is cleared.

A return from interrupt (RTI) instruction pulls the CPU registers from the stack, restoring the interrupt mask to its state before the interrupt was encountered. After any reset, the interrupt mask is set and can only be cleared by the Clear I-Bit (CLI), or WAIT instructions.

### 3.6.3 Negative Bit (N-Bit)

The negative bit is set when the result of the last arithmetic operation, logical operation, or data manipulation was negative. (Bit 7 of the result was a logical one.)

The negative bit can also be used to check an often tested flag by assigning the flag to bit 7 of a register or memory location. Loading the accumulator with the contents of that register or location then sets or clears the negative bit according to the state of the flag.

### 3.6.4 Zero Bit (Z-Bit)

The zero bit is set when the result of the last arithmetic operation, logical operation, data manipulation, or data load operation was zero.

MC68HC05JB3 REV 1 **CENTRAL PROCESSING UNIT** 

**MOTOROLA** 



GENERAL RELEASE SPECIFICATION November 5, 1998

### 3.6.5 Carry/Borrow Bit (C-Bit)

The carry/borrow bit is set when a carry out of bit 7 of the accumulator occurred during the last arithmetic operation, logical operation, or data manipulation. The carry/borrow bit is also set or cleared during bit test and branch instructions and during shifts and rotates. This bit is neither set by an INC nor by a DEC instruction.



### **SECTION 4 INTERRUPTS**

The MCU can be interrupted in six different ways:

- Non-maskable Software Interrupt Instruction (SWI)
- External Asynchronous Interrupt (IRQ)
- External Interrupt via IRQ on PA0-PA3 (mask option)
- **USB** Interrupt
- Timer1 Interrupt (16-bit Timer)
- **Multi-Function Timer Interrupt**

### 4.1 **INTERRUPT VECTORS**

Table 4-1. Reset/Interrupt Vector Addresses

Function	Source	Control Bit	Global Hardware Mask	Local Software Mask	Priority (1 = Highest)	Vector Address
Reset	Power-On Logic RESET Pin Low Voltage Reset Illegal Address Reset	_	_	_	1	\$1FFE-\$1FFF
	COP Watchdog					
Software Interrupt (SWI)	User Code	_	_	_	Same Priority As Instruction	\$1FFC-\$1FFD
External Interrupt (IRQ)	ĪRQ Pin	_	I Bit	IRQE Bit	2	\$1FFA-\$1FFB
USB Interrupts	TXD0F TXD1F RESUMP	_	l Bit	TXD0IE TXD1IE —	3	\$1FF8-\$1FF9
Timer1 Interrupts	ICF Bit OCF Bit TOF Bit	_	l Bit	ICIE Bit OCIE Bit TOIE Bit	4	\$1FF6-\$1FF7
MFT Interrupts	CTOF Bit RTIF Bit	_	I Bit	CTOFE Bit RTIE Bit	5	\$1FF4-\$1FF5
	\$1FF2-\$1FF3					
	\$1FF0-\$1FF1					



GENERAL RELEASE SPECIFICATION November 5, 1998

### NOTE

If more than one interrupt request is pending, the CPU fetches the vector of the higher priority interrupt first. A higher priority interrupt does not actually interrupt a lower priority interrupt service routine unless the lower priority interrupt service routine clears the I bit.

### 4.2 INTERRUPT PROCESSING

The CPU does the following actions to begin servicing an interrupt:

- Stores the CPU registers on the stack in the order shown in **Figure 4-1**.
- Sets the I bit in the condition code register to prevent further interrupts.
- Loads the program counter with the contents of the appropriate interrupt vector locations as shown in **Table 4-1**.

The return from interrupt (RTI) instruction causes the CPU to recover its register contents from the stack as shown in **Figure 4-1**. The sequence of events caused by an interrupt are shown in the flow chart in **Figure 4-2**.

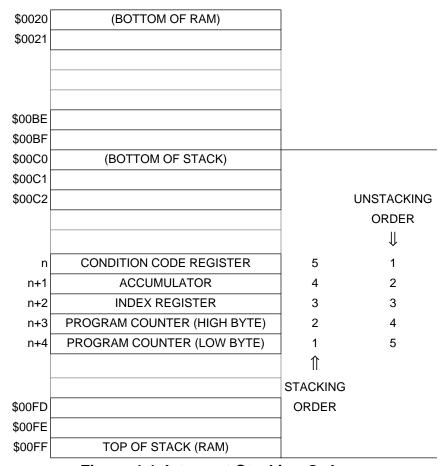


Figure 4-1. Interrupt Stacking Order



November 5, 1998 GENERAL RELEASE SPECIFICATION

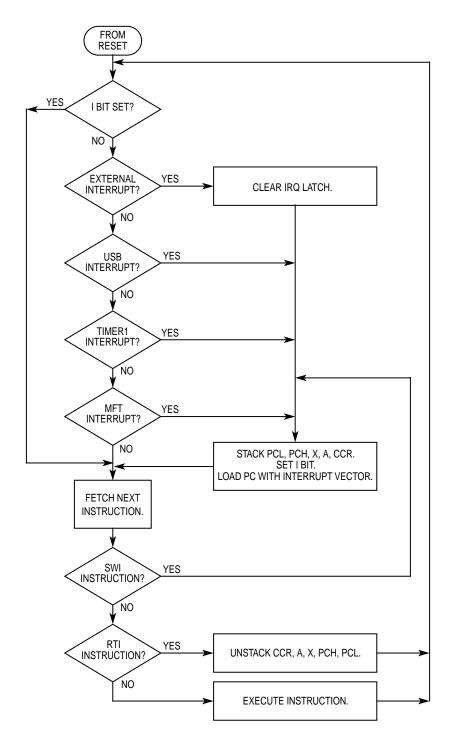


Figure 4-2. Interrupt Flowchart



GENERAL RELEASE SPECIFICATION November 5, 1998

### 4.3 RESET INTERRUPT SEQUENCE

The RESET function is not in the strictest sense an interrupt; however, it is acted upon in a similar manner as shown in **Figure 4-2**. A low level input on the RESET pin or an internally generated RST signal causes the program to vector to its starting address which is specified by the contents of memory locations \$1FFE and \$1FFF. The I-bit in the condition code register is also set.

### 4.4 SOFTWARE INTERRUPT (SWI)

The SWI is an executable instruction and a non-maskable interrupt since it is executed regardless of the state of the I-bit in the CCR. As with any instruction, interrupts pending during the previous instruction will be serviced before the SWI opcode is fetched. The interrupt service routine address is specified by the contents of memory locations \$1FFC and \$1FFD.

### 4.5 HARDWARE INTERRUPTS

All hardware interrupts except RESET are maskable by the I-bit in the CCR. If the I-bit is set, all hardware interrupts (internal and external) are disabled. Clearing the I-bit enables the hardware interrupts. There are two types of hardware interrupts which are explained in the following sections.

### 4.5.1 External Interrupt IRQ

The IRQ pin provides an asynchronous interrupt to the CPU. A block diagram of the IRQ logic is shown in **Figure 4-3**.

The IRQ pin is one source of an IRQ interrupt and a mask option can also enable the four lower Port-A pins (PA0 to PA3) to act as other IRQ interrupt sources.

Refer to **Figure 4-3** for the following descriptions. IRQ interrupt source comes from IRQ latch. The IRQ latch will be set on the falling edge of the  $\overline{\text{IRQ}}$  pin or on any falling edge of PA0-3 pins if PA0-3 interrupts have been enabled. If 'edge-only' sensitivity is chosen by a mask option, only the IRQ latch output can activate an IRQF flag which creates a request to the CPU to generate the IRQ interrupt sequence. This makes the IRQ interrupt sensitive to the following cases:

- 1. Falling edge on the IRQ pin.
- 2. Falling edge on any PA0-PA3 pin with IRQ enabled (via mask option).

If level sensitivity is chosen, the active high state of the signal to the clock input of the IRQ latch can also activate an IRQF flag which creates an IRQ request to the CPU to generate the IRQ interrupt sequence. This makes the IRQ interrupt sensitive to the following cases:

- 1. Low level on the IRQ pin.
- 2. Falling edge on the IRQ pin.
- 3. Low level on any PA0-PA3 pin with IRQ enabled (via mask option).
- 4. Falling edge on any PA0-PA3 pin with IRQ enabled (via mask option).



November 5, 1998 GENERAL RELEASE SPECIFICATION

The IRQE enable bit controls whether an active IRQF flag can generate an IRQ interrupt sequence. This interrupt is serviced by the interrupt service routine located at the address specified by the contents of \$1FFA and \$1FFB.

If IRQF is set, the only way to clear this flag is by writing a logic one to the IRQR acknowledge bit in the ICSR. As long as the output state of the IRQF flag bit is active the CPU will continuously re-enter the IRQ interrupt sequence until the active state is removed or the IRQE enable bit is cleared.

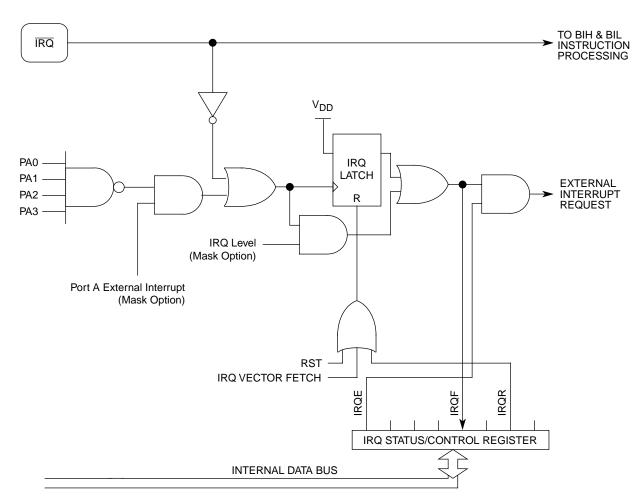


Figure 4-3. External Interrupt (IRQ) Logic

### 4.5.2 IRQ Control/Status Register (ICSR) - \$0A

The IRQ interrupt function is controlled by the ICSR located at \$000A. All unused bits in the ICSR will read as logic zeros. The IRQF bit is cleared and IRQE bit is set by reset.

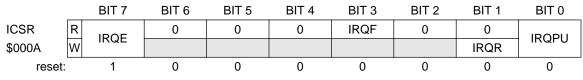


Figure 4-4. IRQ Control and Status Register (ICSR)

MC68HC05JB3 INTERRUPTS MOTOROLA REV 1 4-5



GENERAL RELEASE SPECIFICATION November 5, 1998

### IRQPU — IRQ pin PUII-up resistor enable

This bit enables/disables the internal pull-up resistor on the  $\overline{IRQ}$  pin.

- $1 = Internal pull-up resistor in <math>\overline{IRQ}$  pin enabled.
- 0 = Internal pull-up resistor in IRQ pin disabled.

### IRQR — IRQ Interrupt Acknowledge

This write-only bit clears an IRQ interrupt by clearing the IRQ latch, and hence the IRQF bit. The IRQR bit will always read as a logic zero.

- 1 = Clears IRQ interrupt request (clears IRQF).
- 0 = No effect.

### IRQF — IRQ Interrupt Request Flag

Writing to the IRQF flag bit will have no effect on it. If the additional setting of IRQF flag bit is not cleared in the IRQ service routine and the IRQE enable bit remains set the CPU will re-enter the IRQ interrupt sequence continuously until either the IRQF flag bit or the IRQE enable bit is clear. The IRQF latch is cleared by reset.

- 1 = Indicates that an IRQ request is pending.
- 0 = Indicates that no IRQ request triggered by pins PA0-3 or IRQ is pending. The IRQF flag bit can be cleared by writing a logic one to the IRQR acknowledge bit to clear the IRQ latch and also conditioning the external IRQ sources to be inactive (if the level sensitive interrupts are enabled via mask option). Doing so before exiting the service routine will mask out additional occurrences of the IRQF.

### IRQE — IRQ Interrupt Enable

The IRQE bit enables/disables the IRQF flag bit to initiate an IRQ interrupt sequence.

- 1 = Enables IRQ interrupt, that is, the IRQF flag bit can generate an interrupt sequence. Reset sets the IRQE enable bit, thereby enabling IRQ interrupts once the I-bit is cleared. Execution of the STOP or WAIT instructions causes the IRQE bit to be set in order to allow the external IRQ to exit these modes.
- 0 = The IRQF flag bit cannot generate an interrupt sequence.

### 4.5.3 Port A External Interrupts (PA0-PA3, by mask option)

The IRQ interrupt can also be triggered by the inputs on the PA0 to PA3 port pins if enabled by a single mask option. If enabled, the lower four bits of Port A can activate the IRQ interrupt function, and the interrupt operation will be the same as for inputs to the  $\overline{\text{IRQ}}$  pin. This mask option of PA0-3 interrupt allow all of these input pins to be OR'ed with the input present on the  $\overline{\text{IRQ}}$  pin. All PA0 to PA3 pins must be selected as a group as an additional IRQ interrupt. All the PA0-3 interrupt sources are also controlled by the IRQE enable bit.





November 5, 1998 GENERAL RELEASE SPECIFICATION

#### NOTE

The BIH and BIL instructions will only apply to the level on the  $\overline{IRQ}$  pin itself, and not to the output of the logic OR function with the PA0 to PA3 pins. The state of the individual Port A pins can be checked by reading the appropriate Port A pins as inputs.

### **NOTE**

If enabled, the PA0 to PA3 pins will cause an IRQ interrupt only when the corresponding pin is configured as input.

### 4.5.4 Timer1 Interrupt (TIMER1)

The TIMER1 interrupt is generated by the 16-bit timer when either an overflow or an input capture or output compare has occurred as described in the section on 16-bit timer. The interrupt flags and enable bits for the Timer1 interrupts are located in the Timer1 Control & Status Register (TSR) located at \$0012, \$0013. The I-bit in the CCR must be clear in order for the TIMER1 interrupt to be enabled. Either of these three interrupts will vector to the same interrupt service routine located at the address specified by the contents of memory locations \$1FF6 and \$1FF7.

# 4.5.5 USB Interrupt (USB)

The USB interrupt is generated by the USB module as described in the section on Universal Serial Bus. The interrupt enable bits for the USB interrupt are located at bit3-bit2 of UIR0 register and bit3-bit2 of UIR1 register. Also Once the device goes into Suspend Mode, any bus activities will cause the USB to generate an interrupt to CPU to come out from the Suspend mode. The I-bit in the CCR must be clear in order for the USB interrupt to be enabled. Either of these two interrupts will vector to the same interrupt service routine located at the address specified by the contents of memory locations \$1FF8 and \$1FF9.

### 4.5.6 MFT Interrupt (MFT)

The MFT interrupt is generated by the MFT module as described in the section on Multi-function Timer. These interrupts will vector to the same interrupt service routine located at the address specified by the contents of memory locations \$1FF4 and \$1FF5.



GENERAL RELEASE SPECIFICATION November 5, 1998

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# **SECTION 5** RESETS

This section describes the six reset sources and how they initialize the MCU. A reset immediately stops the operation of the instruction being executed, initializes certain control bits, and loads the program counter with a user defined reset vector address. The following conditions produce a reset:

- Initial power up of device (power on reset).
- A logic zero applied to the RESET pin (external reset).
- Timeout of the COP watchdog (COP reset).
- Low voltage applied to the device (LVR reset).
- Fetch of an opcode from an address not in the memory map (illegal address reset).
- Detection of USB reset signal (USB reset).

Figure 5-1 shows a block diagram of the reset sources and their interaction.

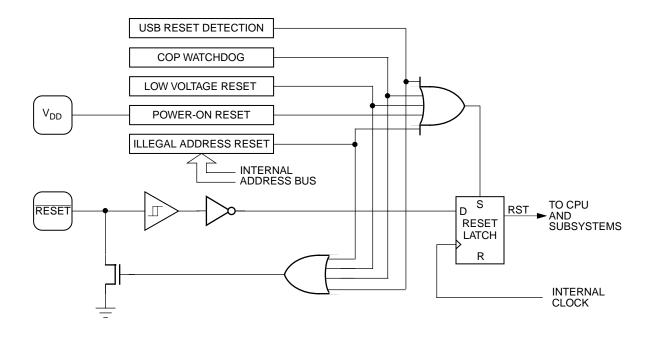


Figure 5-1. Reset Sources



GENERAL RELEASE SPECIFICATION November 5, 1998

### 5.1 POWER-ON RESET

A positive transition on the  $V_{DD}$  pin generates a power-on reset. The power-on reset is strictly for conditions during powering up and cannot be used to detect drops in power supply voltage.

A  $224t_{CYC}$  or  $4064t_{CYC}$  (internal clock cycle) delay after the oscillator becomes active allows the clock generator to stabilize. If the  $\overline{RESET}$  pin is at logic zero at the end of the multiple  $t_{CYC}$  time, the MCU remains in the reset condition until the signal on the  $\overline{RESET}$  pin goes to a logic one.

### 5.2 EXTERNAL RESET

A logic zero applied to the  $\overline{\text{RESET}}$  pin for 1.5t<sub>CYC</sub> generates an external reset. This pin is connected to a Schmitt trigger input gate to provide and upper and lower threshold voltage separated by a minimum amount of hysteresis. The external reset occurs whenever the  $\overline{\text{RESET}}$  pin is pulled below the lower threshold and remains in reset until the  $\overline{\text{RESET}}$  pin rises above the upper threshold. This active low input will generate the internal RST signal that resets the CPU and peripherals.

The  $\overline{\text{RESET}}$  pin can also act as an open drain output. It will be pulled to a low state by an internal pulldown device that is activated by three internal reset sources. This RESET pulldown device will only be asserted for 3 to 4 cycles of the internal clock,  $f_{OP}$  or as long as the internal reset source is asserted. When the external  $\overline{\text{RESET}}$  pin is asserted, the pulldown device will not be turned on.

#### NOTE

Do not connect the RESET pin directly to V<sub>DD</sub>, as this may overload some power supply designs when the internal pulldown on the RESET pin activates.

### 5.3 INTERNAL RESETS

The five internally generated resets are the initial power-on reset function, the COP Watchdog timer reset, the low voltage reset, and the illegal address detector. Only the COP Watchdog timer reset, low voltage reset and illegal address detector will also assert the pulldown device on the RESET pin for the duration of the reset function or 3 to 4 internal clock cycles, whichever is longer.

## 5.3.1 Power-On Reset (POR)

The internal POR is generated on power-up to allow the clock oscillator to stabilize. The POR is strictly for power turn-on conditions and is not able to detect a drop in the power supply voltage (brown-out). There is an oscillator stabilization delay of 224 or 4064 (224 or 4064 is selected by mask option) internal processor bus clock cycles after the oscillator becomes active.



November 5, 1998 GENERAL RELEASE SPECIFICATION

The POR will generate the RST signal which will reset the CPU. If any other reset function is active at the end of the 224 or 4064 cycle delay, the RST signal will remain in the reset condition until the other reset condition(s) end.

POR will not activate the pulldown device on the  $\overline{RESET}$  pin.  $V_{DD}$  must drop below  $V_{POR}$  in order for the internal POR circuit to detect the next rise of  $V_{DD}$ .

#### 5.3.2 USB Reset

The USB reset is generated by a detection on the USB bus reset signal. For MC68HC05JB3, seeing a single-end zero on its upstream port for 4 to 8 bit times will set RSTF bit in UIR0 register. The detections will also generate the RST signal to reset the CPU and other peripherals in the MCU.

### 5.3.3 Computer Operating Properly (COP) Reset

The COP watchdog is enabled by a mask option.

A timeout of the COP watchdog generates a COP reset. The COP watchdog is part of a software error detection system and must be cleared periodically to start a new timeout period. To clear the COP watchdog and prevent a COP reset, write a logic zero to the COPC bit of the COP register at location \$1FF0.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	
COPR	R	0	0	0	0	0	0	0	0	
\$1FF0	W								COPC	
reset	:	U	U	U	U	U	U	U	0	
U = UNAFFECTED BY RESET										

Figure 5-2. COP Watchdog Register (COPR)

#### COPC — COP Clear

COPC is a write-only bit. Periodically writing a logic zero to COPC prevents the COP watchdog from resetting the MCU. Reset clears the COPC bit.

- 1 = No effect on system.
- 0 = Reset COP watchdog timer.

The COP Watchdog reset will assert the pull-down device to pull the RESET pin low for one cycle of the internal bus clock.

Refer to section on Multi-Function Timer for detail on COP watchdog timeout periods.

### 5.3.4 Low Voltage Reset (LVR)

The LVR activates the RST reset signal to reset the device when the voltage on the  $V_{DD}$  pin falls below the LVR trip voltage. The LVR will assert the pulldown device to pull the  $\overline{RESET}$  pin low one cycle of the internal bus clock. The Low Voltage Reset circuit is enabled by a mask option.



GENERAL RELEASE SPECIFICATION November 5, 1998

# 5.3.5 Illegal Address Reset

An opcode fetch from an address that is not in the ROM or the RAM generates an illegal address reset. The illegal address reset will assert the pull-down device to pull the RESET pin low for 3 to 4 cycles of the internal bus clock.



November 5, 1998 GENERAL RELEASE SPECIFICATION

# SECTION 6 LOW POWER MODES

There are three modes of operation that reduce power consumption:

- Stop mode
- Wait mode
- Data retention mode

Figure 6-1 shows the sequence of events in Stop and Wait modes.



GENERAL RELEASE SPECIFICATION November 5, 1998

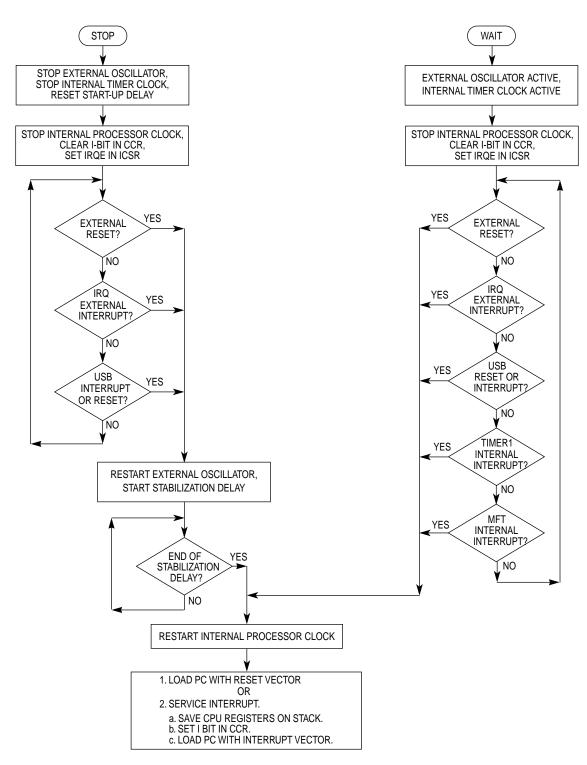


Figure 6-1. STOP and WAIT Flowchart



November 5, 1998 GENERAL RELEASE SPECIFICATION

### 6.1 STOP MODE

STOP mode is entered by executing the STOP instruction. This is the lowest power consumption mode of the MCU. In the STOP Mode the internal oscillator is turned off, halting all internal processing.

Execution of the STOP instruction automatically clears the I-bit in the Condition Code Register and sets the IRQE enable bit in the IRQ Control/Status Register so that the IRQ external interrupt is enabled. All other registers, including the other bits in the TCSR, and memory remain unaltered. All input/output lines remain unchanged.

The MCU can be brought out of the STOP Mode by an IRQ external interrupt or a USB coming out from Suspend Mode Interrupt (Bus activity detection) or an externally generated RESET, USB Reset or an LVR reset. When exiting the STOP Mode the internal oscillator will resume after a 224 or 4064 internal processor clock cycle oscillator stabilization delay.

### 6.2 WAIT MODE

WAIT mode is entered by executing the WAIT instruction. This places the MCU in a low-power mode, which consumes more power than the STOP Mode. In the WAIT Mode the internal processor clock is halted, suspending all processor and internal bus activity. Execution of the WAIT instruction automatically clears the I-bit in the Condition Code Register and sets the IRQE enable bit in the IRQ Control/ Status Register so that the IRQ external interrupt is enabled. All other registers, memory, and input/output lines remain in their previous states.

The WAIT Mode may be exited when an external IRQ, USB, Timer1 or MFT interrupt, an LVR reset, USB reset or an external RESET occurs.

### 6.3 DATA-RETENTION MODE

The Data-Retention mode is only available if the Low Voltage Reset function (mask option) is not enabled.

In the data retention mode, the MCU retains RAM contents and CPU register contents at  $V_{DD}$  voltages as low as 2Vdc. The data retention feature allows the MCU to remain in a low power consumption state during which it retains data, but the CPU cannot execute instructions. The  $\overline{RESET}$  pin must be held low during data-retention mode.





GENERAL RELEASE SPECIFICATION November 5, 1998

# SECTION 7 INPUT/OUTPUT PORTS

In normal operating mode there are 19 usable bidirectional I/O lines arranged as one 8-bit I/O port (Port-A), one 7-bit I/O port (Port-B), and one 4-bit I/O port (Port C). The individual bits in these ports are programmable as either inputs or outputs under software control by the data direction registers (DDRs).

The eight port pins, PB4-PB7 and PC0-PC3, are only available on the 28-pin version of the device.

Table 7-1 shows a summary of Port-A, Port-B, and Port-C functions.

**Table 7-1. Summary of Port Pin Functions** 

Port Pins	Internal Resistor Configuration	Current Drive/Sink	Additional Features		
PA0-PA3		1.6mA sink	External Interrupt <sup>2</sup>	Optical Interface	
PA4-PA7	Pull-down <sup>1</sup>	8mA sink	Optical Inter		
PB0		1.6mA sink	shared with TCAP		
PB1, PB2	Pull-up <sup>1</sup>	25mA sink, open-drain	Slow Transition Output		
PB4-PB7					
PC0	Pull-down <sup>1</sup>	1.6mA sink	shared with OCMP	Pins available only in 28-pin device	
PC1-PC3				1	
			·	·	

#### Notes:

- 1. A pull-up/pull-down resistor is enabled by setting the corresponding register bit to "0" and the port pull-up/down mask option is selected.
- 2. Selected by mask option.

### 7.1 PORT-A

Port-A is an 8-bit bi-directional port. The Port-A data register is at \$0000 and the data direction register (DDRA) is at \$0004. Reset does not affect the data registers, but clears the data direction registers, thereby returning the port pins to inputs. Writing a '1' to a DDR bit sets the corresponding port bit to output mode.

All Port-A pins have programmable pull-down resistors. PA4 to PA7 each has 8mA current sink capability.

The table below summarizes the pin configurations for Port-A.



GENERAL RELEASE SPECIFICATION November 5, 1998

PDRAx	DDRAx	Pin Configuration
0	0	Input with pull-down
0	1	Output Push/Pull
1	0	Input
1	1	Output Push/Pull

### 7.1.1 Port-A Data Register

	В	SIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
PORTA \$0000	R W	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0
reset:		0	0	0	0	0	0	0	0

### 7.1.2 Port-A Data Direction Register

	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
DDRA \$0004	R W DDRA7	DDRA6	DDRA5	DDRA4	DDRA3	DDRA2	DDRA1	DDRA0
reset	. 0	0	0	0	0	0	0	0

### DDRAx — PAx Data Direction

- 1 = Port pin set as output.
- 0 = Port pin set as input.

# 7.1.3 Port-A Pull-down/up Register

With the pull-up/down mask option selected, each pin in Port-A has an internal pull-down resistor which can be enabled by writing a '0' to the corresponding bit in the Port-A pull-down/up control register (PDURA) at location \$0010.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
PDURA	R								
\$0010	W	PDRA7							
rese	t:	0	0	0	0	0	0	0	0

### PDRAx — PAx Pin Pull-down enable

- 1 = Internal pull-down disabled.
- 0 = Internal pull-down enabled.

### 7.1.4 PA0-PA3 Interrupts

A mask option selects the capability for PA0-PA3 to be used as external IRQ interrupt inputs. These four I/O pins also have schmitt trigger input circuits.

See INTERRUPTS section for detail.

MOTOROLA 7-2 Semiconductor, Inc

### 7.1.5 PA0-PA7 Optical Interface

Port-A is programmable to use as ports for the optical interface.

See OPTICAL INTERFACE section for details.

#### 7.2 **PORT-B**

Port-B is a 7-bit bi-directional port. The Port-B data register is at \$0001 and the data direction register (DDRB) is at \$0005. Reset does not affect the data registers, but clears the data direction registers, thereby returning the port pins to inputs. Writing a 'one' to a DDR bit sets the corresponding port bit to output mode.

PB4-PB7 are only available on the 28-pin version of the device.

All Port-B pins have programmable pull-down or pull-up resistors. PB1 and PB2 each has 25mA current sink capability.

PB0 is also used as the 16-timer TCAP input pin. When configured as output, the input to the input capture will be permanently tied "low" and no input capture can be generated.

The table below summarizes the pin configurations for Port-B.

PDRBx/PURBx	DDRBx	Pin Configuration				
0	0	PB0, PB4-PB7: Input with pull-down PB1, PB2: Input with pull-up				
0	1	PB0, PB4-PB7: Output Push-Pull PB1, PB2: Output Open-drain with pull-up				
1	0	Input				
1	1	PB0, PB4-PB7: Output Push-Pull PB1, PB2: Output Open-drain				

### 7.2.1 Port-B Data Register

	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
PORTB \$0001	R W PB7	PB6	PB5	PB4	0	PB2	PB1	PB0
reset:	0	0	0	0	0	0	0	0

### 7.2.2 Port-B Data Direction Register

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
DDRB	R	DDRB7	DDRB6	DDRB5	DDRB4	SLOWE	DDRB2	DDRB1	DDRB0
\$0005	W	וסאטט	DUKBO	משאטט	DDKD4	SLOWE	DUKBZ	ופאטט	סטאטט
rese	et:	0	0	0	0	0	0	0	0

#### DDRBx — PBx Data Direction

1 = Port pin set as output.

0 = Port pin set as input.

MC68HC05JB3 REV 1



GENERAL RELEASE SPECIFICATION November 5, 1998

### **SLOWE** — Slow Transition Enable

See Section 7.2.4 for details.

- 1 = Enable slow falling-edge output transition feature on PB1 and PB2.
- 0 = Disable slow falling-edge output transition feature on PB1 and PB2.

## 7.2.3 Port-B Pull-down/up Register

With the pull-up/down mask option selected, PB0 and PB4-PB7 each has an internal pull-down resistor, while PB1 and PB2 each has an internal pull-up resistor, which can be enabled by writing a '0' to the corresponding bit in the Port-B pull-down/up control register (PDURB) at location \$0011.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
PDURB	R								
\$0011	W	PDRB7	PDRB6	PDRB5	PDRB4		PURB2	PURB1	PDRB0
reset:		0	0	0	0	0	0	0	0

#### PDRBx — PBx Pin Pull-down enable

- 1 = Internal pull-down disabled.
- 0 = Internal pull-down enabled.

### PURBx — PBx Pin Pull-up enable

- 1 = Internal pull-up disabled.
- 0 = Internal pull-up enabled.

### 7.2.4 PB1, PB2 Slow Transition Output

The slow transition output feature is enabled by setting the SLOWE bit in DDRB at \$0005.

PB2 — a high-to-low output transition is a sharp falling edge transition delayed by  $t_{CVC} \div 2$ .

PB1 — a high-to-low output transition is a slow falling edge (drops from 5.0V to 2.2V in 167ns typically at  $f_{OP}$ =3MHz, with 50pF load) followed by a fast transition to  $V_{SS}$ . The fast transition duration is depending on the strength of the output driver defined for each port. See **Figure 7-1**.

Both PB1 and PB2 have 25mA current sink capability.

7-4

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November 5, 1998 GENERAL RELEASE SPECIFICATION

Figure 7-1. PB1 Slow Falling-edge Output

### 7.3 PORT-C

Port-C is a 4-bit bi-directional port. The Port-C data register is at \$0002 and the data direction register (DDRC) is at \$0006. Reset does not affect the data registers, but clears the data direction registers, thereby returning the port pins to inputs. Writing a 'one' to a DDR bit sets the corresponding port bit to output mode.

All Port-C pins have programmable pull-down resistors, and are only available on the 28-pin version of the device.

PC0 is also used as the 16-timer OCMP output pin.

The table below summarizes the pin configurations for Port-A.

PDRCx	DDRCx	Pin Configuration
0	0	Input with pull-down
0	1	Output Push-Pull
1	0	Input
1	1	Output Push-Pull

# 7.3.1 Port-C Data Register

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
PORTC	R					PC3	PC2	PC1	PC0
\$0002	W					1 00	. 02		
rese	t:	0	0	0	0	0	0	0	0

# 7.3.2 Port-C Data Direction Register

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
DDRC \$0006	R W	ОСМРО	VROFF			DDRC3	DDRC2	DDRC1	DDRC0
reset	:	0	0	0	0	0	0	0	0

MC68HC05JB3 REV 1 **INPUT/OUTPUT PORTS** 

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GENERAL RELEASE SPECIFICATION November 5, 1998

### DDRCx — PCx Data Direction

- 1 = Port pin set as output.
- 0 = Port pin set as input.

### VROFF — USB 3.3V Voltage Reference

See USB section for details.

- 1 = Disable 3.3V regulator.
- 0 = Enables 3.3V regulator.

### OCMPO — OCMP Output Enable

See 16-BIT TIMER section for details.

- 1 = PC0 is OCMP pin, OCF from 16-bit timer output compare.
- 0 = PC0 is standard I/O pin, from Port-C data register.

## 7.3.3 Port-C Pull-down/up Register

With the pull-up/down mask option selected, each pin in Port-C has an internal pull-down resistor which can be enabled by writing a '0' to the corresponding bit in the Port-C pull-down/up control register (PDURC) at location \$000F.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	
PDURC	R									
\$000F	W					PDRC3	PDRC2	PDRC1	PDRC0	ı
reset		0	0	0	0	0	0	0	0	

### PDRCx — PCx Pin Pull-down Enable

- 1 = Internal pull-down disabled.
- 0 = Internal pull-down enabled.

# SECTION 8 MULTI-FUNCTION TIMER

The Multi-Function Timer (or Core Timer) module is a 15-stage ripple counter with Timer Over Flow (CTOF), Real Time Interrupt (RTI), and COP Watchdog function.

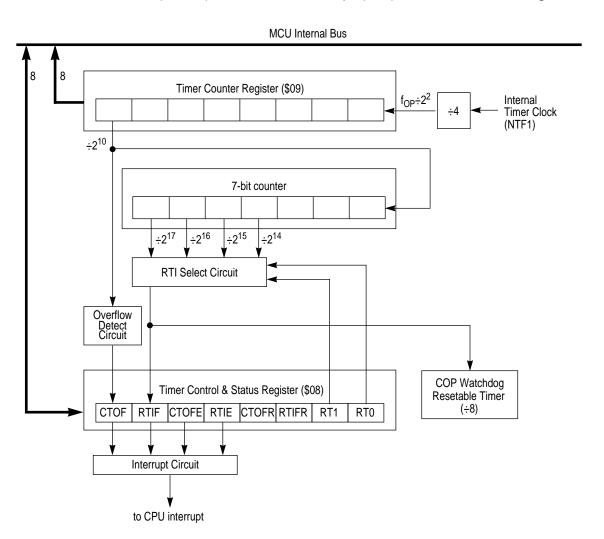


Figure 8-1. Multi-Function Timer Block Diagram



GENERAL RELEASE SPECIFICATION November 5, 1998

### 8.1 OVERVIEW

As shown in **Figure 8-1**, the Timer is driven by the timer clock, NTF1, divided by four. NTF1 has the same phase and frequency as the processor bus clock, PH2, but continues to run in WAIT mode. The NTF1 drives an 8-bit ripple counter. The value of this 8-bit ripple counter can be read by the CPU at any time by accessing the Timer Counter Register (TCNT) at address \$09. A timer overflow function is implemented on the last stage of this 8-bit counter, giving a possible interrupt rate of  $f_{OP} \div 1024$ .

The last stage of the 8-bit counter also drives a further 7-bit counter. The final four stages is used by the RTI circuit, giving possible RTI rates of  $f_{OP} \div 2^{14}$ ,  $2^{15}$ ,  $2^{16}$  or  $2^{17}$ , selected by RT1 and RT0 (see **Table 8-1**). The RTI rate selector bits, and the RTI and CTOF enable bits and flags are located in the Timer Control and Status Register at location \$08.

The power-on cycle clears the entire counter chain and begins clocking the counter. After 224 or 4064 cycles, the power-on reset circuit is released which again clears the counter chain and allows the device to come out of reset. At this point, if RESET is not asserted, the timer will start counting up from zero and normal device operation will begin. If RESET is asserted at any time during operation the counter chain will be cleared.

## 8.2 COMPUTER OPERATING PROPERLY (COP) WATCHDOG

The COP Watchdog is enabled by a mask option.

The COP Watchdog Timer function is implemented by using the output of the RTI circuit and further dividing it by eight. The minimum COP reset rates are listed in **Table 8-1**. If the COP circuit times out, an internal reset is generated and the normal reset vector is fetched.

Preventing a COP time-out is done by writing a "0" to bit-0 of address \$1FF0. When the COP is cleared, only the final divide by eight stage (output of the RTI) is cleared.

### 8.3 MFT REGISTERS

### 8.3.1 Timer Counter Register (TCNT) \$09

The Timer Counter Register is a read-only register which contains the current value of the 8-bit ripple counter at the beginning of the timer chain. This counter is clocked at  $f_{OP} \div 4$  and can be used for various functions including a software input capture. Extended time periods can be attained using the CTOF function to increment a temporary RAM storage location thereby simulating a 16-bit (or more) counter. The value of each bit of the TCNT is shown in **Figure 8-2**. This register is cleared by reset.



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November 5, 1998 GENERAL RELEASE SPECIFICATION

Figure 8-2. Timer Counter Register

### 8.3.2 Timer Control/Status Register (TCSR) \$08

The TCSR contains the timer interrupt flag bits, the timer interrupt enable bits, and the real time interrupt rate select bits. Bit 2 and bit 3 are write-only bits which will read as logical zeros. **Figure 8-3** shows the value of each bit in the TCSR following reset.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
TCSR	R	CTOF	RTIF	CTOFE	RTIE	0	0	RT1	RT0
\$0008	W			CIOFE	KIIE	CTOFR	RTIFR	KII	KIU
rese	t:	0	0	0	0	0	0	1	1

Figure 8-3. Timer Control/Status Register (TCSR)

## RT0, RT1 — Real-Time Interrupt period select bits

These two bits select the Real-Time Interrupt period and the COP Watchdog reset period.

Table 8-1. RTI and COP Rates at f<sub>OP</sub>=3.0MHz

			Bus Frequency	, f <sub>BUS</sub> =f <sub>OP</sub> =3.0 MHz
RT1	RT0	Divide Ratio	RTI Rate	COP Reset Period (RTI × 8)
0	0	2 <sup>14</sup>	5.46ms	43.68ms
0	1	2 <sup>15</sup>	10.92ms	87.36ms
1	0	2 <sup>16</sup>	21.85 ms	174.8ms
1	1	2 <sup>17</sup>	43.69ms	349.52ms

### RTIFR — Real Time Interrupt Acknowledge

The RTIFR is an acknowledge bit that resets the RTIF flag bit. This bit is unaffected by reset. Reading the RTIFR will always return a logical zero.

- 1 = Clears the RTIF flag bit.
- 0 = Does not clear the RTIF flag bit.

### CTOFR — Timer Overflow Acknowledge

The CTOFR is an acknowledge bit that resets the CTOF flag bit. This bit is unaffected by reset. Reading the CTOFR will always return a logical zero.

- 1 = Clears the CTOF flag bit.
- 0 = Does not clear the CTOF flag bit.

MC68HC05JB3 REV 1



GENERAL RELEASE SPECIFICATION November 5, 1998

# RTIE — Real Time Interrupt Enable

The RTIE is an enable bit that allows generation of a TIMER Interrupt by the RTIF bit.

- 1 = When set, the TIMER Interrupt is generated when the RTIF flag bit is set.
- 0 = When cleared, no TIMER interrupt caused by RTIF bit set will be generated. This bit is cleared by reset.

### **CTOFE** — Timer Overflow Enable

The CTOFE is an enable bit that allows generation of a TIMER Interrupt upon overflow of the Timer Counter Register.

- 1 = When set, the TIMER Interrupt is generated when the CTOF flag bit is set.
- 0 = When cleared, no TIMER interrupt caused by CTOF bit set will be generated. This bit is cleared by reset.

### RTIF — Real Time Interrupt Flag

The RTIF is a read-only flag bit.

- 1 = Set when the output of the chosen (1 of 4 selections) Real Time Interrupt stage goes active. A TIMER Interrupt request will be generated if RTIE is also set.
- 0 = Reset by writing a logical one to the RTIF acknowledge bit, RTIFR. Writing to the RTIF flag bit has no effect on its value. This bit is cleared by reset.

### CTOF — Timer Overflow Flag

The CTOF is a read-only flag bit.

- 1 = Set when the 8-bit ripple counter rolls over from \$FF to \$00. A TIMER Interrupt request will be generated if CTOFE is also set.
- 0 = Reset by writing a logical one to the CTOF acknowledge bit, CTOFR. Writing to the CTOF flag bit has no effect on its value. This bit is cleared by reset.

#### 8.4 OPERATION DURING STOP MODE

When STOP is exited by an external interrupt or an LVR reset or an external RESET, the internal oscillator will resume, followed by a 224 or 4064 internal processor oscillator stabilization delay.

### 8.5 COP CONSIDERATION DURING STOP MODE

In STOP mode, the clock to the Watchdog Timer is stopped and is therefore impossible to generate COP reset when in STOP mode. The COP function will resume 224 or 4064 cycles after exiting from STOP.

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# **SECTION 9 16-BIT TIMER**

This 16-bit Programmable Timer (Timer1) has an Input Capture function and an Output Compare function. Figure 9-1 shows a block diagram of the 16-bit programmable timer.

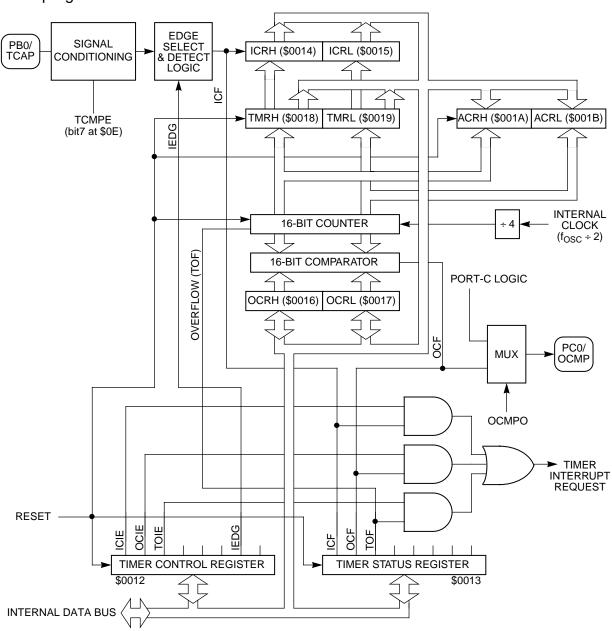


Figure 9-1. Programmable Timer Block Diagram



GENERAL RELEASE SPECIFICATION November 5, 1998

The basis of the 16-bit Timer is a 16-bit free-running counter which increases in count with each internal bus clock cycle. The counter is the timing reference for the input capture and output compare functions. The input capture and output compare functions provide a means to latch the times at which external events occur, to measure input waveforms, and to generate output waveforms and timing delays. Software can read the value in the 16-bit free-running counter at any time without affect the counter sequence.

Because of the 16-bit timer architecture, the I/O registers for the input capture and output compare functions are pairs of 8-bit registers. Each register pair contains the high and low byte of that function. Generally, accessing the low byte of a specific timer function allows full control of that function; however, an access of the high byte inhibits that specific timer function until the low byte is also accessed.

Because the counter is 16 bits long and preceded by a fixed divide-by-four prescaler, the counter rolls over every 262,144 internal clock cycles. Timer resolution with a 4MHz crystal oscillator is 2 microsecond/count.

The interrupt capability, the input capture edge, and the output compare state are controlled by the timer control register (TCR) located at \$0012 and the status of the interrupt flags can be read from the timer status register (TSR) located at \$0013.

### 9.1 TIMER REGISTERS (TMRH, TMRL)

The functional block diagram of the 16-bit free-running timer counter and timer registers is shown in **Figure 9-2**. The timer registers include a transparent buffer latch on the LSB of the 16-bit timer counter.

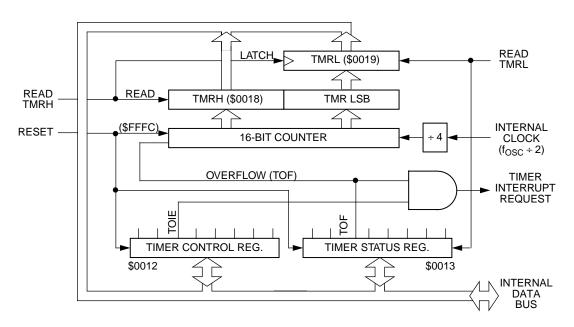


Figure 9-2. Programmable Timer Counter Block Diagram



November 5, 1998 GENERAL RELEASE SPECIFICATION

The timer registers (TMRH, TMRL) shown in **Figure 9-3** are read-only locations which contain the current high and low bytes of the 16-bit free-running counter. Writing to the timer registers has no effect. Reset of the device presets the timer counter to \$FFFC.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
TMRH	R	TMRH7	TMRH6	TMRH5	TMRH4	TMRH3	TMRH2	TMRH1	TMRH0
\$0018	W								
reset:		1	1	1	1	1	1	1	1
TMRL	R	TMRL7	TMRL6	TMRL5	TMRL4	TMRL3	TMRL2	TMRL1	TMRL0
\$0019	W								
reset		1	1	1	1	1	1	0	0

Figure 9-3. Programmable Timer Counter Registers (TMRH, TMRL)

The TMRL latch is a transparent read of the LSB until the a read of the TMRH takes place. A read of the TMRH latches the LSB into the TMRL location until the TMRL is again read. The latched value remains fixed even if multiple reads of the TMRH take place before the next read of the TMRL. Therefore, when reading the MSB of the timer at TMRH the LSB of the timer at TMRL must also be read to complete the read sequence.

During power-on-reset (POR), the counter is initialized to \$FFFC and begins counting after the oscillator start-up delay. Because the counter is sixteen bits and preceded by a fixed divide-by-four prescaler, the value in the counter repeats every 262, 144 internal bus clock cycles (524, 288 oscillator cycles).

When the free-running counter rolls over from \$FFFF to \$0000, the timer overflow flag bit (TOF) is set in the TSR. When the TOF is set, it can generate an interrupt if the timer overflow interrupt enable bit (TOIE) is also set in the TCR. The TOF flag bit can only be reset by reading the TMRL after reading the TSR.

Other than clearing any possible TOF flags, reading the TMRH and TMRL in any order or any number of times does not have any effect on the 16-bit free-running counter.

### **NOTE**

To prevent interrupts from occurring between readings of the TMRH and TMRL, set the I bit in the condition code register (CCR) before reading TMRH and clear the I bit after reading TMRL.



GENERAL RELEASE SPECIFICATION November 5, 1998

# 9.2 ALTERNATE COUNTER REGISTERS (ACRH, ACRL)

The functional block diagram of the 16-bit free-running timer counter and alternate counter registers is shown in **Figure 9-4**. The alternate counter registers behave the same as the timer registers, except that any reads of the alternate counter will not have any effect on the TOF flag bit and Timer interrupts. The alternate counter registers include a transparent buffer latch on the LSB of the 16-bit timer counter.

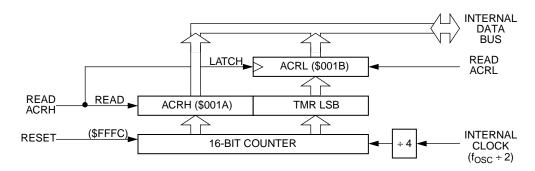


Figure 9-4. Alternate Counter Block Diagram

The alternate counter registers (ACRH, ACRL) shown in **Figure 9-5** are read-only locations which contain the current high and low bytes of the 16-bit free-running counter. Writing to the alternate counter registers has no effect. Reset of the device presets the timer counter to \$FFFC.

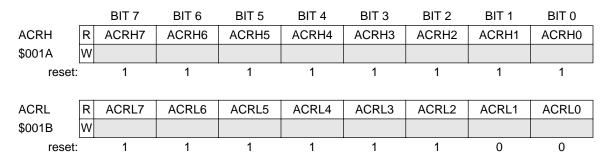


Figure 9-5. Alternate Counter Registers (ACRH, ACRL)

The ACRL latch is a transparent read of the LSB until the a read of the ACRH takes place. A read of the ACRH latches the LSB into the ACRL location until the ACRL is again read. The latched value remains fixed even if multiple reads of the ACRH take place before the next read of the ACRL. Therefore, when reading the MSB of the timer at ACRH the LSB of the timer at ACRL must also be read to complete the read sequence.

During power-on-reset (POR), the counter is initialized to \$FFFC and begins counting after the oscillator start-up delay. Because the counter is sixteen bits and preceded by a fixed divide-by-four prescaler, the value in the counter repeats every 262,144 internal bus clock cycles (524,288 oscillator cycles).

Reading the ACRH and ACRL in any order or any number of times does not have any effect on the 16-bit free-running counter or the TOF flag bit.



November 5, 1998 GENERAL RELEASE SPECIFICATION

#### NOTE

To prevent interrupts from occurring between readings of the ACRH and ACRL, set the I bit in the condition code register (CCR) before reading ACRH and clear the I bit after reading ACRL.

### 9.3 INPUT CAPTURE REGISTERS

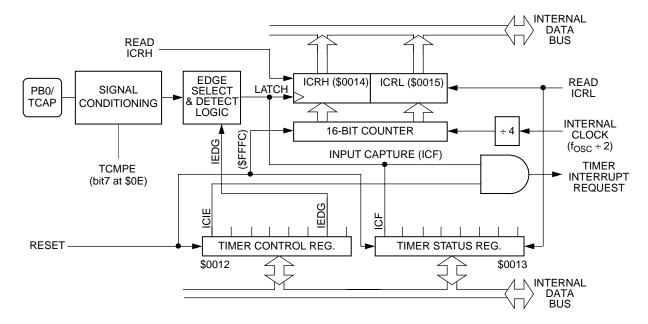


Figure 9-6. Timer Input Capture Block Diagram

The input capture function is a technique whereby an external signal (connected to PB0/TCAP pin) is used to trigger the 16-bit timer counter. In this way it is possible to relate the timing of an external signal to the internal counter value, and hence to elapsed time.

#### NOTE

Since the TCAP pin is shared with the PB0 I/O pin, changing the state of the PB0 DDR or Data Register can cause an unwanted TCAP interrupt. This can be avoided by clearing the ICIE bit before changing the configuration of PB0, and clearing any pending interrupts before enabling ICIE.

The signal on the TCAP pin is first directed to a schmitt trigger or a voltage comparator as shown in **Figure 9-7**. Setting the TCMPE bit to "1" will enable the comparator and the  $V_{DD}/2$  reference voltage.



GENERAL RELEASE SPECIFICATION November 5, 1998

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
OIER \$000E	R W	ТСМРЕ	VREF2	VREF1	VREF0	OIE3	OIE2	OIE1	OIE0
*									
reset:		0	0	0	0	0	0	0	0

### **TCMPE** — Timer Input Capture Comparator Enable

- 1 = Timer input capture comparator is selected.
- 0 = Timer input capture comparator schmitt trigger is selected.

#### NOTE

When the comparator and  $V_{DD}/2$  reference are enabled, PB0 pin will automatically becomes an input pin, irrespective of DDR setting. However, it is recommended to set PB0 as an input first (via DDR), before enabling the comparator. A read of PB0 will reflect the TCAP pin status, not the PB0 register bit.

The comparator uses the  $V_{DD}/2$  reference as the compare voltage, resulting in a typical output as shown in **Figure 9-8**.

Switching off the  $V_{DD}/2$  voltage reference by clearing TCMPE=0 will further save power when the MCU is in a low power mode.

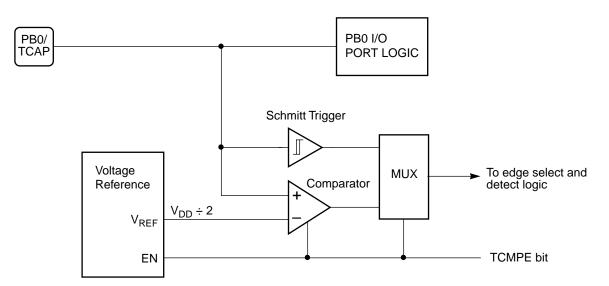


Figure 9-7. TCAP Input Signal Conditioning



November 5, 1998 GENERAL RELEASE SPECIFICATION

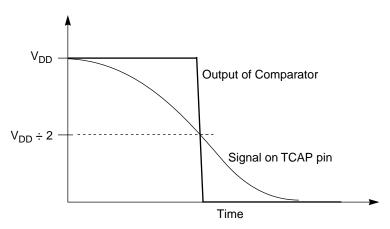


Figure 9-8. TCAP Input Comparator Output

When the input capture circuitry detects an active edge on the TCAP pin, it latches the contents of the free-running timer counter registers into the input capture registers as shown in **Figure 9-6**.

Latching values into the input capture registers at successive edges of the same polarity measures the period of the selected input signal. Latching the counter values at successive edges of opposite polarity measures the pulse width of the signal.

The input capture registers are made up of two 8-bit read-only registers (ICRH, ICRL) as shown in **Figure 9-9**. The input capture edge detector contains a Schmitt trigger to improve noise immunity. The edge that triggers the counter transfer is defined by the input edge bit (IEDG) in the TCR. Reset does not affect the contents of the input capture registers.

The result obtained by an input capture will be one count higher than the value of the free-running timer counter preceding the external transition. This delay is required for internal synchronization. Resolution is affected by the prescaler, allowing the free-running timer counter to increment once every four internal clock cycles (eight oscillator clock cycles).

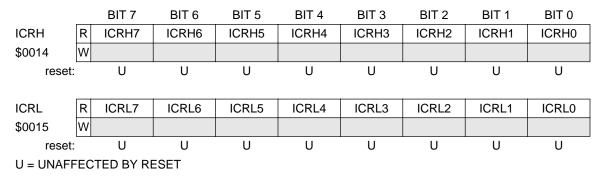


Figure 9-9. Input Capture Registers (ICRH, ICRL)



GENERAL RELEASE SPECIFICATION November 5, 1998

Reading the ICRH inhibits further captures until the ICRL is also read. Reading the ICRL after reading the timer status register (TSR) clears the ICF flag bit. does not inhibit transfer of the free-running counter. There is no conflict between reading the ICRL and transfers from the free-running timer counters. The input capture registers always contain the free-running timer counter value which corresponds to the most recent input capture.

#### NOTE

To prevent interrupts from occurring between readings of the ICRH and ICRL, set the I bit in the condition code register (CCR) before reading ICRH and clear the I bit after reading ICRL.

#### 9.4 OUTPUT COMPARE REGISTERS

The Output Compare function is a means of generating an interrupt when the 16-bit timer counter reaches a selected value as shown in **Figure 9-10**. Software writes the selected value into the output compare registers. On every fourth internal clock cycle (every eight oscillator clock cycle) the output compare circuitry compares the value of the free-running timer counter to the value written in the output compare registers. When a match occurs, the output compare interrupt flag, OCF is set. A timer interrupt request to the CPU is generated if the output compare interrupt enable is set, i.e. OCIE=1.

Port pin, PC0 is configured as the OCMP output pin when the OCMPO bit (bit7 at \$06) is set to "1". The OCMP output reflects the logic of the output compare interrupt flag, OCF, as shown in **Figure 9-10**.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
DDRC	R	ОСМРО	VROFF			DDRC3	DDRC2	DDRC1	DDRC0
\$0006	W	OCIVIPO	VKOFF			DDRC3	DDRC2	DDRCI	DDRCU
rese	et:	0	0	0	0	0	0	0	0

### OCMPO — OCMP Output Enable

- 1 = PC0 is OCMP pin, OCF from 16-bit timer output compare.
- 0 = PC0 is standard I/O pin, from Port-C data register.

Software can use the output compare register to measure time periods, to generate timing delays, or to generate a pulse of specific duration or a pulse train of specific frequency and duty cycle.

Writing to the OCRH before writing to the OCRL inhibits timer compares until the OCRL is written. Reading or writing to the OCRL after reading the TSR will clear the output compare flag bit (OCF).



November 5, 1998 GENERAL RELEASE SPECIFICATION

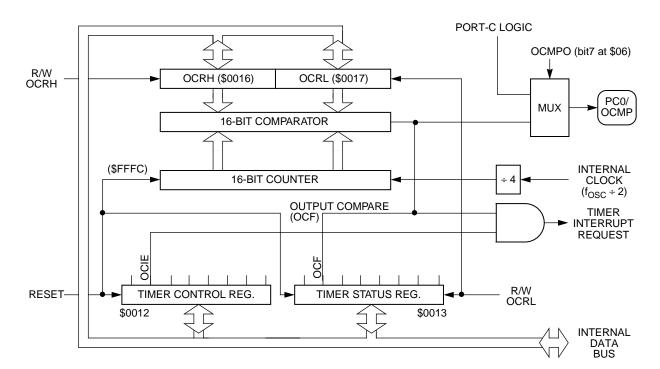


Figure 9-10. Timer Output Compare Block Diagram

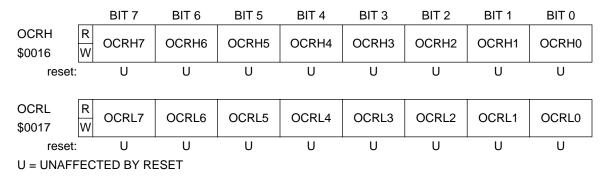


Figure 9-11. Output Compare Registers (OCRH, OCRL)

To prevent OCF from being set between the time it is read and the time the output compare registers are updated, use the following procedure:

- 1. Disable interrupts by setting the I bit in the condition code register.
- 2. Write to the OCRH. Compares are now inhibited until OCRL is written.
- 3. Read the TSR to arm the OCF for clearing.
- 4. Enable the output compare registers by writing to the OCRL. This also clears the OCF flag bit in the TSR.
- 5. Enable interrupts by clearing the I bit in the condition code register.

A software example of this procedure is shown below.



GENERAL RELEASE SPECIFICATION November 5, 1998

9B		SEI		DISABLE INTERRUPTS
•••		•••		••••
• • •		• • •		••••
в7	16	STA	OCRH	INHIBIT OUTPUT COMPARE
В6	13	LDA	TSR	ARM OCF FLAG FOR CLEARING
BF	17	STX	OCRL	READY FOR NEXT COMPARE, OCF CLEARED
•••		•••		••••
• • •		• • •		••••
9A		CLI		ENABLE INTERRUPTS

## 9.5 TIMER CONTROL REGISTER (TCR)

The timer control register is shown in **Figure 9-12** performs the following functions:

- Enables input capture interrupts
- Enables output compare interrupts
- Enables timer overflow interrupts
- Control the active edge polarity of the TCAP signal on pin PB0/TCAP

Reset clears all the bits in the TCR with the exception of the IEDG bit which is unaffected.

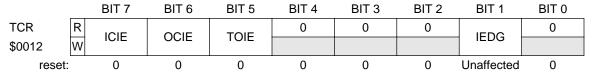


Figure 9-12. Timer Control Register (TCR)

## ICIE - INPUT CAPTURE INTERRUPT ENABLE

This read/write bit enables interrupts caused by an active signal on the PB0/TCAP pin. Reset clears the ICIE bit.

- 1 = Input capture interrupts enabled.
- 0 = Input capture interrupts disabled.

### OCIE - OUTPUT COMPARE INTERRUPT ENABLE

This read/write bit enables interrupts caused by a successful compare between the timer counter and the output compare registers. Reset clears the OCIE bit.

- 1 = Output compare interrupts enabled.
- 0 = Output compare interrupts disabled.

### **TOIE - TIMER OVERFLOW INTERRUPT ENABLE**

This read/write bit enables interrupts caused by a timer overflow. Reset clears the TOIE bit.

- 1 = Timer overflow interrupts enabled.
- 0 = Timer overflow interrupts disabled.



November 5, 1998 GENERAL RELEASE SPECIFICATION

### **IEDG - INPUT CAPTURE EDGE SELECT**

The state of this read/write bit determines whether a positive or negative transition on the TCAP pin triggers a transfer of the contents of the timer register to the input capture register. Reset has no effect on the IEDG bit.

- 1 = Positive edge (low to high transition) triggers input capture.
- 0 = Negative edge (high to low transition) triggers input capture.

### 9.6 TIMER STATUS REGISTER (TSR)

The timer status register (TSR) shown in **Figure 9-13** contains flags for the following events:

- An active signal on the PB0/TCAP pin, transferring the contents of the timer registers to the input capture registers.
- A match between the 16-bit counter and the output compare registers
- An overflow of the timer registers from \$FFFF to \$0000.

Writing to any of the bits in the TSR has no effect. Reset does not change the state of any of the flag bits in the TSR.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
TSR	R	ICF	OCF	TOF	0	0	0	0	0
\$0013	W								
reset	:	U	U	U	0	0	0	0	0
U = UNAF	FEC1	TED BY RE	ESET						

Figure 9-13. Timer Status Registers (TSR)

### **ICF - INPUT CAPTURE FLAG**

The ICF bit is automatically set when an edge of the selected polarity occurs on the PB0/TCAP pin. Clear the ICF bit by reading the timer status register with the ICF set, and then reading the low byte (ICRL, \$0015) of the input capture registers. Reset has no effect on ICF.

### OCF - OUTPUT COMPARE FLAG

The OCF bit is automatically set when the value of the timer registers matches the contents of the output compare registers. Clear the OCF bit by reading the timer status register with the OCF set, and then accessing the low byte (OCRL, \$0017) of the output compare registers. Reset has no effect on OCF.

OCF status will be latched to the output of OCMP (PC0 pin) if the OCMPO bit is set to "1" (bit7 at \$06).

### **TOF-TIMER OVERFLOW FLAG**

The TOF bit is automatically set when the 16-bit timer counter rolls over from \$FFFF to \$0000. Clear the TOF bit by reading the timer status register with the TOF set, and then accessing the low byte (TMRL, \$0019) of the timer registers. Reset has no effect on TOF.



GENERAL RELEASE SPECIFICATION November 5, 1998

### 9.7 TIMER OPERATION DURING WAIT MODE

During WAIT mode the 16-bit timer continues to operate normally and may generate an interrupt to trigger the MCU out of the WAIT mode.

### 9.8 TIMER OPERATION DURING STOP MODE

When the MCU enters the STOP mode the free-running counter stops counting (the internal processor clock is stopped). It remains at that particular count value until the STOP mode is exited by applying a low signal to the  $\overline{IRQ}$  pin, at which time the counter resumes from its stopped value as if nothing had happened. If STOP mode is exited via an external reset (logic low applied to the  $\overline{RESET}$  pin) the counter is forced to \$FFFC.

If a valid input capture edge occurs at the PB0/TCAP pin during the STOP mode the input capture detect circuitry will be armed. This action does not set any flags or "wake up" the MCU, but when the MCU does "wake up" there will be an active input capture flag (and data) from the first valid edge. If the STOP mode is exited by an external reset, no input capture flag or data will be present even if a valid input capture edge was detected during the STOP mode.

# SECTION 10 UNIVERSAL SERIAL BUS MODULE

This USB Module is designed for USB application in LS products. With minimized software effort, it can fully comply with USB LS device specification. See USB specification version 1.0 for the detail description of USB.

### 10.1 FEATURES

- Integrated 3.3 Volt Regulator with 3.3V Output Pin
- Integrated USB transceiver supporting Low Speed functions
- USB Data Control Logic
  - Packet decoding/generation
  - CRC generation and checking
  - NRZI encoding/decoding
  - Bit-stuffing
- USB reset support
- Control Endpoint 0 and Interrupt Endpoints 1 and 2
- Two 8-byte transmit buffers
- One 8-byte receive buffer
- Suspend and resume operations
- Remote Wake-up support
- USB generated interrupts
- Transaction interrupt driven
- Resume interrupt
- End of Packet interrupt
- STALL, NAK, and ACK handshake generation



GENERAL RELEASE SPECIFICATION November 5, 1998

### 10.2 OVERVIEW

This section provides an overview of the Universal Serial Bus (USB) module in the MC68HC05JB3. This USB module is designed to serve as a low-speed (LS) USB device per the Universal Serial Bus Specification Rev 1.0. Three types of USB data transfers are supported: control, interrupt, and bulk (transmit only). Endpoint 0 functions as a receive/transmit control endpoint. Endpoints 1 and 2 can function as interrupt or bulk, but only in the transmit direction.

A block diagram of the USB module is shown **Figure 10-1**. The USB module manages communications between the host and the USB function. The module is partitioned into four functional blocks. These blocks consist of a 3.3 volt regulator, a dual function transceiver, the USB control logic, and the endpoint registers. The blocks are further detailed in **Section 10.4**.

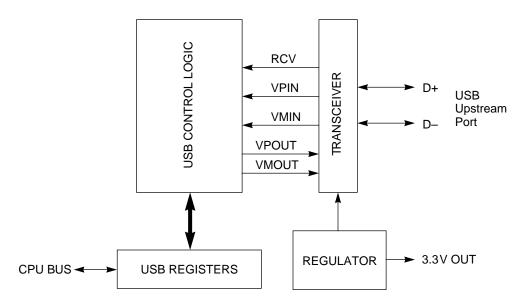


Figure 10-1. USB Block Diagram



November 5, 1998 GENERAL RELEASE SPECIFICATION

### 10.2.1 USB Protocol

**Figure 10-2** shows the various transaction types supported by the MC68HC05JB3 USB module. The transactions are portrayed as error free. The effect of errors in the data flow are discussed later.

#### **ENDPOINT 0 TRANSACTIONS:**

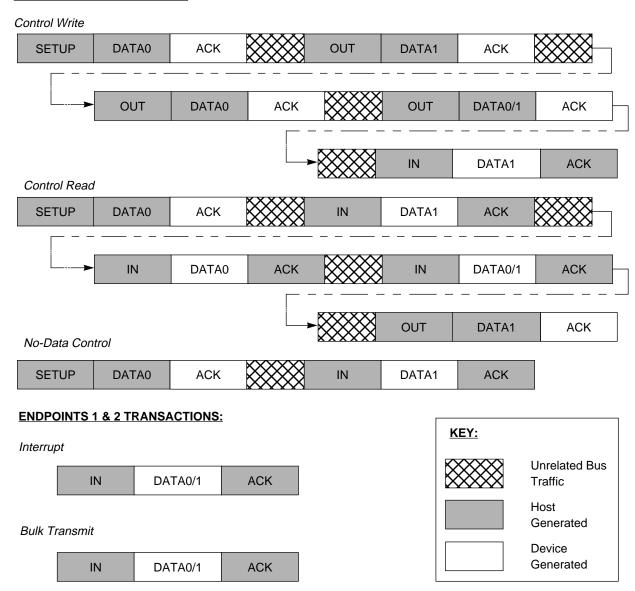


Figure 10-2. Supported Transaction Types per Endpoint

Each USB transaction is comprised of a series of packets. The MC68HC05JB3 USB module supports the packet types shown in **Figure 10-3**. Token packets are generated by the USB host and decoded by the USB device. Data and Handshake packets are both decoded and generated by the USB device depending on the type of transaction.

MC68HC05JB3 REV 1



GENERAL RELEASE SPECIFICATION November 5, 1998

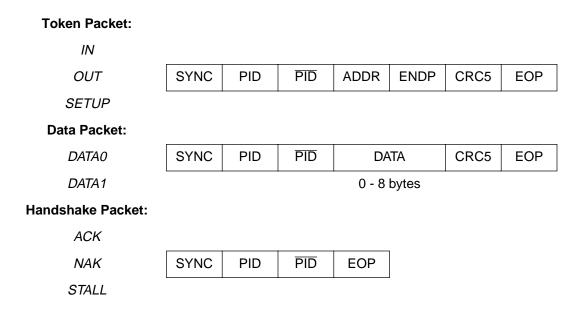


Figure 10-3. Supported USB Packet Types

The following sections will give some detail on each segment used to form a complete USB transaction.

# 10.2.1.1 Sync Pattern

The NRZI (See **Section 10.4.4.1**) bit pattern shown in **Figure 10-4** is used as a synchronization pattern and is prefixed to each packet. This pattern is equivalent to a data pattern of seven 0's followed by a 1 (0x80).

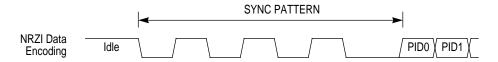


Figure 10-4. Sync Pattern

The start of a packet (SOP) is signaled by the originating port by driving the D+ and D- lines from the idle state (also referred to as the "J" state) to the opposite logic level (also referred to as the "K" state). This switch in levels represents the first bit of the Sync field. **Figure 10-5** shows the data signaling and voltage levels for the start of packet and the sync pattern.

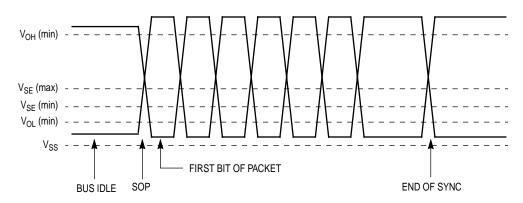


Figure 10-5. SOP, Sync Signaling and Voltage Levels

#### 10.2.1.2 Packet Identifier Field

The Packet Identifier field is an eight bit number comprised of the four bit packet identification (PID) and its complement. The field follows the sync pattern and determines the direction and type of transaction on the bus. **Table 10-1** shows the PID values for the supported packet types.

**Table 10-1. Supported Packet Identifiers** 

PID Value	PID Type
%1001	IN Token
%0001	OUT Token
%1101	SETUP Token
%0011	DATA0 Packet
%1011	DATA1 Packet
%0010	ACK Handshake
%1010	NAK Handshake
%1110	STALL Handshake

## 10.2.1.3 Address Field (ADDR)

The Address field is a seven bit number that is used to select a particular USB device. This field is compared to the lower seven bits of the UADDR register to determine if a given transaction is targeting the MC68HC05JB3 USB device.

The Endpoint field is a four bit number that is used to select a particular endpoint within a USB device. For the MC68HC05JB3, this will be a binary number between zero and two inclusive. Any other value will cause the transaction to be ignored.

## 10.2.1.5 Cyclic Redundancy Check (CRC)

Cyclic Redundancy Checks are used to verify the address and data stream of a USB transaction. This field is five bits wide for token packets and sixteen bits wide for data packets. CRCs are generated in the transmitter and sent on the USB data lines after both the endpoint field and the data field. **Figure 10-6** shows how the five bit CRC value is calculated from the data stream and verified for the address and endpoint fields of a token packet. **Figure 10-7** shows how the sixteen bit CRC value is calculated and either transmitted or verified for the data packet of a given transaction.

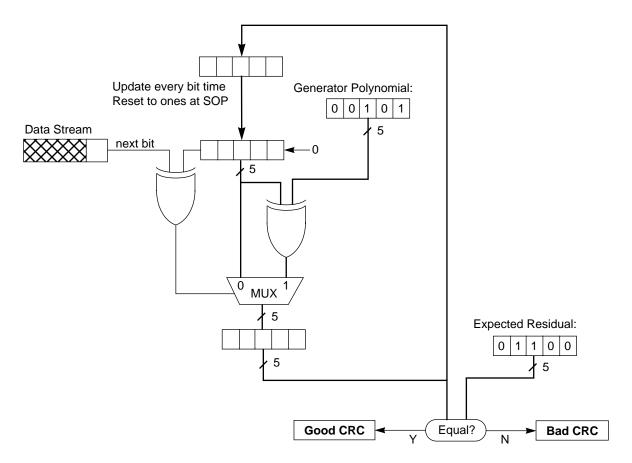


Figure 10-6. CRC Block Diagram for Address and Endpoint Fields

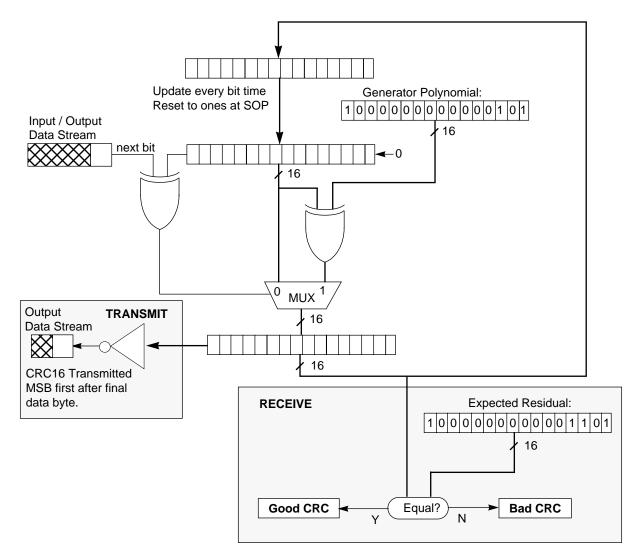


Figure 10-7. CRC Block Diagram for Data Packets

## 10.2.1.6 End Of Packet (EOP)

The single-ended 0 (SE0) state is used to signal an end of packet (EOP). The single-ended 0 state is indicated by both D+ and D- being below 0.8 V. EOP will be signaled by driving D+ and D- to the single-ended 0 state for two bit times followed by driving the lines to the idle state for one bit time. The transition from the single-ended 0 to the idle state defines the end of the packet. The idle state is asserted for one bit time and then both the D+ and D- output drivers are placed in their high-impedance state. The bus termination resistors hold the bus in the idle state. **Figure 10-8** shows the data signaling and voltage levels for an end of packet transaction.

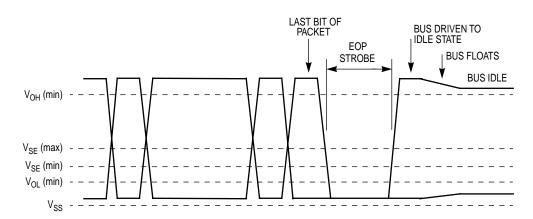


Figure 10-8. EOP Transaction Voltage Levels

The width of the SE0 in the EOP is about two bit times. The EOP width is measured with the same capacitive load used for maximum rise and fall times and is measured at the same level as the differential signal crossover points of the data lines.

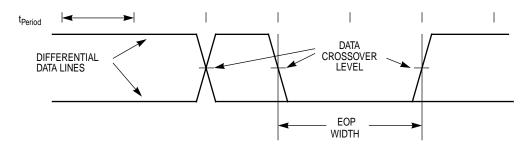


Figure 10-9. EOP Width Timing

## 10.2.2 Reset Signaling

A reset is signaled on the bus by the presence of an extended SE0 at the USB data pins of a device. The reset signaling is specified to be present for a minimum of 10 ms. An active device (powered and not in the suspend state) seeing a single-ended zero on its USB data inputs for more than 2.5us may treat that signal as a reset, but must have interpreted the signaling as a reset within 5.5 µs. For a Low speed device, an SE0 condition between 4 and 8 low speed bit times represents a valid USB reset.

A USB sourced reset will hold the MC68HC05JB3 in reset for the duration of the reset on the USB bus. The RSTF bit in the USB interrupt register 0 (UIR0) will be set after the internal reset is removed (See Section 10.5.2 for more detail).

After a reset is removed, the device will be in the attached, but not yet addressed or configured state (refer to Section 9.1 of the USB specification). The device must be able to accept a device address via a SET ADDRESS command (refer to section 9.4 of the USB specification) no later than 10 ms after the reset is removed.

**MOTOROLA** 10-8



November 5, 1998 GENERAL RELEASE SPECIFICATION

Reset can wake a device from the suspended mode. A device may take up to 10ms to wake up from the suspended state.

# 10.2.3 Suspend

The MC68HC05JB3 supports suspend mode for low power. Suspend mode should be entered when the USB data lines are in the idle state for more than 3.0 ms. Entry into Suspend mode is controlled by the SUSPND bit in the USB Interrupt Register. Any low speed bus activity should keep the device out of the suspend state. Low speed devices are kept awake by periodic low speed EOP signals from the host. This is referred to as Low speed keep alive (refer to Section 11.2.5.1 of the USB specification).

Firmware should monitor the EOPF flag and enter suspend mode by setting the SUSPND bit if an EOP is not detected for 3 ms.

Per the USB specification, the MC68HC05JB3 is required to draw less than 500  $\mu A$  from the  $V_{DD}$  supply when in the suspend state. This includes the current supplied by the voltage regulator to the 15 K $\Omega$  to ground termination resistors placed at the host end of the USB bus. This low current requirement means that firmware is responsible for entering STOP mode once the USB module has been placed in the suspend state.

## 10.2.4 Resume After Suspend

The MC68HC05JB3 can be activated from the suspend state by normal bus activity, a USB reset signal, or by a forced resume driven from the MC68HC05JB3.

#### 10.2.4.1 Host Initiated Resume

The host signals resume by initiating resume signalling ("K" state) for at least 20 ms followed by a standard low speed EOP signal. This 20 ms ensures that all devices in the USB network are awakened.

After resuming the bus, the host must begin sending bus traffic within 3 ms to prevent the device from re-entering suspend mode.

#### 10.2.4.2 USB Reset Signalling

Reset can wake a device from the suspended mode. A device may take up to 10 ms to wake up from the suspended state.

#### 10.2.4.3 Remote Wake-up

The MC68HC05JB3 also supports the remote wake-up feature. The firmware has the ability to exit suspend mode by signaling a resume state to the upstream Host or Hub. A non-idle state ("K" state) on the USB data lines is accomplished by asserting the FRESUM bit in the UCR1 register.

MC68HC05JB3 REV 1 **UNIVERSAL SERIAL BUS MODULE** 

**MOTOROLA** 



GENERAL RELEASE SPECIFICATION November 5, 1998

When using the remote wake-up capability, the firmware must wait for at least 5 ms after the bus is in the idle state before sending the remote wake-up resume signaling. This allows the upstream devices to get into their suspend state and prepare for propagating resume signaling. The FRESUM bit should be asserted to cause the resume state on the USB data lines for at least 10ms, but not more than 15ms. Note that the resume signalling is controlled by the FRESUM bit and meeting the timing specifications is dependent on the firmware. When FRESUM is cleared by firmware, the data lines will return to their high impedance state. Refer to **Section 10.5.5** for more information about how the Force Resume (FRESUM) bit can be used to initiate the remote wake-up feature.

#### 10.2.5 Low Speed Device

Externally, low speed devices are configured by the position of a pull-up resistor on the USB D- pin of the MC68HC05JB3. Low speed devices are terminated as shown in **Figure 10-10** with the pull-up on the D- line.

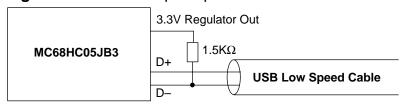


Figure 10-10. External Low Speed Device Configuration

For low speed transmissions, the transmitter's EOP width must be between  $1.25\mu s$  and  $1.50\mu s$ . These ranges include timing variations due to differential buffer delay and rise/fall time mismatches and to noise and other random effects. A low speed receiver must accept a 670ns wide SE0 followed by a J transition as a valid EOP. An SE0 narrower than 330ns or an SE0 not followed by a J transition must be rejected as an EOP. An EOP between 330ns and 670ns may be rejected or accepted as above. Any SE0 that is  $2.5\mu s$  or wider is automatically a reset.

#### 10.3 CLOCK REQUIREMENTS

The low speed data rate is nominally 1.5 Mbs. The OSCXCLK signal driven by the oscillator circuits is the clock source for the USB module and requires that a 6 MHz oscillator circuit be connected to the OSC1 and OSC2 pins. The permitted frequency tolerance for low speed functions is approximately  $\pm 1.5\%$  (15000 ppm). This tolerance includes inaccuracies from all sources: initial frequency accuracy, crystal capacitive loading, supply voltage on the oscillator, temperature, and aging. The jitter in the low speed data rate must be less than 10 ns. This tolerance allows the use of resonators in low cost, low speed devices.

#### 10.4 HARDWARE DESCRIPTION

The USB module as previously shown in **Figure 10-1** contains four functional blocks: a 3.3 volt regulator, a LS USB transceiver, the USB control logic, and the USB registers. The following will detail the function of the regulator, transceiver and control logic. See **Section 10.5** for the register discussion.

MOTOROLA 10-10 UNIVERSAL SERIAL BUS MODULE

MC68HC05JB3

## 10.4.1 Voltage Regulator

The USB data lines are required by the USB Specification to have a maximum output voltage between 2.8V and 3.6V. The data lines are also required to have an external  $1.5 \mathrm{K}\Omega$  pullup resistor connected between a data line and a voltage source between 3.0V and 3.6V. Since the power provided by the USB cable is specified to be between 4.4V and 5.0V, an on-chip regulator is used to drop the voltage to the appropriate level for sourcing the USB transceiver and external pullup resistor. An output pin driven by the regulator voltage is provided to source the  $1.5 \mathrm{K}\Omega$  external resistor. **Figure 10-11** shows the worst case electrical connection for the voltage regulator.

This regulator can be switched off by user program to save power when the device is in suspend mode. Please note that if the regulator is off, the D– line should be tied to another voltage source with an external pull-up resistor.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	
DDRC	R	ОСМРО	VROFF			DDRC3	DDRC2	DDRC1	DDRC0	
\$0006	W	00	*****			DDITOO	551(62	221(01		
reset	:	0	0	0	0	0	0	0	0	

#### VROFF — USB 3.3V Voltage Reference

The 3.3V Voltage Regulator for the USB transmitter and external D– pull-up can be switched off to reduce power consumption when device is in suspend mode.

- 1 = Disable 3.3V regulator.
- 0 = Enables 3.3V regulator.

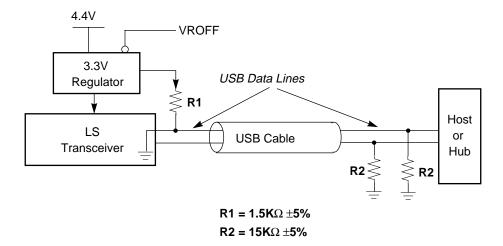


Figure 10-11. Regulator Electrical Connections



GENERAL RELEASE SPECIFICATION November 5, 1998

#### 10.4.2 USB Transceiver

The USB transceiver provides the physical interface to the USB D+ and D- data lines. The transceiver is composed of two parts: an output drive circuit and a differential receiver.

#### 10.4.2.1 Output Driver Characteristics

The USB transceiver uses a differential output driver to drive the USB data signal onto the USB cable. The static output swing of the driver in its low state is below the  $V_{OL}$  of 0.3 V with a 1.5 k $\Omega$  load to 3.6 V and in its high state is above the  $V_{OH}$  of 2.8 V with a 15 k $\Omega$  load to ground. The output swings between the differential high and low state are well balanced to minimize signal skew. Slew rate control on the driver is used to minimize the radiated noise and cross talk. The driver's outputs support three-state operation to achieve bi-directional half duplex operation. The driver can tolerate a voltage on the signal pins of –0.5 V to 3.8 V with respect to local ground reference without damage.

# 10.4.2.2 Low Speed (1.5 Mbs) Driver Characteristics

The rise and fall time of the signals on this cable are greater than 75 ns to keep RFI emissions under FCC class B limits, and less than 300 ns to limit timing delays and signaling skews and distortions. The driver reaches the specified static signal levels with smooth rise and fall times, and minimal reflections and ringing when driving the cable. This driver is used only on network segments between low speed devices and the ports to which they are connected.

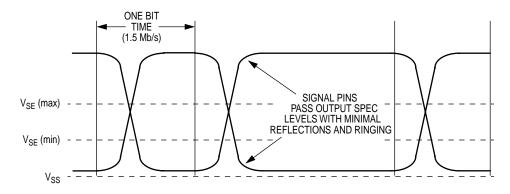


Figure 10-12. Low Speed Driver Signal Waveforms

#### 10.4.3 Receiver Characteristics

USB data transmission is done with differential signals. A differential input receiver is used to accept the USB data signal. A differential 1 on the bus is represented by D+ being at least 200 mV more positive than D- as seen at the receiver, and a differential 0 is represented by D- being at least 200 mV more positive than D+ as seen at the receiver. The signal cross over point must be between 1.3V and 2.0V.

MOTOROLA 10-12 The receiver features an input sensitivity of 200 mV when both differential data inputs are in the range of 0.8 V to 2.5 V with respect to the local ground reference. This is called the common mode input voltage range. Proper data reception is also achieved when the differential data lines are outside the common mode range, as shown in **Figure 10-13**. The receiver can tolerate static input voltages between -0.5 V to 3.8 V with respect to its local ground reference without damage. In addition to the differential receiver, there is a single-ended receiver (schmitt trigger) for each of the two data lines.

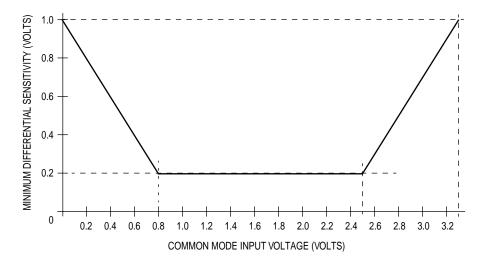


Figure 10-13. Differential Input Sensitivity Over Entire Common Mode Range

#### 10.4.3.1 Receiver Data Jitter

The data receivers for all types of devices must be able to properly decode the differential data in the presence of jitter. The more of the bit cell that any data edge can occupy and still be decoded, the more reliable the data transfer will be. Data receivers are required to decode differential data transitions that occur in a window plus and minus a nominal quarter bit cell from the nominal (centered) data edge position.

Jitter will be caused by the delay mismatches and by mismatches in the source and destination data rates (frequencies). The receive data jitter budget for low speed is given in the electrical section of the this specification. The specification includes the consecutive (next) and paired transition values for each source of jitter.

#### 10.4.3.2 Data Source Jitter

The source of data can have some variation (jitter) in the timing of edges of the data transmitted. The time between any set of data transitions is N x  $T_{PERIOD} \pm$  jitter time, where 'N' is the number of bits between the transitions and  $T_{PERIOD}$  is defined as the actual period of the data rate. The data jitter is

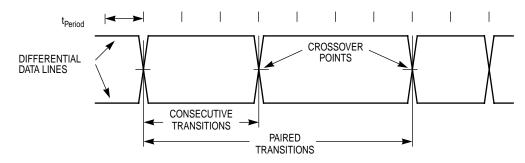
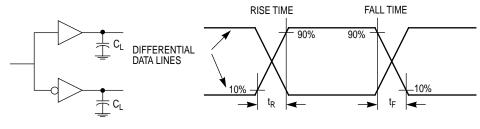


Figure 10-14. Data Jitter

For low speed transmissions, the jitter time for any consecutive differential data transitions must be within  $\pm 25$  ns and within  $\pm 10$  ns for any set of paired differential data transitions. These jitter numbers include timing variations due to differential buffer delay, rise/fall time mismatches, internal clock source jitter, and to noise and other random effects.

## 10.4.3.3 Data Signal Rise and Fall Time

The output rise time and fall time are measured between 10% and 90% of the signal. Edge transition time for the rising and falling edges of low speed signals is 75 ns (minimum) into a capacitive load ( $C_L$ ) of 50 pF and 300 ns (maximum) into a capacitive load of 350 pF. The rising and falling edges should be smooth transitional (monotonic) when driving the cable to avoid excessive EMI.



LOW SPEED: 75 ns at  $C_1 = 50$  pF, 300 ns at  $C_1 = 350$  pF

Figure 10-15. Data Signal Rise and Fall Time

## **10.4.4 USB Control Logic**

The USB control logic manages data movement between the CPU and the transceiver. The control logic handles both transmit and receive operations on the USB. It contains the logic used to manipulate the transceiver and the endpoint registers. The logic contains byte count buffers for transmit operations that load the active transmit endpoints byte count and use this to determine the number of bytes to transfer. This same buffer is used for receive transactions to count the number of bytes received and, upon the end of the transaction, transfer that number to the receive endpoints byte count register.

MOTOROLA 10-14 UNIVERSAL SERIAL BUS MODULE

MC68HC05JB3

REV<sub>1</sub>

November 5, 1998 GENERAL RELEASE SPECIFICATION

When transmitting, the control logic handles parallel to serial conversion, CRC generation, NRZI encoding, and bit stuffing.

When Receiving, the control logic handles Sync detection, packet identification, end of packet detection, bit (un)stuffing, NRZI decoding, CRC validation, and serial to parallel conversion. Errors detected by the control logic include bad CRC, time-out while waiting for EOP, and bit stuffing violations.

## 10.4.4.1 Data Encoding/Decoding

The USB employs NRZI data encoding when transmitting packets. In NRZI encoding, a 1 is represented by no change in level and a 0 is represented by a change in level. **Figure 10-16** shows a data stream and the NRZI equivalent and **Figure 10-17** is a flow diagram for NRZI. The high level represents the J state on the data lines in this and subsequent figures showing NRZI encoding. A string of zeros causes the NRZI data to toggle each bit time. A string of ones causes long periods with no transitions in the data.

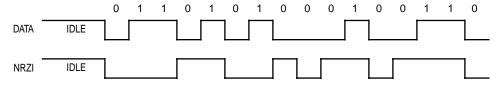


Figure 10-16. NRZI Data Encoding

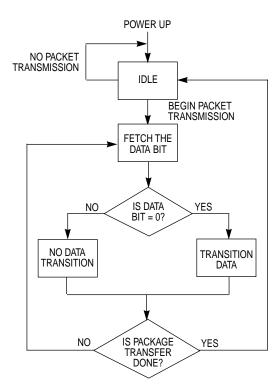


Figure 10-17. Flow Diagram for NRZI



GENERAL RELEASE SPECIFICATION November 5, 1998

#### 10.4.4.2 Bit Stuffing

In order to ensure adequate signal transitions, bit stuffing is employed by the transmitting device when sending a packet on the USB (see **Figure 10-18** and **Figure 10-19**). A 0 is inserted after every six consecutive 1's in the data stream before the data is NRZI encoded to force a transition in the NRZI data stream. This gives the receiver logic a data transition at least once every seven bit times to guarantee the data and clock lock. The receiver must decode the NRZI data, recognize the stuffed bits, and discard them. Bit stuffing is enabled beginning with the Sync Pattern and throughout the entire transmission. The data "one" that ends the Sync Pattern is counted as the first one in a sequence. Bit stuffing is always enforced, without exception. If required by the bit stuffing rules, a zero bit will be inserted even if it is the last bit before the end-of-packet (EOP) signal.

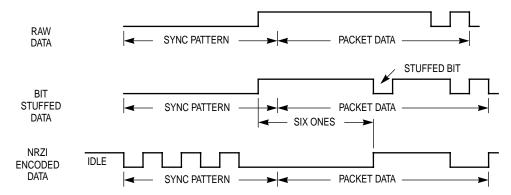


Figure 10-18. Bit Stuffing

GENERAL RELEASE SPECIFICATION



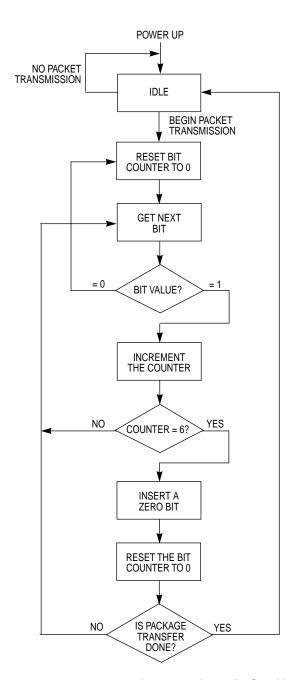


Figure 10-19. Flow Diagram for Bit Stuffing



GENERAL RELEASE SPECIFICATION November 5, 1998

#### 10.5 I/O REGISTER DESCRIPTION

The USB Endpoint registers are comprised of a set of control/status registers and twenty-four data registers that provide storage for the buffering of data between the USB and the CPU. These registers are shown in **Table 10-2**.

**Table 10-2. Register Summary** 

		able it	)-2. IVE	jister St	ıııııaı y				
Register Name	Bit 7	6	5	4	3	2	1	Bit 0	Addr
USB Control Register 2		0	TX1ST	0	ENABLE2	ENIARI E1	STALL2	STALL1	\$0037
(UCR2)		TX1STR			LINABLLZ	LINABLLI	JIALLZ	SIALLI	φυσι
USB Address Register (UADDR)	USBEN	UADD6	UADD5	UADD4	UADD3	UADD2	UADD1	UADD0	\$0038
USB Interrupt Register 0	TXD0F	RXD0F	RSTF	SUSPND	TXD0IE	RXD0IE	0	0	\$0039
(UIR0)				SUSFIND	INDUIL	KADUIE	TXD0FR	RXD0FR	φυυσ9
USB Interrupt Register 1	TXD1F	EOPF	RESUMF	0	TVD4IF	FODIE	0	0	Ф000 A
(UIR1)				RESUMFR	TXD1IE	EOPIE	TXD1FR	EOPFR	\$003A
USB Control Register 0 (UCR0)	T0SEQ	STALL0	TX0E	RX0E	TP0SIZ3	TP0SIZ2	TP0SIZ1	TP0SIZ0	\$003B
USB Control Register 1 (UCR1)	T1SEQ	ENDADD	TX1E	FRESUM	TP1SIZ3	TP1SIZ2	TP1SIZ1	TP1SIZ0	\$003C
USB Status Register	RSEQ	SETUP	0	0	RPSIZ3	RPSIZ2	RPSIZ1	RPSIZ0	\$003D
(USR)									φυυ3D
									_
USB Endpoint 0 Data	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0	\$0020
Register 0 (UE0D0)	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0	Ψ0020
$\downarrow$									$\downarrow$
USB Endpoint 0 Data	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0	<b></b>
Register 7 (UE0D7)	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0	\$0027
									J
USB Endpoint 1/2 Data									\$0028
Register 0 (UE1D0)	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0	\$0026
<b>\</b>									, \
USB Endpoint 1/2 Data									\$002F
Register 7 (UE1D7)	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0	70021
					= Unimple	mented			

November 5, 1998 GENERAL RELEASE SPECIFICATION

## 10.5.1 USB Address Register (UADDR)

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
UADDR \$0038	R W	USBEN	UADD6	UADD5	UADD4	UADD3	UADD2	UADD1	UADD0
reset	t:	0	0	0	0	0	0	0	0

Figure 10-20. USB Address Register (UADDR)

#### **USBEN** — USB Module Enable

This read/write bit enables and disables the USB module and the USB pins. When USBEN is clear, the USB module will not respond to any tokens. Reset clears this bit.

- 1 = USB function enabled.
- 0 = USB function disabled.

#### **UADD6-UADD0** — **USB Function Address**

These bits specify the USB address of the device. Reset clears these bits.

# 10.5.2 USB Interrupt Register 0 (UIR0)

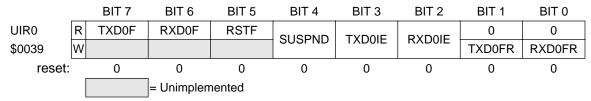


Figure 10-21. USB Interrupt Register 0 (UIR0)

#### TXD0F — Endpoint 0 Data Transmit Flag

This read only bit is set after the data stored in Endpoint 0 transmit buffers has been sent and an ACK handshake packet from the host is received. Once the next set of data is ready in the transmit buffers, software must clear this flag by writing a logic 1 to the TXD0FR bit. To enable the next data packet transmission, TX0E must also be set. If TXD0F bit is not cleared, a NAK handshake will be returned in the next IN transaction.

Reset clears this bit. Writing a logic 0 to TXD0F has no effect.

- 1 = Transmit on Endpoint 0 has occurred.
- 0 = Transmit on Endpoint 0 has not occurred.

#### RXD0F — Endpoint 0 Data Receive Flag

This read only bit is set after the USB module has received a data packet and responded with an ACK handshake packet. Software must clear this flag by writing a logic 1 to the RXD0FR bit after all of the received data has been read.



GENERAL RELEASE SPECIFICATION November 5, 1998

Software must also set RX0E bit to one to enable the next data packet reception. If RXD0F bit is not cleared, a NAK handshake will be returned in the next OUT transaction.

Reset clears this bit. Writing a logic 0 to RXD0F has no effect.

- 1 = Receive on Endpoint 0 has occurred.
- 0 = Receive on Endpoint 0 has not occurred.

# RSTF — USB Reset Flag

This read only bit is set when a valid reset signal state is detected on the D+ and D- lines. This reset detection will also generate an internal reset signal to reset the CPU and other peripherals including the USB module. This bit is cleared by writing a logic 1 to the RSTFR bit in the UCR2 register. This bit is cleared by a POR reset.

# SUSPND — USB Suspend Flag

To save power, this read/write bit should be set by the software if a 3ms constant idle state is detected on USB bus. Setting this bit stops the clock to the USB and causes the USB module to enter Suspend mode. Unnecessary analog circuitry will be powered down. Software must clear this bit after the Resume flag (RESUMF) is set while this Resume interrupt flag is serviced.

#### **TXD0IE** — Endpoint 0 Transmit Interrupt Enable

This read/write bit enables the Transmit Endpoint 0 to generate a USB interrupt when the TXD0F bit becomes set.

- 1 = USB interrupts enabled for Transmit Endpoint 0.
- 0 = USB interrupts disabled for Transmit Endpoint 0.

#### RXD0IE — Endpoint 0 Receive Interrupt Enable

This read/write bit enables the Transmit Endpoint 0 to generate a USB interrupt when the RXD0F bit becomes set.

- 1 = USB interrupts enabled for Receive Endpoint 0.
- 0 = USB interrupts disabled for Receive Endpoint 0.

#### TXD0FR — Endpoint 0 Transmit Flag Reset

Writing a logic 1 to this write only bit will clear the TXD0F bit if it is set. Writing a logic 0 to TXD0FR has no effect. Reset clears this bit.

#### RXD0FR — Endpoint 0 Receive Flag Reset

Writing a logic 1 to this write only bit will clear the RXD0F bit if it is set. Writing a logic 0 to RXD0FR has no effect. Reset clears this bit.

November 5, 1998 GENERAL RELEASE SPECIFICATION

## 10.5.3 USB Interrupt Register 1 (UIR1)

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
UIR1	R	TXD1F	EOPF	RESUMF	0	TXD1IE	EOPIE	0	0
\$003A	W				RESUMFR	IVDIIE	LOPIE	TXD1FR	EOPFR
reset:		0	0	0	0	0	0	0	0
			= Unimplem	nented					

Figure 10-22. USB Interrupt Register 1(UIR1)

## TXD1F — Endpoint 1/Endpoint 2 Data Transmit Flag

This read only bit is shared by Endpoint 1 and Endpoint 2. It is set after the data stored in the shared Endpoint 1/Endpoint 2 transmit buffer has been sent and an ACK handshake packet from the host is received. Once the next set of data is ready in the transmit buffers, software must clear this flag by writing a logic 1 to the TXD1FR bit. To enable the next data packet transmission, TX1E must also be set. If TXD1F bit is not cleared, a NAK handshake will be returned in the next IN transaction.

Reset clears this bit. Writing a logic 0 to TXD1F has no effect.

- 1 = Transmit on Endpoint 1 or Endpoint 2 has occurred.
- 0 = Transmit on Endpoint 1 or Endpoint 2 has not occurred.

#### **EOPF** — End of Packet Detect Flag

This read only bit is set when a valid End-of-Packet sequence is detected on the D+ and D- lines. Software must clear this flag by writing a logic 1 to the EOPFR bit.

Reset clears this bit. Writing a logic 0 to EOPF has no effect.

- 1 = End-of-Packet sequence has been detected.
- 0 = End-of-Packet sequence has not been detected.

#### **RESUMF** — Resume Flag

This read only bit is set when USB bus activity is detected while the SUSPND bit is set. Software must clear this flag by writing a logic 1 to the RESUMFR bit. Reset clears this bit. Writing a logic 0 to RESUMF has no effect.

- 1 = USB bus activity has been detected.
- 0 = No USB bus activity has been detected.

## RESUMFR — Resume Flag Reset

Writing a logic 1 to this write only bit will clear the RESUMF bit if it is set. Writing a logic 0 to RESUMFR has no effect. Reset clears this bit.

#### TXD1IE — Endpoint 1/Endpoint 2 Transmit Interrupt Enable

This read/write bit enables the USB to generate an interrupt when the shared Transmit Endpoint 1/Endpoint 2 interrupt flag (TXD1F) bit becomes set. Reset clears this bit.

- 1 = USB interrupts enabled for Transmit Endpoints 1 and 2.
- 0 = USB interrupts disabled for Transmit Endpoints 1 and 2.



GENERAL RELEASE SPECIFICATION November 5, 1998

## **EOPIE** — End of Packet Detect Interrupt Enable

This read/write bit enables the USB to generate an interrupt when the EOPF bit becomes set. Reset clears this bit.

- 1 = USB interrupts enabled for Transmit Endpoints 1 and 2.
- 0 = USB interrupts disabled for Transmit Endpoint 1 and 2.

## TXD1FR — Endpoint 1/Endpoint 2 Transmit Flag Reset

Writing a logic 1 to this write only bit will clear the TXD1F bit if it is set. Writing a logic 0 to TXD1FR has no effect. Reset clears this bit.

#### **EOPFR** — End of Packet Flag Reset

Writing a logic 1 to this write only bit will clear the EOPF bit if it is set. Writing a logic 0 to the EOPFR has no effect. Reset clears this bit.

## 10.5.4 USB Control Register 0 (UCR0)

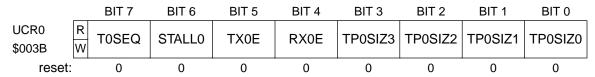


Figure 10-23. USB Control Register 0 (UCR0)

#### T0SEQ — Endpoint 0 Transmit Sequence Bit

This read/write bit determines which type of data packet (DATA0 or DATA1) will be sent during the next IN transaction. Toggling of this bit must be controlled by software. Reset clears this bit.

- 1 = DATA1 Token active for next Endpoint 0 transmit.
- 0 = DATA0 Token active for next Endpoint 0 transmit.

#### STALL0 — Endpoint 0 Force Stall Bit

This read/write bit causes Endpoint 0 to return a STALL handshake when polled by either an IN or OUT token by the USB Host Controller. The USB hardware clears this bit when a SETUP token is received. Reset clears this bit.

- 1 = Send STALL handshake.
- 0 = Default.

## **TX0E** — Endpoint 0 Transmit Enable

This read/write bit enables a transmit to occur when the USB Host controller sends an IN token to Endpoint 0. Software should set this bit when data is ready to be transmitted. It must be cleared by software when no more Endpoint 0 data needs to be transmitted.

If this bit is 0 or the TXD0F is set, the USB will respond with a NAK handshake to any Endpoint 0 IN tokens. Reset clears this bit.

- 1 = Data is ready to be sent.
- 0 = Data is not ready. Respond with NAK.

MOTOROLA 10-22

November 5, 1998 GENERAL RELEASE SPECIFICATION

## RX0E — Endpoint 0 Receive Enable

This read/write bit enables a receive to occur when the USB Host controller sends an OUT token to Endpoint 0. Software should set this bit when data is ready to be received. It must be cleared by software when data cannot be received.

If this bit is 0 or the RXD0F is set, the USB will respond with a NAK handshake to any Endpoint 0 OUT tokens. Reset clears this bit.

- 1 = Data is ready to be received.
- 0 = Not ready for data. Respond with NAK.

## TP0SIZ3-TP0SIZ0 — Endpoint 0 Transmit Data Packet Size

These read/write bits store the number of transmit data bytes for the next IN token request for Endpoint 0. These bits are cleared by reset.

#### 10.5.5 USB Control Register 1 (UCR1)

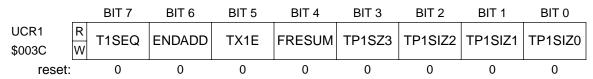


Figure 10-24. USB Control Register 1 (UCR1)

# T1SEQ — Endpoint1/Endpoint 2 Transmit Sequence Bit

This read/write bit determines which type of data packet (DATA0 or DATA1) will be sent during the next IN transaction directed to Endpoint 1 or Endpoint 2. Toggling of this bit must be controlled by software. Reset clears this bit.

- 1 = DATA1 Token active for next Endpoint 1/Endpoint 2 transmit.
- 0 = DATA0 Token active for next Endpoint 1/Endpoint 2 transmit.

#### **ENDADD** — Endpoint Address Select

This read/write bit specifies whether the data inside the registers UE1D0-UE1D7 are used for Endpoint 1 or Endpoint 2. If all the conditions for a successful Endpoint 2 USB response to a hosts IN token are satisfied (TXD1F=0, TX1E=1, STALL2=0, and ENABLE2=1) except that the ENDADD bit is configured for Endpoint 1, the USB responds with a NAK handshake packet.

- 1 = The data buffers are used for Endpoint 2.
- 0 = The data buffers are used for Endpoint 1.



GENERAL RELEASE SPECIFICATION November 5, 1998

## TX1E — Endpoint 1/Endpoint 2 Transmit Enable

This read/write bit enables a transmit to occur when the USB Host controller sends an IN token to Endpoint 1 or Endpoint 2. The appropriate endpoint enable bit, ENABLE1 or ENABLE2 bit in the UCR2 register, should also be set. Software should set the TX1E bit when data is ready to be transmitted. It must be cleared by software when no more data needs to be transmitted.

If this bit is 0 or the TXD1F is set, the USB will respond with a NAK handshake to any Endpoint 1 or Endpoint 2 directed IN tokens. Reset clears this bit.

- 1 = Data is ready to be sent.
- 0 = Data is not ready. Respond with NAK.

#### FRESUM — Force Resume

This read/write bit forces a resume state ("K" or non-idle state) onto the USB data lines to initiate a remote wake-up. Software should control the timing of the forced resume to be between 10ms and 15 ms. Setting this bit will not cause the RESUMF bit to set.

- 1 = Force data lines to "K" state.
- 0 = Default.

#### TP1SIZ3-TP1SIZ0 — Endpoint 1/Endpoint 2 Transmit Data Packet Size

These read/write bits store the number of transmit data bytes for the next IN token request for Endpoint 1 or Endpoint 2. These bits are cleared by reset.

# 10.5.6 USB Control Register 2 (UCR2)

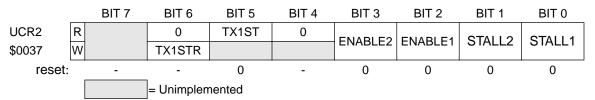


Figure 10-25. USB Control Register 2 (UCR2)

# TX1STR — Clear Transmit First Flag

Writing a logic 1 to this write-only bit will clear the TX1ST bit if it is set. Writing a logic 0 to the TX1STR has no effect. Reset clears this bit.

#### TX1ST — Transmit First Flag

This read-only bit is set if the Endpoint 0 Data Transmit Flag (TXD0F) is set when the USB control logic is setting the Endpoint 0 Data Receive Flag (RXD0F). That is, this bit will be set if an Endpoint 0 Transmit Flag is still set at the end of an Endpoint 0 reception. This bit lets the firmware know that the Endpoint 0 transmission happened before the Endpoint 0 reception. Reset clears this bit.

- 1 = IN transaction occurred before SETUP/OUT.
- 0 = IN transaction occurred after SETUP/OUT.

MOTOROLA 10-24



Semiconductor, Inc

# Freescale Semiconductor, Inc.

November 5, 1998 GENERAL RELEASE SPECIFICATION

## **ENABLE2** — Endpoint 2 Enable

This read/write bit enables Endpoint 2 and allows the USB to respond to IN packets addressed to Endpoint 2. Reset clears this bit.

- 1 = Endpoint 2 is enabled and can respond to an IN token.
- 0 = Endpoint 2 is disabled.

#### **ENABLE1** — Endpoint 1 Enable

This read/write bit enables Endpoint 1 and allows the USB to respond to IN packets addressed to Endpoint 1. Reset clears this bit.

- 1 = Endpoint 1 is enabled and can respond to an IN token.
- 0 = Endpoint 1 is disabled.

#### STALL2 — Endpoint 2 Force Stall Bit

This read/write bit causes Endpoint 2 to return a STALL handshake when polled by either an IN or OUT token by the USB Host Controller. Reset clears this bit.

- 1 = Send STALL handshake.
- 0 = Default.

## STALL1 — Endpoint 1 Force Stall Bit

This read/write bit causes Endpoint 1 to return a STALL handshake when polled by either an IN or OUT token by the USB Host Controller. Reset clears this bit.

- 1 = Send STALL handshake.
- 0 = Default.

# 10.5.7 USB Status Register (USR)

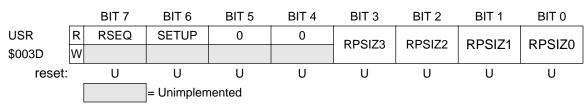


Figure 10-26. USB Status Register (USR)

#### RSEQ — Endpoint 0 Receive Sequence Bit

This read only bit indicates the type of data packet last received for Endpoint 0 (DATA0 or DATA1).

- 1 = DATA1 Token received in last Endpoint 0 receive.
- 0 = DATA0 Token received in last Endpoint 0 receive.

#### **SETUP — SETUP Token Detect Bit**

This read only bit indicates that a valid SETUP token has been received.

- 1 = Last token received for Endpoint 0 was a SETUP token.
- 0 = Last token received for Endpoint 0 was not a SETUP token.



GENERAL RELEASE SPECIFICATION November 5, 1998

## RPSIZ3-RPSIZ0 — Endpoint 0 Receive Data Packet Size

These read only bits store the number of data bytes received for the last OUT or SETUP transaction for Endpoint 0. These bits are not affected by reset.

#### 10.5.8 USB Endpoint 0 Data Registers (UE0D0-UE0D7)

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
UE0D0	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0020	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
to		:	:	:	:	:	:	:	:
ιο		:	:	:	:	:	:	:	:
UE0D7	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0027	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
reset	:	Χ	Χ	Χ	Χ	Χ	Χ	Χ	Χ

Figure 10-27. USB Endpoint 0 Data Register (UE0D0-UE0D7)

#### UE0RD7 - UE0RD0 — Endpoint 0 Receive Data Buffer

These read only bits are serially loaded with OUT token or SETUP token data received over the USB's D+ and D- pins.

#### **UE0TD7 - UE0TD0 — Endpoint 0 Transmit Data Buffer**

These write only buffers are loaded by software with data to be sent on the USB bus on the next IN token directed at Endpoint 0.

# 10.5.9 USB Endpoint 1/Endpoint 2 Data Registers (UE1D0-UE1D7)

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
UE1D0	R								
\$0028	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
to		:	:	:	:	:	:	:	:
10		:	:	:	:	:	:	:	:
UE1D7	R								
\$002F	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
reset	:	X	X	X	X	X	X	X	X

Figure 10-28. USB Endpoint 1/Endpoint2 Data Registers (UE1D0-UE1D7)

#### UE1TD7 - UE1TD0 — Endpoint 1/ Endpoint 2 Transmit Data Buffer

These write only buffers are loaded by software with data to be sent on the USB bus on the next IN token directed at Endpoint 1 or Endpoint 2. These buffers are shared by Endpoints 1 and 2 and depend on proper configuration of the ENDADD bit.

#### 10.6 USB INTERRUPTS

The USB module is capable of generating interrupts and causing the CPU to execute the USB interrupt service routine. There are three types of USB interrupts:

MOTOROLA 10-26 UNIVERSAL SERIAL BUS MODULE



November 5, 1998 GENERAL RELEASE SPECIFICATION

- End of Transaction interrupts signify a completed transaction (receive or transmit)
- Resume interrupts signify that the USB bus is reactivated after having been suspended
- End of Packet interrupts signify that a low speed end of packet signal was detected

All USB interrupts share the same interrupt vector. Firmware is responsible for determining which interrupt is active.

## 10.6.1 USB End of Transaction Interrupt

There are three possible end of transaction interrupts: Endpoint 0 Receive, Endpoint 0 Transmit, and a shared Endpoint 1 or Endpoint 2 Transmit. End of transaction interrupts occur as detailed in the following sections.

# 10.6.1.1 Receive Control Endpoint 0

For a Control OUT transaction directed at Endpoint 0, the USB module will generate an interrupt by setting the RXD0F flag in the UIR0 register. The conditions necessary for the interrupt to occur are shown in the flowchart of **Figure 10-29**.

SETUP transactions cannot be stalled by the USB function. A SETUP received by a control endpoint will clear the STALL0 bit if it is set. The conditions for receiving a SETUP interrupt are shown in **Figure 10-30**.

## 10.6.1.2 Transmit Control Endpoint 0

For a Control IN transaction directed at Endpoint 0, the USB module will generate an interrupt by setting the TXD0F flag in the UIR0 register. The conditions necessary for the interrupt to occur are shown in the flowchart of **Figure 10-31**.

#### 10.6.1.3 Transmit Endpoint 1 and Transmit Endpoint 2

Transmit Endpoints 1 and 2 share their interrupt flag. For an IN transaction directed at Endpoint 1 or 2, the USB module will generate an interrupt by setting the TXD1F flag in the UIR1 register. The conditions necessary for the interrupt to occur are shown in the flowchart of **Figure 10-32**.

#### 10.6.2 Resume Interrupt

The USB module will generate a USB interrupt if low speed bus activity is detected after entering the suspend state. A transition of the USB data lines to the non-idle state ("K" state) while in the suspend mode will set the RESUMF flag in the UIR1 register. There is no interrupt enable bit for this interrupt source and an interrupt will be executed if the I bit in the CCR is cleared. A resume interrupt can only occur while the MC68HC05JB3 is in the suspend mode.



GENERAL RELEASE SPECIFICATION November 5, 1998

# 10.6.3 End of Packet Interrupt

The USB module can generate a USB interrupt upon detection of an end of packet signal (a single ended 0) for low speed devices. Upon detection of an SE0 sequence, the USB module sets the EOPF bit and will generate an interrupt if the EOPIE bit in the UIR1 register is set.

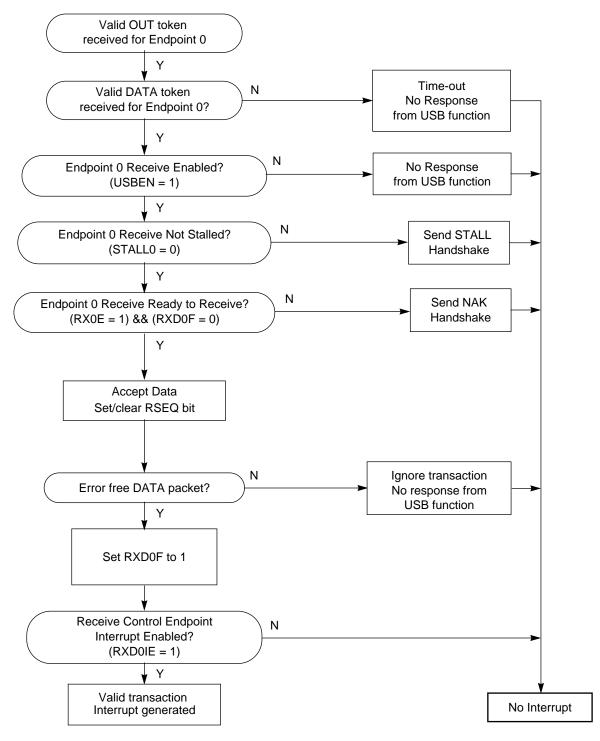


Figure 10-29. OUT Token Data Flow for Receive Endpoint 0



GENERAL RELEASE SPECIFICATION November 5, 1998

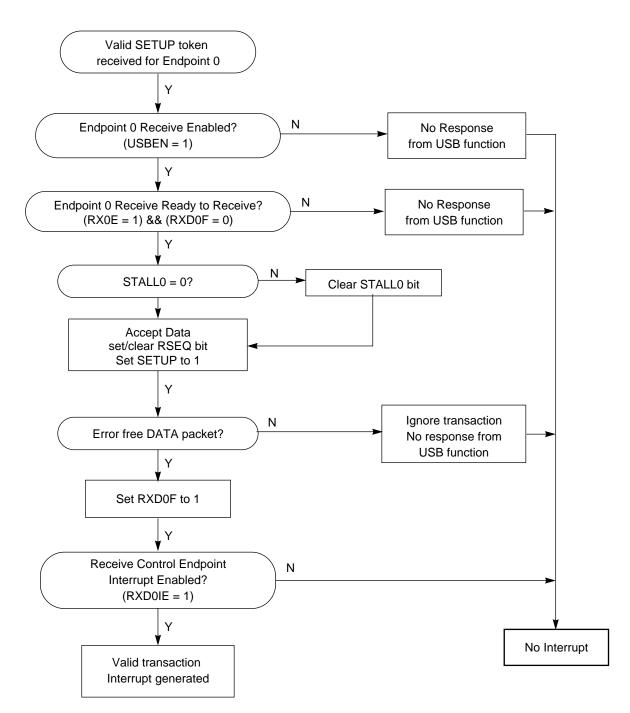


Figure 10-30. SETUP Token Data Flow for Receive Endpoint 0

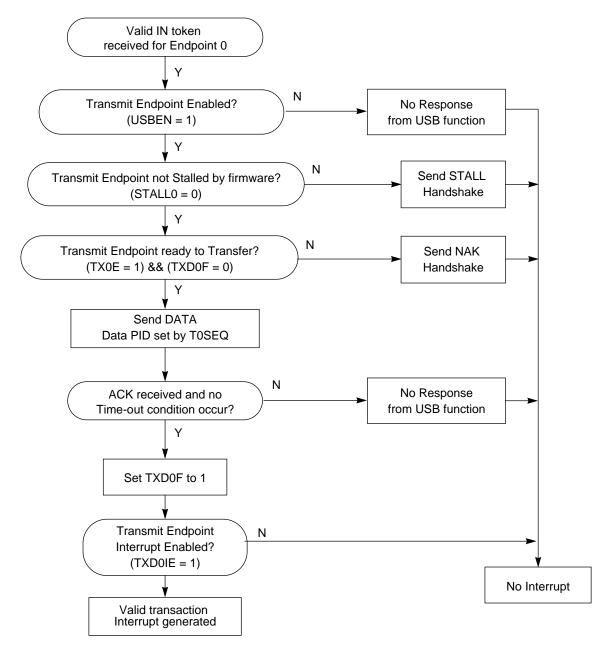


Figure 10-31. IN Token Data Flow for Transmit Endpoint 0



GENERAL RELEASE SPECIFICATION November 5, 1998

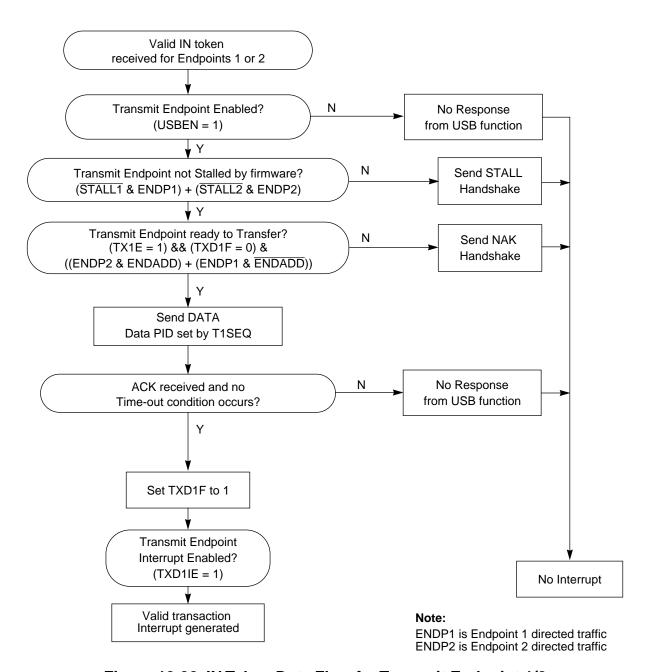


Figure 10-32. IN Token Data Flow for Transmit Endpoint 1/2



The MC68HC05JB3 MCU has four pairs of Optical Interfaces, configured through Port-A. This port has built-in optical coupler interface devices, which can be directly connected to IR displacement encoders, such as in optical mouse and optical joystick applications.

#### 11.1 OVERVIEW

In practical designs, each axis requires two optical couplers to detect the displacement. Hence, the eight optical interfaces on port-A are enabled in pairs, with each pair enabled by a bit in the Optical Interface Enable Register (\$0E). **Figure 11-1** shows a one pair of the optical interface. **Table 11-1** shows the port-A configuration for the four pairs.

**Table 11-1. Port-A Optical Interface Pairs** 

Optical Coupler	Port pin used	Enable bit in OIER
Pair 1	PA0 and PA1	OIE0
Pair 2	PA2 and PA3	OIE1
Pair 3	PA4 and PA5	OIE2
Pair 4	PA6 and PA7	OIE3

For optimal performance, the reference voltage used in the optical interface module is selectable from eight predefined values, as shown in **Figure 11-2**. and **Table 11-2**. This allows the optical interface to be easily configured by software to match the IR displacement encoders. The reference voltage is selected using the bits VREF0-VREF2 in the OIER.



GENERAL RELEASE SPECIFICATION November 5, 1998

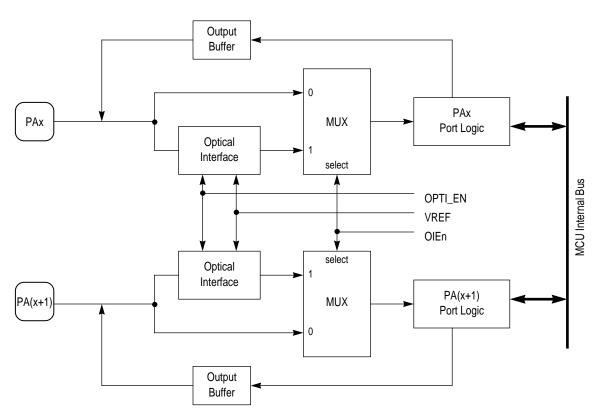


Figure 11-1. A pair of Optical Coupler Interface

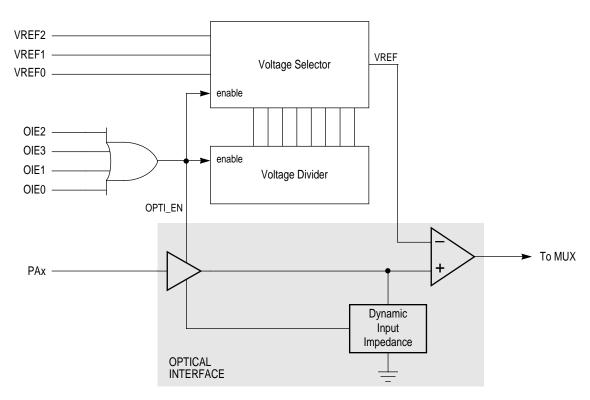


Figure 11-2. Optical Interface Comparator

MOTOROLA **OPTICAL INTERFACE** 11-2



November 5, 1998 GENERAL RELEASE SPECIFICATION

#### 11.2 OPTICAL INTERFACE ENABLE REGISTER

The OIER register controls the operation of the optical interface devices on Port-A. This register is located at address \$0E.

	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
OIER	R TCMPE	VREF2	VREF1	VREF0	OIE3	OIE2	OIE1	OIE0
\$000E	W							
reset	: 0	0	0	0	0	0	0	0

Figure 11-3. Optical Interface Enable Register (TCSR)

## OIE0 — Optical Interface pair 0 Enable

- 1 = PA0 and PA1 optical interface are enabled.
- 0 = PA0 and PA1 optical interface are disabled.

## **OIE1** — Optical Interface pair 1 Enable

- 1 = PA2 and PA3 optical interface are enabled.
- 0 = PA2 and PA3 optical interface are disabled.

## OIE2 — Optical Interface pair 2 Enable

- 1 = PA4 and PA5 optical interface are enabled.
- 0 = PA4 and PA5 optical interface are disabled.

# OIE3 — Optical Interface pair 3 Enable

- 1 = PA6 and PA7 optical interface are enabled.
- 0 = PA6 and PA7 optical interface are disabled.

## VREF[0:2] — Reference Voltage Selection

These 3 bits are used to select the optical interface reference voltage.

Table 11-2. Optical Interface Reference Voltage Selection

VREF2	VREF1	VREF0	Reference Voltage (mV), V <sub>DD</sub> =5V
0	0	0	300
0	0	1	430
0	1	0	560
0	1	1	690
1	0	0	820
1	0	1	950
1	1	0	1080
1	1	1	1210



GENERAL RELEASE SPECIFICATION November 5, 1998

# **TCMPE** — Timer Input Capture Comparator Enable

This bit is used to enable the comparator in the 16-bit timer input capture circuit. Please refer to 16-BIT TIMER section.

- 1 = Timer input capture comparator is selected.
- 0 = Timer input capture comparator schmitt trigger is selected.



November 5, 1998 GENERAL RELEASE SPECIFICATION

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# SECTION 12 INSTRUCTION SET

This section describes the addressing modes and instruction types.

## 12.1 ADDRESSING MODES

The CPU uses eight addressing modes for flexibility in accessing data. The addressing modes define the manner in which the CPU finds the data required to execute an instruction. The eight addressing modes are the following:

- Inherent
- Immediate
- Direct
- Extended
- Indexed, No Offset
- Indexed, 8-Bit Offset
- Indexed, 16-Bit Offset
- Relative

#### 12.1.1 Inherent

Inherent instructions are those that have no operand, such as return from interrupt (RTI) and stop (STOP). Some of the inherent instructions act on data in the CPU registers, such as set carry flag (SEC) and increment accumulator (INCA). Inherent instructions require no memory address and are one byte long.

#### 12.1.2 Immediate

Immediate instructions are those that contain a value to be used in an operation with the value in the accumulator or index register. Immediate instructions require no memory address and are two bytes long. The opcode is the first byte, and the immediate data value is the second byte.



GENERAL RELEASE SPECIFICATION November 5, 1998

#### 12.1.3 Direct

Direct instructions can access any of the first 256 memory addresses with two bytes. The first byte is the opcode, and the second is the low byte of the operand address. In direct addressing, the CPU automatically uses \$00 as the high byte of the operand address. BRSET and BRCLR are three-byte instructions that use direct addressing to access the operand and relative addressing to specify a branch destination.

#### 12.1.4 Extended

Extended instructions use only three bytes to access any address in memory. The first byte is the opcode; the second and third bytes are the high and low bytes of the operand address.

When using the Motorola assembler, the programmer does not need to specify whether an instruction is direct or extended. The assembler automatically selects the shortest form of the instruction.

#### 12.1.5 Indexed, No Offset

Indexed instructions with no offset are one-byte instructions that can access data with variable addresses within the first 256 memory locations. The index register contains the low byte of the conditional address of the operand. The CPU automatically uses \$00 as the high byte, so these instructions can address locations \$0000–\$00FF.

Indexed, no offset instructions are often used to move a pointer through a table or to hold the address of a frequently used RAM or I/O location.

#### 12.1.6 Indexed, 8-Bit Offset

Indexed, 8-bit offset instructions are two-byte instructions that can access data with variable addresses within the first 511 memory locations. The CPU adds the unsigned byte in the index register to the unsigned byte following the opcode. The sum is the conditional address of the operand. These instructions can access locations \$0000–\$01FE.

Indexed 8-bit offset instructions are useful for selecting the kth element in an n-element table. The table can begin anywhere within the first 256 memory locations and could extend as far as location 510 (\$01FE). The k value is typically in the index register, and the address of the beginning of the table is in the byte following the opcode.



November 5, 1998 GENERAL RELEASE SPECIFICATION

## 12.1.7 Indexed, 16-Bit Offset

Indexed, 16-bit offset instructions are three-byte instructions that can access data with variable addresses at any location in memory. The CPU adds the unsigned byte in the index register to the two unsigned bytes following the opcode. The sum is the conditional address of the operand. The first byte after the opcode is the high byte of the 16-bit offset; the second byte is the low byte of the offset. These instructions can address any location in memory.

Indexed, 16-bit offset instructions are useful for selecting the kth element in an n-element table anywhere in memory.

As with direct and extended addressing, the Motorola assembler determines the shortest form of indexed addressing.

#### 12.1.8 Relative

Relative addressing is only for branch instructions. If the branch condition is true, the CPU finds the conditional branch destination by adding the signed byte following the opcode to the contents of the program counter. If the branch condition is not true, the CPU goes to the next instruction. The offset is a signed, two's complement byte that gives a branching range of –128 to +127 bytes from the address of the next location after the branch instruction.

When using the Motorola assembler, the programmer does not need to calculate the offset, because the assembler determines the proper offset and verifies that it is within the span of the branch.

# 12.1.9 Instruction Types

The MCU instructions fall into the following five categories:

- Register/Memory Instructions
- Read-Modify-Write Instructions
- Jump/Branch Instructions
- Bit Manipulation Instructions
- Control Instructions



GENERAL RELEASE SPECIFICATION November 5, 1998

# 12.1.10 Register/Memory Instructions

Most of these instructions use two operands. One operand is in either the accumulator or the index register. The CPU finds the other operand in memory. **Table 12-1** lists the register/memory instructions.

Table 12-1. Register/Memory Instructions

Instruction	Mnemonic
Add Memory Byte and Carry Bit to Accumulator	ADC
Add Memory Byte to Accumulator	ADD
AND Memory Byte with Accumulator	AND
Bit Test Accumulator	BIT
Compare Accumulator	CMP
Compare Index Register with Memory Byte	CPX
EXCLUSIVE OR Accumulator with Memory Byte	EOR
Load Accumulator with Memory Byte	LDA
Load Index Register with Memory Byte	LDX
Multiply	MUL
OR Accumulator with Memory Byte	ORA
Subtract Memory Byte and Carry Bit from Accumulator	SBC
Store Accumulator in Memory	STA
Store Index Register in Memory	STX
Subtract Memory Byte from Accumulator	SUB



November 5, 1998 GENERAL RELEASE SPECIFICATION

### 12.1.11 Read-Modify-Write Instructions

These instructions read a memory location or a register, modify its contents, and write the modified value back to the memory location or to the register. The test for negative or zero instruction (TST) is an exception to the read-modify-write sequence because it does not write a replacement value. **Table 12-2** lists the read-modify-write instructions.

Table 12-2. Read-Modify-Write Instructions

Instruction	Mnemonic
Arithmetic Shift Left	ASL
Arithmetic Shift Right	ASR
Clear Bit in Memory	BCLR
Set Bit in Memory	BSET
Clear	CLR
Complement (One's Complement)	COM
Decrement	DEC
Increment	INC
Logical Shift Left	LSL
Logical Shift Right	LSR
Negate (Two's Complement)	NEG
Rotate Left through Carry Bit	ROL
Rotate Right through Carry Bit	ROR
Test for Negative or Zero	TST

### 12.1.12 Jump/Branch Instructions

Jump instructions allow the CPU to interrupt the normal sequence of the program counter. The unconditional jump instruction (JMP) and the jump to subroutine instruction (JSR) have no register operand. Branch instructions allow the CPU to interrupt the normal sequence of the program counter when a test condition is met. If the test condition is not met, the branch is not performed. All branch instructions use relative addressing.

Bit test and branch instructions cause a branch based on the state of any readable bit in the first 256 memory locations. These three-byte instructions use a combination of direct addressing and relative addressing. The direct address of the byte to be tested is in the byte following the opcode. The third byte is the signed offset byte. The CPU finds the conditional branch destination by adding the



GENERAL RELEASE SPECIFICATION November 5, 1998

third byte to the program counter if the specified bit tests true. The bit to be tested and its condition (set or clear) is part of the opcode. The span of branching is from –128 to +127 from the address of the next location after the branch instruction. The CPU also transfers the tested bit to the carry/borrow bit of the condition code register. **Table 12-3** lists the jump and branch instructions.

Table 12-3. Jump and Branch Instructions

Instruction	Mnemonic
Branch if Carry Bit Clear	BCC
Branch if Carry Bit Set	BCS
Branch if Equal	BEQ
Branch if Half-Carry Bit Clear	внсс
Branch if Half-Carry Bit Set	BHCS
Branch if Higher	ВНІ
Branch if Higher or Same	BHS
Branch if IRQ Pin High	BIH
Branch if IRQ Pin Low	BIL
Branch if Lower	BLO
Branch if Lower or Same	BLS
Branch if Interrupt Mask Clear	BMC
Branch if Minus	ВМІ
Branch if Interrupt Mask Set	BMS
Branch if Not Equal	BNE
Branch if Plus	BPL
Branch Always	BRA
Branch if Bit Clear	BRCLR
Branch Never	BRN
Branch if Bit Set	BRSET
Branch to Subroutine	BSR
Unconditional Jump	JMP
Jump to Subroutine	JSR



November 5, 1998 GENERAL RELEASE SPECIFICATION

### 12.1.13 Bit Manipulation Instructions

The CPU can set or clear any writable bit in the first 256 bytes of memory. Port registers, port data direction registers, timer registers, and on-chip RAM locations are in the first 256 bytes of memory. The CPU can also test and branch based on the state of any bit in any of the first 256 memory locations. Bit manipulation instructions use direct addressing. **Table 12-4** lists these instructions.

**Table 12-4. Bit Manipulation Instructions** 

Instruction	Mnemonic
Clear Bit	BCLR
Branch if Bit Clear	BRCLR
Branch if Bit Set	BRSET
Set Bit	BSET

### **12.1.14 Control Instructions**

These register reference instructions control CPU operation during program execution. Control instructions, listed in **Table 12-5**, use inherent addressing.

Table 12-5. Control Instructions

Instruction	Mnemonic
Clear Carry Bit	CLC
Clear Interrupt Mask	CLI
No Operation	NOP
Reset Stack Pointer	RSP
Return from Interrupt	RTI
Return from Subroutine	RTS
Set Carry Bit	SEC
Set Interrupt Mask	SEI
Stop Oscillator and Enable IRQ Pin	STOP
Software Interrupt	SWI
Transfer Accumulator to Index Register	TAX
Transfer Index Register to Accumulator	TXA
Stop CPU Clock and Enable Interrupts	WAIT



GENERAL RELEASE SPECIFICATION November 5, 1998

## 12.1.15 Instruction Set Summary

**Table 12-6** is an alphabetical list of all M68HC05 instructions and shows the effect of each instruction on the condition code register.

**Table 12-6. Instruction Set Summary** 

Source	Operation	Description			ect		1	Address Mode	Opcode	Operand	Cycles
Form	-	·	Н	I	N	Z	С	Adç	o	do	ે
ADC #opr ADC opr ADC opr,X ADC opr,X ADC opr,X ADC ,X	Add with Carry	$A \leftarrow (A) + (M) + (C)$	<b>‡</b>		<b>‡</b>	\$	<b>‡</b>	IMM DIR EXT IX2 IX1 IX	A9 B9 C9 D9 E9	ii dd hh II ee ff ff	2 3 4 5 4 3
ADD #opr ADD opr ADD opr, ADD opr,X ADD opr,X ADD ,X	Add without Carry	$A \leftarrow (A) + (M)$	<b>‡</b>		<b>‡</b>	<b>‡</b>	<b>‡</b>	IMM DIR EXT IX2 IX1 IX	AB BB CB DB EB FB	ii dd hh II ee ff ff	2 3 4 5 4 3
AND #opr AND opr AND opr AND opr,X AND opr,X AND ,X	Logical AND	$A \leftarrow (A) \land (M)$	_	_	<b>‡</b>	<b>‡</b>	_	IMM DIR EXT IX2 IX1 IX	A4 B4 C4 D4 E4 F4	ii dd hh II ee ff ff	2 3 4 5 4 3
ASL opr ASLA ASLX ASL opr,X ASL ,X	Arithmetic Shift Left (Same as LSL)	© - 0 b0	_	_	<b>‡</b>	<b>‡</b>	<b>‡</b>	<b>1</b>	38 48 58 68 78	dd ff	5 3 3 6 5
ASR <i>opr</i> ASRA ASRX ASR <i>opr</i> ,X ASR ,X	Arithmetic Shift Right	b7 b0		_	<b>‡</b>	<b>‡</b>	<b>‡</b>	DIR INH INH IX1 IX	37 47 57 67 77	dd ff	5 3 3 6 5
BCC rel	Branch if Carry Bit Clear	$PC \leftarrow (PC) + 2 + rel? C = 0$	-	_		_	_	REL	24	rr	3
BCLR n opr	Clear Bit n	Mn ← 0	_				_	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	13 15 17 19 1B 1D	dd dd dd dd dd dd dd	5 5 5 5 5 5 5 5
BCS rel	Branch if Carry Bit Set (Same as BLO)	PC ← (PC) + 2 + <i>rel</i> ? C = 1		_		_	_	REL	25	rr	3
BEQ rel	Branch if Equal	$PC \leftarrow (PC) + 2 + rel? Z = 1$		_	_	_	_	REL	27	rr	3



November 5, 1998 GENERAL RELEASE SPECIFICATION

**Table 12-6. Instruction Set Summary (Continued)** 

Source	Operation	Description			ect		1	Address Mode	Opcode	Operand	Cycles
Form			Н	I	N	Z	С	Adc	op	Ope	خَ
BHCC rel	Branch if Half-Carry Bit Clear	PC ← (PC) + 2 + rel ? H = 0	_	_	_	_	_	REL	28	rr	3
BHCS rel	Branch if Half-Carry Bit Set	PC ← (PC) + 2 + rel ? H = 1		_	_	_	_	REL	29	rr	3
BHI rel	Branch if Higher	$PC \leftarrow (PC) + 2 + rel? C \lor Z = 0$	<u> </u>	_	_	_	_	REL	22	rr	3
BHS rel	Branch if Higher or Same	PC ← (PC) + 2 + rel ? C = 0	_	_	_	_	_	REL	24	rr	3
BIH rel	Branch if IRQ Pin High	PC ← (PC) + 2 + <i>rel</i> ? <del>IRQ</del> = 1	_	_	_	_	_	REL	2F	rr	3
BIL rel	Branch if IRQ Pin Low	$PC \leftarrow (PC) + 2 + rel ? \overline{IRQ} = 0$		_	_	_	_	REL	2E	rr	3
BIT #opr BIT opr BIT opr, BIT opr,X BIT opr,X BIT,X	Bit Test Accumulator with Memory Byte	(A) ∧ (M)			<b>‡</b>	\$	_	IMM DIR EXT IX2 IX1 IX	A5 B5 C5 D5 E5 F5	ii dd hh II ee ff ff p	2 3 4 5 4 3
BLO rel	Branch if Lower (Same as BCS)	PC ← (PC) + 2 + rel ? C = 1	_	_	_	_	_	REL	25	rr	3
BLS rel	Branch if Lower or Same	$PC \leftarrow (PC) + 2 + rel? C \lor Z = 1$	_	_	_	_	_	REL	23	rr	3
BMC rel	Branch if Interrupt Mask Clear	PC ← (PC) + 2 + rel ? I = 0	_	_	_	_	_	REL	2C	rr	3
BMI rel	Branch if Minus	PC ← (PC) + 2 + <i>rel</i> ? N = 1	_	_	_	_	_	REL	2B	rr	3
BMS rel	Branch if Interrupt Mask Set	PC ← (PC) + 2 + rel ? I = 1		_	_	_	_	REL	2D	rr	3
BNE rel	Branch if Not Equal	$PC \leftarrow (PC) + 2 + rel? Z = 0$	_	_	_	_	_	REL	26	rr	3
BPL rel	Branch if Plus	$PC \leftarrow (PC) + 2 + rel? N = 0$	_	_	_	_	_	REL	2A	rr	3
BRA rel	Branch Always	PC ← (PC) + 2 + rel? 1 = 1	_	_	_	_	_	REL	20	rr	3
BRCLR n opr rel	Branch if bit n clear	PC ← (PC) + 2 + <i>rel</i> ? Mn = 0					<b>‡</b>	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	05 07 09 0B 0D	dd rr dd rr dd rr dd rr dd rr dd rr dd rr dd rr	5 5 5 5 5 5 5 5
	Branch if Bit n Set	PC ← (PC) + 2 + <i>rel</i> ? Mn = 1	_		_		<b>‡</b>	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	02 04 06 08 0A 0C	dd rr dd rr dd rr dd rr dd rr dd rr dd rr dd rr	5 5 5 5 5 5
BRN rel	Branch Never	$PC \leftarrow (PC) + 2 + rel? 1 = 0$						REL	21	rr	3



GENERAL RELEASE SPECIFICATION November 5, 1998

**Table 12-6. Instruction Set Summary (Continued)** 

Source	Operation	Description	I		ect	or	1	Address Mode	Opcode	Operand	Cycles
Form		P	Н	I	N	Z	С	Ado	ď	Ope	کَ
BSET n opr	Set Bit n	Mn ← 1	_	_				DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	14 16 18 1A 1C	dd dd dd dd dd dd dd	5 5 5 5 5 5 5
BSR rel	Branch to Subroutine	$\begin{array}{c} PC \leftarrow (PC) + 2;  push \; (PCL) \\ SP \leftarrow (SP) - 1;  push \; (PCH) \\ SP \leftarrow (SP) - 1 \\ PC \leftarrow (PC) + \mathit{rel} \end{array}$	_					REL	AD	rr	6
CLC	Clear Carry Bit	$C \leftarrow 0$	_	_	_	_	0	INH	98		2
CLI	Clear Interrupt Mask	I ← 0		0	_	—	_	INH	9A		2
CLR opr CLRA CLRX CLR opr,X CLR ,X	Clear Byte	$\begin{array}{l} M \leftarrow \$00 \\ A \leftarrow \$00 \\ X \leftarrow \$00 \\ M \leftarrow \$00 \\ M \leftarrow \$00 \\ M \leftarrow \$00 \end{array}$	_		0	1	_	DIR INH INH IX1 IX	3F 4F 5F 6F 7F	dd ff	5 3 3 6 5
CMP #opr CMP opr CMP opr,X CMP opr,X CMP opr,X	Compare Accumulator with Memory Byte	(A) – (M)	_	_	<b>‡</b>	<b>‡</b>	<b>‡</b>	IMM DIR EXT IX2 IX1 IX	A1 B1 C1 D1 E1	ii dd hh II ee ff ff	2 3 4 5 4 3
COM opr COMA COMX COM opr,X COM ,X	Complement Byte (One's Complement)	$\begin{aligned} M &\leftarrow (\overline{M}) = \$FF - (M) \\ A &\leftarrow (\overline{A}) = \$FF - (M) \\ X &\leftarrow (\overline{X}) = \$FF - (M) \\ M &\leftarrow (\overline{M}) = \$FF - (M) \\ M &\leftarrow (\overline{M}) = \$FF - (M) \end{aligned}$	_		<b>‡</b>	<b>‡</b>	1	DIR INH INH IX1 IX	33 43 53 63 73	dd ff	5 3 6 5
CPX #opr CPX opr CPX opr CPX opr,X CPX opr,X CPX,X	Compare Index Register with Memory Byte	(X) – (M)	_		<b>‡</b>	<b>‡</b>	<b>‡</b>	IMM DIR EXT IX2 IX1 IX	A3 B3 C3 D3 E3 F3	hh II	2 3 4 5 4 3
DEC opr DECA DECX DEC opr,X DEC ,X	Decrement Byte	$\begin{aligned} M &\leftarrow (M) - 1 \\ A &\leftarrow (A) - 1 \\ X &\leftarrow (X) - 1 \\ M &\leftarrow (M) - 1 \\ M &\leftarrow (M) - 1 \end{aligned}$		_	<b>‡</b>	<b>‡</b>	_	DIR INH INH IX1 IX	3A 4A 5A 6A 7A	dd ff	5 3 3 6 5
EOR #opr EOR opr EOR opr,X EOR opr,X EOR,X	EXCLUSIVE OR Accumulator with Memory Byte	$A \leftarrow (A) \oplus (M)$	_	_	<b>‡</b>	<b>‡</b>		IMM DIR EXT IX2 IX1 IX	A8 B8 C8 D8 E8 F8	hh II	2 3 4 5 4 3



November 5, 1998 GENERAL RELEASE SPECIFICATION

**Table 12-6. Instruction Set Summary (Continued)** 

Source	Operation	Description			ect		1	Address Mode	Opcode	Operand	Cycles
Form		·	Н	ı	N	Z	С	Adc	g	obe	ં
INC opr INCA INCX INC opr,X INC ,X	Increment Byte	$\begin{aligned} M &\leftarrow (M) + 1 \\ A &\leftarrow (A) + 1 \\ X &\leftarrow (X) + 1 \\ M &\leftarrow (M) + 1 \\ M &\leftarrow (M) + 1 \end{aligned}$	_	_	<b>‡</b>	<b>‡</b>	_	DIR INH INH IX1 IX	3C 4C 5C 6C 7C	dd ff	5 3 3 6 5
JMP opr JMP opr JMP opr,X JMP opr,X JMP ,X	Unconditional Jump	PC ← Jump Address		_	_	_	_	DIR EXT IX2 IX1 IX	BC CC DC EC FC	dd hh II ee ff ff	2 3 4 3 2
JSR opr JSR opr JSR opr,X JSR opr,X JSR ,X	Jump to Subroutine	$ \begin{array}{l} PC \leftarrow (PC) + n \; (n = 1,  2,  or \; 3) \\ Push \; (PCL); \; SP \leftarrow (SP) - 1 \\ Push \; (PCH); \; SP \leftarrow (SP) - 1 \\ PC \leftarrow Conditional \; Address \\ \end{array} $						DIR EXT IX2 IX1 IX	BD CD DD ED FD	dd hh II ee ff ff	5 6 7 6 5
LDA #opr LDA opr LDA opr, LDA opr,X LDA opr,X LDA ,X	Load Accumulator with Memory Byte	$A \leftarrow (M)$	_	_	<b>‡</b>	<b>‡</b>	_	IMM DIR EXT IX2 IX1 IX	A6 B6 C6 D6 E6 F6	ii dd hh II ee ff ff	2 3 4 5 4 3
LDX #opr LDX opr LDX opr LDX opr,X LDX opr,X LDX ,X	Load Index Register with Memory Byte	$X \leftarrow (M)$	_		\$	\$	_	IMM DIR EXT IX2 IX1 IX	AE BE CE DE EE FE	ii dd hh II ee ff ff	2 3 4 5 4 3
LSL opr LSLA LSLX LSL opr,X LSL ,X	Logical Shift Left (Same as ASL)	C 0 0 b0		_	<b>‡</b>	<b>‡</b>	<b>‡</b>	DIR INH INH IX1 IX	38 48 58 68 78	dd ff	5 3 3 6 5
LSR opr LSRA LSRX LSR opr,X LSR ,X	Logical Shift Right	0 - C b7 b0		_	0	<b>‡</b>	<b>‡</b>	DIR INH INH IX1 IX	34 44 54 64 74	dd ff	5 3 3 6 5
MUL	Unsigned Multiply	$X : A \leftarrow (X) \times (A)$	0	_	_	_	0	INH	42		11
NEG opr NEGA NEGX NEG opr,X NEG ,X	Negate Byte (Two's Complement)	$\begin{aligned} M &\leftarrow -(M) = \$00 - (M) \\ A &\leftarrow -(A) = \$00 - (A) \\ X &\leftarrow -(X) = \$00 - (X) \\ M &\leftarrow -(M) = \$00 - (M) \\ M &\leftarrow -(M) = \$00 - (M) \end{aligned}$			<b>‡</b>	<b>‡</b>	<b>‡</b>	DIR INH INH IX1 IX	30 40 50 60 70	ii ff	5 3 3 6 5
NOP	No Operation		1—	_	<u> </u>	_	_	INH	9D		2



GENERAL RELEASE SPECIFICATION November 5, 1998

## **Table 12-6. Instruction Set Summary (Continued)**

Source	Operation	Description	I		ect		n	Address Mode	Opcode	Operand	Cycles
Form	-	·	Н	I	N	Z	С	Αdα	o	Ope	ပ်
ORA #opr ORA opr ORA opr ORA opr,X ORA opr,X ORA ,X	Logical OR Accumulator with Memory	$A \leftarrow (A) \lor (M)$			\$	<b>‡</b>	_	IMM DIR EXT IX2 IX1 IX	AA BA CA DA EA FA	ii dd hh II ee ff ff	2 3 4 5 4 3
ROL opr ROLA ROLX ROL opr,X ROL ,X	Rotate Byte Left through Carry Bit	b7 b0		_	<b>‡</b>	<b>‡</b>	<b>‡</b>	DIR INH INH IX1 IX	39 49 59 69 79	dd ff	5 3 3 6 5
ROR opr RORA RORX ROR opr,X ROR ,X	Rotate Byte Right through Carry Bit	b7 b0	_	_	<b>‡</b>	<b>‡</b>	<b>‡</b>	DIR INH INH IX1 IX	36 46 56 66 76	dd ff	5 3 6 5
RSP	Reset Stack Pointer	SP ← \$00FF	-	_	_	_	-	INH	9C		2
RTI	Return from Interrupt	$\begin{array}{c} SP \leftarrow (SP) + 1;  Pull  (CCR) \\ SP \leftarrow (SP) + 1;  Pull  (A) \\ SP \leftarrow (SP) + 1;  Pull  (X) \\ SP \leftarrow (SP) + 1;  Pull  (PCH) \\ SP \leftarrow (SP) + 1;  Pull  (PCL) \end{array}$	<b>‡</b>	<b>‡</b>	<b>‡</b>	<b>‡</b>	<b>‡</b>	INH	80		9
RTS	Return from Subroutine	$SP \leftarrow (SP) + 1$ ; Pull (PCH) $SP \leftarrow (SP) + 1$ ; Pull (PCL)	-	_	_	_	_	INH	81		6
SBC #opr SBC opr SBC opr SBC opr,X SBC opr,X SBC ,X	Subtract Memory Byte and Carry Bit from Accumulator	$A \leftarrow (A) - (M) - (C)$			<b>‡</b>	<b>‡</b>	<b>‡</b>	IMM DIR EXT IX2 IX1 IX	A2 B2 C2 D2 E2 F2	ii dd hh II ee ff ff	2 3 4 5 4 3
SEC	Set Carry Bit	C ← 1	-	_	_	_	1	INH	99		2
SEI	Set Interrupt Mask	I ← 1		1	_	_	-	INH	9B		2
STA opr STA opr STA opr,X STA opr,X STA ,X	Store Accumulator in Memory	M ← (A)		_	<b>‡</b>	<b>‡</b>	_	DIR EXT IX2 IX1 IX	B7 C7 D7 E7 F7	dd hh II ee ff ff	4 5 6 5 4
STOP	Stop Oscillator and Enable IRQ Pin		-	0	_	_	_	INH	8E		2
STX opr STX opr STX opr,X STX opr,X STX ,X	Store Index Register In Memory	$M \leftarrow (X)$	_		<b>‡</b>	<b>‡</b>	_	DIR EXT IX2 IX1 IX	BF CF DF EF FF	dd hh II ee ff ff	4 5 6 5 4



November 5, 1998 GENERAL RELEASE SPECIFICATION

## **Table 12-6. Instruction Set Summary (Continued)**

Source Form	Operation	Description			ec CC	t oi R	1	Address Mode	Opcode	Operand	Cycles
FOIII			Н	I	N	Z	С	Ade	o	ope	िं
SUB #opr SUB opr SUB opr SUB opr,X SUB opr,X SUB ,X	Subtract Memory Byte from Accumulator	A ← (A) − (M)	_	_	<b>‡</b>	<b>‡</b>	<b>‡</b>	IMM DIR EXT IX2 IX1 IX	A0 B0 C0 D0 E0 F0	ii dd hh II ee ff ff	2 3 4 5 4 3
SWI	Software Interrupt	$\begin{array}{c} \text{PC} \leftarrow (\text{PC}) + 1;  \text{Push (PCL)} \\ \text{SP} \leftarrow (\text{SP}) - 1;  \text{Push (PCH)} \\ \text{SP} \leftarrow (\text{SP}) - 1;  \text{Push (X)} \\ \text{SP} \leftarrow (\text{SP}) - 1;  \text{Push (A)} \\ \text{SP} \leftarrow (\text{SP}) - 1;  \text{Push (CCR)} \\ \text{SP} \leftarrow (\text{SP}) - 1;  \text{I} \leftarrow 1 \\ \text{PCH} \leftarrow \text{Interrupt Vector High Byte} \\ \text{PCL} \leftarrow \text{Interrupt Vector Low Byte} \end{array}$	_	1				INH	83		10
TAX	Transfer Accumulator to Index Register	X ← (A)	_		_	_	_	INH	97		2
TST opr TSTA TSTX TST opr,X TST ,X	Test Memory Byte for Negative or Zero	(M) – \$00	_	_	<b>‡</b>	<b>‡</b>	_	DIR INH INH IX1 IX	3D 4D 5D 6D 7D	dd ff	4 3 3 5 4
TXA	Transfer Index Register to Accumulator	A ← (X)	_	_	_	_	_	INH	9F		2
WAIT	Stop CPU Clock and Enable Interrupts		_	0	_	_	_	INH	8F		2

Α	Accumulator	opr	Operand (one or two bytes)
С	Carry/borrow flag	PC	Program counter
C	CR Condition code register	PCH	Program counter high byte
do	d Direct address of operand	PCL	Program counter low byte
do	d rr Direct address of operand and relative offset of branch instruction	REL	Relative addressing mode
D	IR Direct addressing mode	rel	Relative program counter offset byte
ee	e ff High and low bytes of offset in indexed, 16-bit offset addressing	rr	Relative program counter offset byte
E)	XT Extended addressing mode	SP	Stack pointer
ff	Offset byte in indexed, 8-bit offset addressing	X	Index register
Н	Half-carry flag	Z	Zero flag
hh	h II High and low bytes of operand address in extended addressing	#	Immediate value
- 1	Interrupt mask	^	Logical AND
ii	Immediate operand byte	<b>V</b>	Logical OR
IN	MM Immediate addressing mode	<b>⊕</b>	Logical EXCLUSIVE OR
IN	IH Inherent addressing mode	()	Contents of
IX	Indexed, no offset addressing mode	-( )	Negation (two's complement)
IX	Indexed, 8-bit offset addressing mode	←	Loaded with
IX	12 Indexed, 16-bit offset addressing mode	?	If
M	Memory location	:	Concatenated with
Ν	Negative flag	<b>‡</b>	Set or cleared
n	Any bit	_	Not affected

Table 12-7. Opcode Map

		on	Branch		Read	l-Modify-	Write		Con	trol			Register	/Memory	,		
		· ·	REL	DIR	INH	INH	IX1	IX	INH	INH	IMM	DIR	EXT	IX2	IX1	IX	
LSB	U	1	2	3	4	5	6	7	8	9	А	В	С	D	E	F	MSB LSB
0	BRSET0 3 DIR	BSET0 2 DIR	BRA REL	NEG DIR	NEGA 1 INH	NEGX 1 INH	NEG 1X1	NEG 1 IX	9 RTI 1 INH		SUB 2 IMM	SUB DIR	SUB EXT	SUB 1X2	SUB	SUB IX	0
1	FRCLR0 DIR		BRN 2 REL						RTS 1 INH		CMP 2 IMM	CMP 2 DIR	CMP 3 EXT	5 3 CMP 3 IX2	CMP	CMP IX	1
2	SRSET1 DIR	BSET1 2 DIR	BHI 2 REL		MUL 1 INH						SBC 2 IMM	SBC DIR	SBC EXT	SBC 1X2	SBC	SBC IX	2
3	БRCLR1 U DIR	BCLR1 2 DIR	BLS REL		COMA 1 INH	COMX 1 INH	COM IX1	COM IX	SWI 1 INH		CPX 2 IMM	CPX DIR	CPX 4	CPX 1X2	CPX .	CPX IX	3
4	5 ⊮RSET2 3 DIR	BSET2 2 DIR	BCC REL	LSR 2 DIR	LSRA 1 INH	LSRX 1 INH	LSR 2 IX1	LSR 1X			AND 2 IMM	AND DIR	AND EXT	AND 1X2	AND	AND IX	4
5	PRCLR2 DIR	BCLR2 2 DIR	BCS/BLO 2 REL								BIT 2 IMM	BIT DIR	BIT 8	5 BIT 1X2	BIT 1X1	BIT IX	5
6	RSET3 DIR	BSET3 2 DIR	BNE 2 REL	ROR DIR	RORA 1 INH	RORX 1 INH	ROR IX1	ROR IX			LDA 2 IMM	LDA 2 DIR	LDA EXT	LDA IX2	LDA IX1	LDA IX	6
7	FRCLR3	BCLR3 2 DIR	BEQ REL	ASR DIR	ASRA 1 INH	ASRX 1 INH	ASR 1X1	ASR IX		TAX 1 INH		STA DIR	STA EXT	STA IX2	STA IX1	STA IX	7
8	DRSET4	BSET4 2 DIR	BHCC 2 REL	ASL/LSL 2 DIR	3 ASLA/LSLA 1 INH	ASLX/LSLX 1 INH	ASL/LSL 2 IX1	ASL/LSL 1 IX		CLC 1 INH	EOR 2	EOR 2 DIR	EOR 3 EXT	EOR 1X2	EOR 1X1	EOR IX	8
9	PRCLR4 DIR	BCLR4 2 DIR	BHCS 2 REL	ROL 2 DIR	ROLA 1 INH	ROLX 1 INH	ROL 2 IX1	ROL 1		SEC 1 INH	ADC 2 IMM	ADC 2 DIR	ADC 3 EXT	ADC 3 IX2	ADC 1X1	ADC IX	9
А	LRSET5 3 DIR	BSET5 2 DIR	BPL 2 REL	DEC 2 DIR	DECA 1 INH	DECX 1 INH	DEC 2 IX1	DEC 1 IX		CLI 1 INH	ORA 2 IMM	ORA 2 DIR	ORA 3 EXT	ORA 3 IX2	ORA 2 IX1	ORA IX	Α
В	RCLR5 DIR	BCLR5 2 DIR	BMI 2 REL							SEI 1 INH	ADD 2 IMM	ADD 2 DIR	ADD 3 EXT	ADD 3 IX2	ADD 1X1	ADD IX	В
С	DRSET6 DIR	BSET6 2 DIR	BMC 2 REL	INC 2 DIR	INCA 1 INH	INCX 1 INH	INC 1X1	INC 1 IX		RSP 1 INH		JMP 2 DIR	JMP 3 EXT	JMP 3 IX2	JMP 2 IX1	JMP IX	С
D	PRCLR6 DIR	BCLR6 2 DIR	BMS 2 REL	TST DIR	TSTA 1 INH	TSTX 1 INH	TST 5	TST 1		NOP 1 INH	BSR 2 REL	JSR 2 DIR	JSR <sup>6</sup>	JSR	JSR 2 IX1	JSR IX	D
Е	RSET7 DIR	BSET7 2 DIR	BIL 2 REL						STOP 2		LDX 2 IMM	LDX 3	LDX 4	LDX 5	LDX 4	LDX 3	Е
F	БRCLR7 3 DIR	BCLR7 2 DIR	BIH 2 REL	CLR 2 DIR	CLRA 1 INH	CLRX 1 INH	CLR 1X1	CLR 1 IX	WAIT 1 INH	TXA 1 INH		STX DIR	STX 5	STX IX2	STX SIX1	STX 4	F

INH = Inherent

REL = Relative

IMM = Immediate
DIR = Direct
EXT = Extended

IX = Indexed, No Offset IX1 = Indexed, 8-Bit Offset

IX2 = Indexed, 16-Bit Offset

LSB of Opcode in Hexadecimal

MSB of Opcode in Hexadecimal

5 Number of Cycles
Opcode Mnemonic
3 DIR Number of Bytes/Addressing Mode

MSB

LSB



This section provides the electrical and timing specifications for the MC68HC05JB3.

### 13.1 MAXIMUM RATINGS

(Voltages referenced to V<sub>SS</sub>)

Rating	Symbol	Value	Unit
Supply Voltage	V <sub>DD</sub>	-0.3 to +7.0	V
Bootloader Mode (IRQ/V <sub>PP</sub> Pin Only)	V <sub>IN</sub>	V <sub>SS</sub> – 0.3 to 17	V
Current Drain Per Pin Excluding V <sub>DD</sub> and V <sub>SS</sub>	I	25	mA
Operating Junction Temperature	TJ	+150	°C
Operating Temperature Range MC68HC05JB3 (Standard) MC68HC05JB3 (Extended)	T <sub>A</sub> T <sub>A</sub>	T <sub>L</sub> to T <sub>H</sub> 0 to +70 -40 to +85	°C
Storage Temperature Range	T <sub>stg</sub>	-65 to +150	°C

### **NOTE**

Maximum ratings are the extreme limits the device can be exposed to without causing permanent damage to the chip. The device is **not** intended to operate at these conditions.

The MCU contains circuitry that protect the inputs against damage from high static voltages; however, do not apply voltages higher than those shown in the table below. Keep  $V_{IN}$  and  $V_{OUT}$  within the range from  $V_{SS} \leq (V_{IN} \text{ or } V_{OUT}) \leq V_{DD}$ . Connect unused inputs to the appropriate voltage level, either  $V_{SS}$  or  $V_{DD}$ .

#### 13.2 THERMAL CHARACTERISTICS

Characteristic	Symbol	Value	Unit
Thermal Resistance			
20-pin PDIP	$\theta_{JA}$		°C/W
20-pin SOIC	$\theta_{JA}$		°C/W
28-pin PDIP	$\theta_{JA}$		°C/W
28-pin SOIC	$\theta_{JA}$		°C/W



GENERAL RELEASE SPECIFICATION November 5, 1998

### 13.3 DC ELECTRICAL CHARACTERISTICS

### **Table 13-1. DC Electrical Characteristics**

(V<sub>DD</sub> = 4.2 V to 5.5 V, V<sub>SS</sub> = 0 Vdc,  $T_A$  = 0°C to +70°C, unless otherwise noted)

Characteristic	Symbol	Min	Тур	Max	Unit
Output Voltage I <sub>Load</sub> = 10.0 μA	V <sub>OL</sub> V <sub>OH</sub>	— V <sub>DD</sub> – 0.1	<u> </u>	0.1 —	V
Output High Voltage (I <sub>Load</sub> =-0.8 mA) PA0-7, PB0-2, PB4-7, PC0-3	V <sub>OH</sub>	V <sub>DD</sub> - 0.8	_	_	V
Output Low Voltage (I <sub>Load</sub> = 1.6mA) PA0-3, PB0, PB4-7, PC0-3 (I <sub>Load</sub> = 8mA) PA4-7 (I <sub>Load</sub> = 25mA) PB1, PB2 (see note 8)	V <sub>OL</sub>			0.4 0.4 0.5	V
Input High Voltage PA0-7, PB0-2, PB4-7, PC0-3, IRQ, RESET, OSC1	V <sub>IH</sub>	0.7×V <sub>DD</sub>		V <sub>DD</sub>	V
Input Low Voltage PA0-7, PB0-2, PB4-7, PC0-3, IRQ, RESET, OSC1	V <sub>IL</sub>	V <sub>SS</sub>	_	0.2×V <sub>DD</sub>	V
Supply Current (see Notes) Run (USB active) Run (USB suspended) Wait (USB active) Wait (USB suspended) Stop (USB suspended) 3.3V regulator on	I <sub>DD</sub>		8 7.5 3 2.5	10 9 5 4	mA mA mA mA
I/O Ports Hi-Z Leakage Current PA0-7, PB0-2, PB4-7, PC0-3 (without individual pull-down/up activated)	I <sub>Z</sub>	_	_	±10	μΑ
Input Pull-down Current PA0-7, PB0, PB4-7, PC0-3 (with individual pull-down activated)	I <sub>IL</sub>	50	100	200	μΑ
Input Current RESET, IRQ, OSC1	I <sub>in</sub>	_	_	5	μΑ
Capacitance Ports (as Input or Output) RESET, IRQ, OSC1, OSC2	C <sub>out</sub> C <sub>in</sub>		_	12 8	pF pF
Crystal/Ceramic Resonator Oscillator Mode Internal Resistor OSC1 to OSC2	R <sub>OSC</sub>	1	2	3	ΜΩ



November 5, 1998 GENERAL RELEASE SPECIFICATION

### Table 13-1. DC Electrical Characteristics

 $(V_{DD} = 4.2 \text{V to } 5.5 \text{V}, V_{SS} = 0 \text{ Vdc}, T_A = 0^{\circ}\text{C to } +70^{\circ}\text{C}, \text{ unless otherwise noted})$ 

Characteristic	Symbol	Min	Тур	Max	Unit
Pullup Resistor PB1, PB2	R <sub>PULLUP</sub>	30	50	75	ΚΩ
LVR Inhibit (see note 9)	V <sub>LVRI</sub>		3.3		V
LVR Recover (see note 9)	V <sub>LVRR</sub>		3.5		V
TCAP Input Threshold Voltage	V <sub>TCAP</sub>		V <sub>DD</sub> /2		V

#### NOTES:

- 1. All values shown reflect average measurements.
- 2. Typical values at midpoint of voltage range, 25°C only.
- 3. Wait I<sub>DD</sub>: Only MFT and Timer1 active.
- 4. Run (Operating)  $I_{DD}$ , Wait  $I_{DD}$ : Measured using external square wave clock source to OSC1 ( $f_{OSC} = 6.0$  MHz), all inputs 0.2 VDC from rail; no DC loads, less than 50pF on all outputs,  $C_L = 20$  pF on OSC2.
- 5. Wait, Stop  $I_{DD}$ : All ports configured as inputs,  $V_{IL} = 0.2 \text{ VDC}$ ,  $V_{IH} = V_{DD} 0.2 \text{ VDC}$ .
- 6. Stop  $I_{DD}$  measured with OSC1 =  $V_{SS}$ .
- 7. Wait I<sub>DD</sub> is affected linearly by the OSC2 capacitance.
- 8.  $T_A = 0^{\circ}C$  to +40°C.
- 9. These are preliminary specifications.

### 13.4 USB DC ELECTRICAL CHARACTERISTICS

#### Table 13-2, USB DC Electrical Characteristics

 $(V_{DD} = 4.2 \text{ V to } 5.5 \text{ V}, V_{SS} = 0 \text{ Vdc}, T_A = 0^{\circ}\text{C to } +70^{\circ}\text{C}, \text{ unless otherwise noted})$ 

Characteristic	Symbol	Conditions	Min	Тур	Max	Unit
Hi-Z State Data Line Leakage	I <sub>LO</sub>	0V <vin<3.3v< td=""><td>-10</td><td></td><td>+10</td><td>μΑ</td></vin<3.3v<>	-10		+10	μΑ
Differential Input Sensitivity	V <sub>DI</sub>	(D+)-(D-)	0.2			V
Differential Common Mode Range	V <sub>CM</sub>	Includes V <sub>DI</sub> range	0.8		2.5	V
Single Ended Receiver Threshold	V <sub>SE</sub>		0.8		2.0	V
Static Output Low	V <sub>OL</sub>	R <sub>L</sub> of 1.5k to 3.6V			0.3	V
Static Output High	V <sub>OH</sub>	R <sub>L</sub> of 15k to GND	2.8		3.6	V
3.3V External Reference Pin	V <sub>3.3</sub>	I <sub>L</sub> =200μA	3.0	3.3	3.6	V



GENERAL RELEASE SPECIFICATION November 5, 1998

#### 13.5 USB LOW SPEED SOURCE ELECTRICAL CHARACTERISTICS

### **Table 13-3. USB Low Speed Source Electrical Characteristics**

Parameter	Symbol	Conditions (Notes 1,2,3)	Min	Тур	Max	Unit
Transition time: Rise Time	T <sub>R</sub>	Notes 4, 5, 8 C <sub>L</sub> =50pF C <sub>L</sub> =350pF	75		300	ns ns
Fall Time	T <sub>F</sub>	$C_L=50pF$ $C_L=350pF$	75		300	ns ns
Rise/Fall Time Matching	T <sub>RFM</sub>	T <sub>R</sub> /T <sub>F</sub>	80		120	%
Output Signal Crossover Voltage	V <sub>CRS</sub>		1.3		2.0	V
Low Speed Data Rate	T <sub>DRATE</sub>	1.5Mbs ±1.5%	1.4775 676.8	1.500 666.0	1.5225 656.8	Mbs ns
Source Differential Driver Jitter To Next Transition For Paired Transitions	T <sub>UDJ1</sub> T <sub>UDJ2</sub>	C <sub>L</sub> =350pF Notes 6 and 7	-25 -10		25 10	ns ns
Receiver Data Jitter Tolerance To Next Transition For Paired Transitions	T <sub>DJR1</sub> T <sub>DJR2</sub>	C <sub>L</sub> =350pF Notes 7	-75 -45		75 45	ns ns
Source EOP Width	TEOPT	Note 7	1.25		1.50	μs
Differential to EOP Transition Skew	TDEOP	Note 7	-40		100	ns
Receiver EOP Width  Must Reject as EOP  Must Accept	T <sub>EOPR1</sub>	Note 7	330 675			ns ns

#### NOTES:

- 1. All voltages measured from local ground, unless otherwise specified.
- 2. All timings use a capacitive load of 50pF, unless otherwise specified.
- 3. Low speed timings have a 1.5k pull-up to 2.8V on the D- data line.
- 4. Measured from 10% to 90% of the data signal.
- 5. The rising and falling edges should be smooth transitions (monotonic).
- 6. Timing differences between the differential data signals.
- 7. Measured at crossover point of differential data signals.
- 8. Capacitive loading includes 50pF of tester capacitance.

### **Table 13-4. Control Timing**

 $(V_{DD} = 4.2 \text{V to } 5.5 \text{V}, V_{SS} = 0 \text{ Vdc}, T_A = 0^{\circ}\text{C to } +70^{\circ}\text{C}, \text{ unless otherwise noted})$ 

Characteristic	Symbol	Min	Max	Units
Frequency of Operation Crystal Oscillator Option External Clock Source	fosc fosc	_ DC	6 6	MHz MHz
Internal Operating Frequency Crystal Oscillator (f <sub>OSC</sub> ÷ 2) External Clock (f <sub>OSC</sub> ÷ 2)	f <sub>OP</sub> f <sub>OP</sub>	_ DC	3 3	MHz MHz
Cycle Time (1/f <sub>OP</sub> )	t <sub>CYC</sub>	330	_	ns
RESET Pulse Width Low	t <sub>RL</sub>	1.5	_	t <sub>CYC</sub>
IRQ Interrupt Pulse Width Low (Edge-Triggered)	t <sub>ILIH</sub>	0.5	_	t <sub>CYC</sub>
IRQ Interrupt Pulse Period	t <sub>ILIL</sub>	note 1	_	t <sub>CYC</sub>
PA0 to PA3 Interrupt Pulse Width High (Edge-Triggered)	t <sub>IHIL</sub>	0.5	_	t <sub>CYC</sub>
PA0 to PA3 Interrupt Pulse Period	t <sub>IHIH</sub>	note 1	_	t <sub>CYC</sub>
OSC1 Pulse Width	t <sub>OH</sub> , t <sub>OL</sub>	_	_	ns
Output High to Low Transition Period on PA6, PA7, PB0-4	t <sub>SLOW</sub>			ns

#### NOTES:

- 1. The minimum period  $t_{ILIL}$  or  $t_{IHIH}$  should not be less than the number of cycles it takes to execute the interrupt service routine plus 19  $t_{CYC}$ .
- 2. Effects of processing, temperature, and supply voltage (excluding tolerances of external R and C)
- 3. t<sub>slow</sub> is a parameter dependent on f<sub>OSC</sub> and loading. Typical value of t<sub>slow</sub> is TENTATIVELY set at 170 ns with minimal value of 130ns and maximal value of 185ns under the SIMULATION conditions that f<sub>OSC</sub> is 6.0 MHz and slow output transition feature is enabled. Actual transition time will be specified to replace the TBDs when enough characterization has been done on various wafers from different lots. The values listed here represent data off simulation runs under the specified conditions. Under no circumstances should they be treated as the final specification.

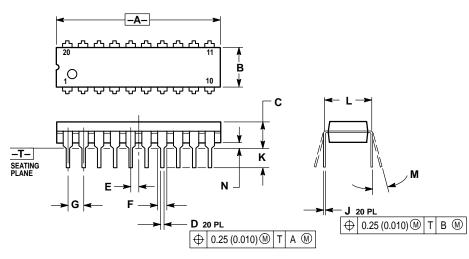


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### **SECTION 14 MECHANICAL SPECIFICATIONS**

This section provides the mechanical dimensions for the 20-pin SDIP, and 28-pin SDIP, 20-pin SOIC, and 28-pin SOIC packages.

#### 14.1 **20-PIN PDIP (CASE 738)**

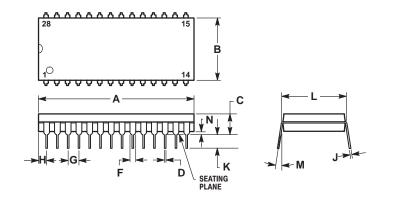


- 1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
- 2. CONTROLLING DIMENSION: INCH.
  3. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
- 4. DIMENSION B DOES NOT INCLUDE MOLD

	INC	HES	MILLIN	IETERS	
DIM	MIN	MAX	MIN	MAX	
Α	1.010	1.070	25.66	27.17	
В	0.240	0.260	6.10	6.60	
С	0.150	0.180	3.81	4.57	
D	0.015	0.022	0.39	0.55	
Е	0.050	BSC	1.27 BSC		
т	0.050	0.070	1.27 1.77		
G	0.100	BSC	2.54 BSC		
J	0.008	0.015	0.21	0.38	
K	0.110	0.140	2.80	3.55	
L	0.300 BSC		7.62 BSC		
M	0°	15°	0°	15°	
N	0.020	0.040	0.51	1.01	

Figure 14-1. 20-Pin PDIP Mechanical Dimensions

### 28-PIN PDIP (CASE 710)



#### NOTES:

- POSITIONAL TOLERANCE OF LEADS (D), SHALL BE WITHIN 0.25mm (0.010) AT MAXIMUM MATERIAL CONDITION, IN RELATION TO SEATING PLANE AND EACH OTHER.
- DIMENSION L TO CENTER OF LEADS WHEN FORMED PARALLEL.
- DIMENSION B DOES NOT INCLUDE MOLD FLASH.

	MILLIM	ETERS	INC	HES	
DIM	MIN	MAX	MIN	MAX	
Α	36.45	37.21	1.435	1.465	
В	13.72	14.22	0.540	0.560	
С	3.94	5.08	0.155	0.200	
D	0.36	0.56	0.014	0.022	
F	1.02	1.52	0.040	0.060	
G	2.54	BSC	0.100 BSC		
Н	1.65	2.16	0.065	0.085	
J	0.20	0.38	0.008	0.015	
K	2.92	3.43	0.115	0.135	
L	15.24	BSC	0.600	BSC	
M	0°	15°	0°	15°	
N	0.51	1.02	0.020	0.040	

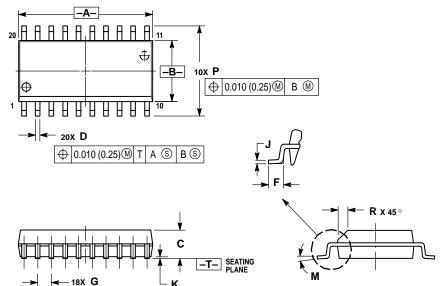
Figure 14-2. 28-Pin PDIP Mechanical Dimensions

MC68HC05JB3 REV 1

**MECHANICAL SPECIFICATIONS** 

**MOTOROLA** 

#### **20-PIN SOIC (CASE 751D)** 14.3



#### NOTES:

- DIMENSIONING AND TOLERANCING PER
- ANSI Y14.5M, 1982.

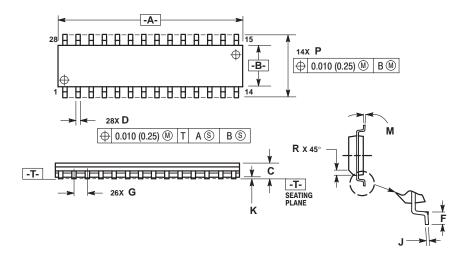
  2. CONTROLLING DIMENSION: MILLIMETER.
- 3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
  4. MAXIMUM MOLD PROTRUSION 0.150
- (0.006) PER SIDE.

  5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.13 (0.005) TOTAL IN EXCESS OF D DIMENSION AT MAXIMUM MATERIAL CONDITION.

	MILLIN	IETERS	INC	HES
DIM	MIN	MAX	MIN	MAX
Α	12.65	12.95	0.499	0.510
В	7.40	7.60	0.292	0.299
С	2.35	2.65	0.093	0.104
D	0.35	0.49	0.014	0.019
F	0.50	0.90	0.020	0.035
G	1.27	BSC	0.050	BSC
J	0.25	0.32	0.010	0.012
K	0.10	0.25	0.004	0.009
M	0 °	7°	0 °	7°
Р	10.05	10.55	0.395	0.415
R	0.25	0.75	0.010	0.029

Figure 14-3. 20-Pin SOIC Mechanical Dimensions

### **28-PIN SOIC (CASE 751F)**



- DIMENSIONING AND TOLERANCING PER
- ANSI Y14.5M, 1982.
  CONTROLLING DIMENSION: MILLIMETER.
  DIMENSION A AND B DO NOT INCLUDE MOLD
- DIMENSION A AND B DO NOT INCLUDE PROTRUSION.

  MAXIMUM MOLD PROTRUSION 0.15
  (0.006) PER SIDE.
  DIMENSION D DOES NOT INCLUDE
  DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.13 (0.005) TOTAL IN EXCESS OF D DIMENSION AT MAXIMUM MATERIAL CONDITION.

	MILLIM	ETERS	INC	HES	
DIM	MIN	MAX	MIN	MAX	
Α	17.80	18.05	0.701	0.711	
В	7.40	7.60	0.292	0.299	
С	2.35	2.65	0.093	0.104	
D	0.35	0.49	0.014	0.019	
F	0.41	0.90	0.016	0.035	
G	1.27	BSC	0.050 BSC		
J	0.23	0.32	0.009	0.013	
K	0.13	0.29	0.005	0.011	
M	0°	8°	0°	8°	
P	10.05	10.55	0.395	0.415	
R	0.25	0.75	0.010	0.029	

Figure 14-4. 28-Pin SOIC Mechanical Dimensions



This appendix describes the MC68HC705JB3, the emulation part for MC68HC05JB3. The entire MC68HC05JB3 data sheet applies to the MC68HC705JB3, with exceptions outlined in this appendix.

### A.1 INTRODUCTION

The MC68HC705JB3 is an EPROM version of the MC68HC05JB3, and is available for user system evaluation and debugging. The MC68HC705JB3 is functionally identical to the MC68HC05JB3 with the exception of the 2560 bytes user ROM is replaced by 2560 bytes user EPROM. Also, the mask options available on the MC68HC05JB3 are implemented using the Mask Option Register (MOR) in the MC68HC705JB3.

The MC68HC705JB3 is not available in the 20-pin SOIC package.

#### A.2 MEMORY

The MC68HC705JB3 memory map is shown in Figure A-1.

### A.3 MASK OPTION REGISTER (MOR)

The Mask Option Register (MOR) is a byte of EPROM used to select the features controlled by mask options on the MC68HC05JB3. In order to program this register the MORON bit in PCR need to be set to "1" before doing the EPROM programming process.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
MOR \$01FF	R W			COPEN	IRQTRIG	PULLREN	PAINTEN	OSCDLY	LVREN
reset		0	0	1	1	1	1	1	

#### **COPEN – COP Enable**

- 1 = COP watchdog function disabled.
- 0 = COP watchdog function enabled.

### IRQTRIG - IRQ, PA0-PA3 Interrupt Option

- 1 = Edge-triggered only.
- 0 = Edge-and-level-triggered.



GENERAL RELEASE SPECIFICATION November 5, 1998

### PULLREN – Port A, B, and C Pull-up/down Option

- 1 = Connected.
- 0 = Disconnected

### PAINTEN - PA0-PA3 External Interrupt Option

- 1 = External interrupt capability on PA0-PA3 disabled.
- 0 = External interrupt capability on PA0-PA3 enabled.

### **OSCDLY – Oscillator Delay Option**

- 1 = 224 internal clock cycles.
- 0 = 4064 internal clock cycles.

### LVREN - LVR Option

- 1 = Low Voltage Reset circuit enabled.
- 0 = Low Voltage Reset circuit disabled.

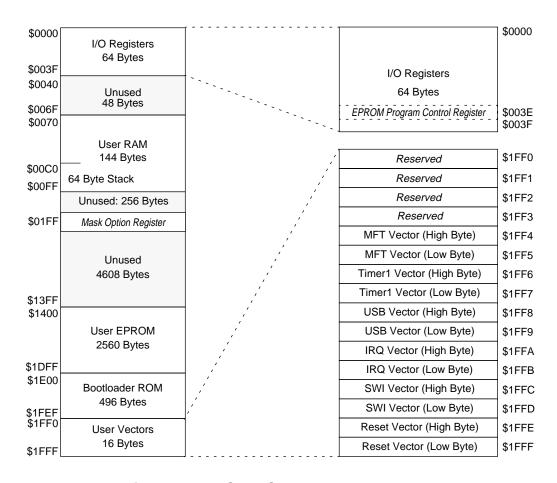


Figure A-1. MC68HC705JB3 Memory Map

MOTOROLA MC68HC05JB3
A-2 REV 1

Semiconductor, Inc

### A.4 BOOTSTRAP MODE

Bootloader mode is entered upon the rising edge of  $\overline{RESET}$  if the  $\overline{IRQ}/V_{PP}$  pin is at  $V_{TST}$  and the PB0 pin is at logic zero. The Bootloader program is masked in the ROM area from \$1E00 to \$1FEF. This program handles copying of user code from an external EPROM into the on-chip EPROM. The bootload function has to be done from an external EPROM. The bootloader performs one programming pass at 1 ms per byte then does a verify pass.

The user code must be a one-to-one correspondence with the internal EPROM addresses.

### A.5 EPROM PROGRAMMING

Programming the on-chip EPROM is achieved by using the Program Control Register located at address \$3E.

Please contact Motorola for programming board availability.

### A.5.1 EPROM Program Control Register (PCR)

This register is provided for programming the on-chip EPROM in the MC68HC705JB3.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0		
PCR	R	0	0	0	0	0	MORON	ELAT	PGM		
\$003E	W	R	R	R	R	R	IVIORON	ELAI	PGIVI		
reset:		0	0	0	0	0	0	0	0		
		R	R = Reserved								

### **MORON – Mask Option Register ON**

- 0 = Disable programming to Mask Option Register (\$01FF)
- 1 = Enable programming to Mask Option Register (\$01FF)

### **ELAT – EPROM LATch control**

- 0 = EPROM address and data bus configured for normal reads
- 1 = EPROM address and data bus configured for programming (writes to EPROM cause address and data to be latched). EPROM is in programming mode and cannot be read if ELAT is 1. This bit should not be set when no programming voltage is applied to the V<sub>DD</sub> pin.

#### PGM - EPROM ProGraM command

- 0 = Programming power is switched OFF from EPROM array.
- 1 = Programming power is switched ON to EPROM array. If ELAT≠1, then PGM=0.

### Bits [7:3] - Reserved

These are reserved bits and should remain zero.



GENERAL RELEASE SPECIFICATION November 5, 1998

### A.5.2 Programming Sequence

The EPROM programming sequence is:

- 1. Set the ELAT bit
- 2. Write the data to the address to be programmed
- 3. Set the PGM bit
- Delay for a time t<sub>PGMR</sub>
- 5. Clear the PGM bit
- 6. Clear the ELAT bit

The last two steps must be performed with separate CPU writes.

#### **CAUTION**

It is important to remember that an external programming voltage must be applied to the  $V_{PP}$  pin while programming, but it should be equal to  $V_{DD}$  during normal operations.

**Figure A-2** shows the flow required to successfully program the EPROM.

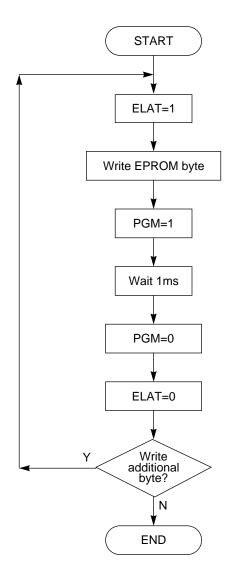


Figure A-2. EPROM Programming Sequence

### A.6 EPROM PROGRAMMING SPECIFICATIONS

## **Table A-1. EPROM Programming Electrical Characteristics**

( $V_{DD}$  = 4.2V to 5.5V,  $V_{SS}$  = 0 Vdc,  $T_A$  = 0°C to +70°C, unless otherwise noted)

Characteristic	Symbol	Min	Тур	Max	Unit
Programming Voltage IRQ/V <sub>PP</sub>	V <sub>PP</sub>	10	12	15	V
Programming Current IRQ/V <sub>PP</sub>	I <sub>PP</sub>	_	3	_	mA
Programming Time per byte	t <sub>EPGM</sub>	1	4	_	ms

MC68HC05JB3 REV 1





GENERAL RELEASE SPECIFICATION November 5, 1998

MOTOROLA MC68HC05JB3
A-6 REV 1





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