1 General description

The TEA1999TK is a member of a new generation of Synchronous Rectifier (SR) controller ICs for switched mode power supplies with adaptive gate drive for maximum efficiency at any load.

The TEA1999TK is a dedicated controller IC for synchronous rectification on the secondary side of flyback converters. It incorporates the sensing stage and driver stage for driving the SR MOSFET, which is rectifying the output of the secondary transformer winding.

The TEA1999TK can generate its own supply voltage for battery charging applications with low output voltage or for applications with high-side rectification.

The TEA1999TK is fabricated in a Silicon-On-Insulator (SOI) process.

2 Features and benefits

2.1 Efficiency features

- Adaptive gate drive for maximum efficiency at any load
- Typical supply current in no-load operation below 250 μA

2.2 Application features

- Operates in an output voltage range between 26 V and 0 V
- Drain sense pin capable of handling input voltages up to 120 V
- Self-supplying for operation with low output voltage
- Self-supplying for high-side rectification without the use of an auxiliary winding
- Operates with standard and logic level SR MOSFETs
- Supports USB BC, QuickCharge, and smart charging applications
- HVSON8 package

2.3 Control features

- Adaptive gate drive for fast turn-off at the end of conduction
- UnderVoltage LockOut (UVLO) with active gate pull-down
- Blanking input for low and high switching frequency
- Enable input for CCM operation and for disabling at start-up or shorted output
3 Applications

The TEA1999TK is intended for flyback power supplies. In such applications, it can drive the external synchronous rectifier MOSFET, which replaces the diode for the rectification of the voltage on the secondary winding of the transformer.

It can be used in all power supplies that require a high efficiency, like:

• Chargers
• Adapters
• Flyback power supplies with very low and/or variable output voltage

4 Ordering information

Table 1. Ordering information

<table>
<thead>
<tr>
<th>Type number</th>
<th>Package Name</th>
<th>Description</th>
<th>Version</th>
</tr>
</thead>
<tbody>
<tr>
<td>TEA1999TK/2</td>
<td>HVSON8</td>
<td>plastic thermal enhanced very thin small outline package; no leads; 8 terminals; body 3 mm × 3 mm × 0.85 mm</td>
<td>SOT782-1</td>
</tr>
</tbody>
</table>

5 Marking

Table 2. Marking code

<table>
<thead>
<tr>
<th>Type number</th>
<th>Marking code</th>
</tr>
</thead>
<tbody>
<tr>
<td>TEA1999TK/2</td>
<td>E1999</td>
</tr>
</tbody>
</table>
6 Block diagram

Figure 1. TEA1999TK block diagram
7 Pinning information

7.1 Pinning

![Figure 2. TEA1999TK pin configuration (SOT782-1)](image)

7.2 Pin description

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Pin</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>GATE</td>
<td>1</td>
<td>gate driver output for SR MOSFET</td>
</tr>
<tr>
<td>SOURCE</td>
<td>2</td>
<td>source sense input of SR MOSFET</td>
</tr>
<tr>
<td>ENABLE</td>
<td>3</td>
<td>enable input for SR operation</td>
</tr>
<tr>
<td>XV</td>
<td>4</td>
<td>external supply input</td>
</tr>
<tr>
<td>CAP</td>
<td>5</td>
<td>capacitor input for internal supply voltage</td>
</tr>
<tr>
<td>BLANKING</td>
<td>6</td>
<td>blanking input for minimum active time selection</td>
</tr>
<tr>
<td>GND</td>
<td>7</td>
<td>ground</td>
</tr>
<tr>
<td>DRAIN</td>
<td>8</td>
<td>drain sense input of SR MOSFET</td>
</tr>
</tbody>
</table>
8 Functional description

8.1 Introduction

The TEA1999TK is a controller IC for Synchronous Rectification (SR) in flyback applications. It can drive the external synchronous rectifier MOSFET for the rectification of the voltage on the secondary winding of the transformer. Figure 3 shows a typical configuration.

![Figure 3. TEA1999TK configuration with low-side rectification](image)

8.2 Start-up and UnderVoltage LockOut (UVLO; CAP and XV pins)

The capacitor on the CAP pin supplies the TEA1999TK. At a low CAP voltage (< 3.7 V), the capacitor is charged via the DRAIN pin with a limited start-up current of typically 15 mA. When the CAP voltage exceeds 3.7 V, the DRAIN pin or the XV pin can charge the capacitor. When the XV voltage < 4.7 V, the capacitor is charged via the DRAIN pin with a typical charge current of 125 mA. When the XV voltage ≥ 4.7 V, the capacitor is charged via the XV pin and an internal regulator. The regulator reduces the voltage difference between the XV and CAP pins to a level below 100 mV.

When the voltage on the CAP pin exceeds $V_{\text{start(CAP)}}$ (3.7 V typical), the IC leaves the UVLO state and activates the synchronous rectifier circuitry. When the voltage drops below 3.6 V (typical), the UVLO state is re-entered and the SR MOSFET gate driver output is actively kept low.

8.3 Drain sense (DRAIN pin)

The drain sense pin is an input pin capable of handling input voltages up to 120 V. At positive drain sense voltages, the gate driver is in off-mode with the gate driver pulled down (pin GATE). At negative drain sense voltages, the IC enables the Synchronous Rectification (SR) by sensing the drain source differential voltage.
8.4 Synchronous rectification (DRAIN and SOURCE pins)

The IC senses the voltage difference between the drain sense (DRAIN pin) and the source sense (SOURCE pin) connections. This drain source differential voltage of the SR MOSFET is used to drive the gate of the SR MOSFET.

When this absolute voltage difference is higher than $V_{\text{act(drv)}}$, the corresponding gate driver output turns on the external SR MOSFET. When the external SR MOSFET is switched on, the absolute voltage difference between the drain and the source sense connections drops to below $V_{\text{act(drv)}}$. The regulation phase follows the turn-on phase.

In the regulation phase, the IC regulates the difference between the drain and the source sense inputs to an absolute level of 25 mV. When the absolute difference exceeds 25 mV ($V_{\text{reg(drv)}}$), the gate driver output increases the gate voltage of the external SR MOSFET until the 25 mV level is reached. The SR MOSFET does not switch off at low current. To avoid that the device switches off because of ringing, a minimum on-time of 1.5 μs ($t_{\text{act(sr)}}$) is integrated.

When the absolute difference < 20 mV, the gate driver output decreases the gate voltage of the external SR MOSFET. The voltage waveform on the gate of the SR MOSFET follows the waveform of the current through the SR MOSFET. When the current through the SR MOSFET reaches zero, the SR MOSFET is switched off quickly.

After SR MOSFET switch-off, the drain voltage increases. When the drain voltage exceeds 250 mV, a low ohmic gate pull-down of 3 Ω keeps the gate of the SR MOSFET switched off.

8.5 Gate driver (GATE pin)

The gate driver circuit charges the gate of the external SR MOSFET during the rising part of the current. The driver circuit discharges the gate during the falling part of the current. The gate driver has a source capability of typically 0.70 A. It has a sink capability of typically 0.50 A. The source and sink capabilities allow fast turn-on and fast turn-off of the external SR MOSFET.

The maximum output voltage of the driver is limited to the voltage on the CAP pin. The maximum output voltage ranges between 4.7 V and 10 V, depending on the voltage on the CAP pin. The high output gate voltage drives all MOSFET brands to the minimum on-state resistance. In applications where the IC is supplied with 5 V, the maximum output voltage of the driver is 4.90 V, and logic level SR MOSFETs can be used.

The IC is self-supplying in applications with high-side rectification or in battery charging applications with an output voltage < 4.7 V. When the XV pin is connected to ground for driving standard SR MOSFETs, the driver is regulated to 10 V. When the XV pin is connected to the converter output for driving logic-level SR MOSFETs, the driver is regulated to the voltage on the XV pin with a minimum of 4.7 V.
8.6 Source sense (SOURCE pin)

The IC is equipped with an additional source sense pin (SOURCE). This pin is used for measuring the drain-to-source voltage of the external SR MOSFET. Voltage differences on PCB tracks because of parasitic inductance in combination with large dI/dt values, can cause errors. To minimize these errors, the source sense input must be connected as close as possible to the SOURCE pin of the external SR MOSFET.

8.7 Overtemperature protection (GATE pin)

Overtemperature protection is triggered when the output of the gate driver:

- Has a load that is too high
- Is short-circuited to ground
- Is short-circuited to the SOURCE pin

The OTP circuit is triggered at 165 °C. It actively pulls down the gate driver output. When the temperature has decreased to 145 °C, the circuit resumes normal operation.

8.8 Enable input (ENABLE pin)

The enable input can be used for enabling and disabling the SR driver.

Disabling the SR driver can be desired during start-up or during a short-circuit of the output.

The enable input can be used for turning off the SR in CCM operation.

If the output voltage is higher than 2 V, input connect to the XV pin enables the SR operation. Pulling the input to ground disables the driver.

An open input enables the SR operation by an internal 1 μA pull-up current.
8.9 Blanking input (BLANKING pin)

The blanking input can set the minimum active time ($t_{\text{act}(sr)(\text{min})}$).

An open pin or a pin connected to the CAP pin can be used for a long blanking time (1.5 $\mu$s) for applications with a switching frequency of up to 150 kHz.

A pin connected to ground can be used for a short blanking time (0.8 $\mu$s) for applications with a switching frequency of up to 300 kHz.
### 9 Limiting values

**Table 4. Limiting values**

In accordance with the Absolute Maximum Rating System (IEC 60134). All voltages are measured with respect to ground (pin 2); positive currents flow into the chip. Voltage ratings are valid provided other ratings are not violated; current ratings are valid provided the other ranges are not violated.

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{XV}$</td>
<td>voltage on pin XV</td>
<td></td>
<td>−0.4</td>
<td>+26</td>
<td>V</td>
</tr>
<tr>
<td>$V_{\text{sense(DRAIN)}}$</td>
<td>sense voltage on pin DRAIN</td>
<td></td>
<td>−0.8</td>
<td>+120</td>
<td>V</td>
</tr>
<tr>
<td>$V_{\text{sense(SOURCE)}}$</td>
<td>sense voltage on pin SOURCE</td>
<td></td>
<td>−0.4</td>
<td>+0.4</td>
<td>V</td>
</tr>
<tr>
<td>$V_{\text{ENABLE}}$</td>
<td>voltage on pin ENABLE</td>
<td></td>
<td>−0.4</td>
<td>+26</td>
<td>V</td>
</tr>
<tr>
<td>$V_{\text{BLANKING}}$</td>
<td>voltage on pin BLANKING</td>
<td></td>
<td>−0.4</td>
<td>$V_{\text{CAP}}$</td>
<td>V</td>
</tr>
</tbody>
</table>

**General**

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>$P_{\text{tot}}$</td>
<td>total power dissipation</td>
<td>$T_{\text{amb}} = 90 , ^{\circ}\text{C}$</td>
<td>-</td>
<td>1</td>
<td>W</td>
</tr>
<tr>
<td>$T_{\text{stg}}$</td>
<td>storage temperature</td>
<td></td>
<td>−55</td>
<td>+150</td>
<td>°C</td>
</tr>
<tr>
<td>$T_{\text{j}}$</td>
<td>junction temperature</td>
<td></td>
<td>−40</td>
<td>+150</td>
<td>°C</td>
</tr>
</tbody>
</table>

**ElectroStatic Discharge (ESD)**

| $V_{\text{ESD}}$ | electrostatic discharge voltage | class 2 | human body model | [1] | 2000 | V |
| $V_{\text{ESD}}$ | electrostatic discharge voltage | class 2 | charged device model | [1] | 500  | V |
| $V_{\text{ESD}}$ | electrostatic discharge voltage | class 2 | machine model | [2] | 200  | V |

[1] Equivalent to discharging a 100 pF capacitor through a 1.5 kΩ series resistor.
[2] Equivalent to discharging a 200 pF capacitor through a 10 Ω series resistor and a 0.75 μH inductor.

### 10 Recommended operating conditions

**Table 5. Recommended operating conditions**

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{XV}$</td>
<td>voltage on pin XV</td>
<td></td>
<td>0</td>
<td>-</td>
<td>21</td>
<td>V</td>
</tr>
<tr>
<td>$V_{\text{DRAIN}}$</td>
<td>voltage on pin DRAIN</td>
<td>peak voltage in switching application</td>
<td>8</td>
<td>-</td>
<td>120</td>
<td>V</td>
</tr>
</tbody>
</table>
11 Thermal characteristics

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Typ</th>
</tr>
</thead>
<tbody>
<tr>
<td>R\text{th(j-a)}</td>
<td>thermal resistance from junction to ambient</td>
<td>JEDEC test board</td>
<td>57</td>
</tr>
<tr>
<td>R\text{th(j-c)}</td>
<td>thermal resistance from junction to case</td>
<td>JEDEC test board</td>
<td>48</td>
</tr>
</tbody>
</table>

12 Characteristics

Table 7. Characteristics

−25 °C < \( T_j \) < +125 °C; \( V_xv = 5 \, V \); ENABLE connected to XV; BLANKING connected to CAP; \( C_{\text{CAP}} = 1 \, \mu F \); \( C_{\text{GATE}} = 10 \, nF \) (capacitor between the GATE and the GND pins); all voltages are measured with respect to ground (pin 2); currents are positive when flowing into the IC; unless otherwise specified.

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>( V_{\text{start(CAP)}} )</td>
<td>start voltage on pin CAP</td>
<td>( V_{XV} = 0 , V )</td>
<td>3.5</td>
<td>3.7</td>
<td>3.9</td>
<td>V</td>
</tr>
<tr>
<td>( V_{\text{stop(CAP)}} )</td>
<td>stop voltage on pin CAP</td>
<td>( V_{XV} = 0 , V )</td>
<td>3.4</td>
<td>3.6</td>
<td>3.8</td>
<td>V</td>
</tr>
<tr>
<td>( I_{\text{start(CAP)}} )</td>
<td>start current on pin CAP</td>
<td>( V_{XV} = 5 , V ); ( V_{\text{CAP}} = 0 , V ); ( V_{\text{DRAIN}} = 12 , V )</td>
<td>−24</td>
<td>−15</td>
<td>−8</td>
<td>mA</td>
</tr>
<tr>
<td>( I_{\text{ch(CAP)}} )</td>
<td>charge current on pin CAP</td>
<td>power save operation</td>
<td>−120</td>
<td>−80</td>
<td>−50</td>
<td>mA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>( V_{XV} = 0 , V ); ( V_{\text{CAP}} = 8 , V ); ( V_{\text{DRAIN}} = 12 , V ); ( T_j = 25 , ^\circ C )</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>( V_{XV} = 2 , V ); ( V_{\text{CAP}} = 4 , V ); ( V_{\text{DRAIN}} = 12 , V ); ( T_j = 25 , ^\circ C )</td>
<td>−160</td>
<td>−110</td>
<td>−50</td>
<td>mA</td>
</tr>
<tr>
<td>( V_{\text{i(CAP)}} )</td>
<td>input voltage on pin CAP</td>
<td>( V_{XV} = 0 , V ); ( V_{\text{DRAIN}} = 15 , V )</td>
<td>9.0</td>
<td>9.4</td>
<td>9.8</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>( V_{XV} = 2 , V ); ( V_{\text{DRAIN}} = 12 , V )</td>
<td>4.5</td>
<td>4.6</td>
<td>4.8</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>( V_{XV} = 5 , V )</td>
<td>4.8</td>
<td>4.9</td>
<td>5.0</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>( V_{XV} = 10 , V )</td>
<td>9.8</td>
<td>9.9</td>
<td>10.0</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>( V_{XV} = 26 , V )</td>
<td>10.3</td>
<td>10.7</td>
<td>11.1</td>
<td>V</td>
</tr>
<tr>
<td>( I_{\text{i(XV)}} )</td>
<td>input current on pin XV</td>
<td>power save operation; ( V_{\text{DRAIN}} = 5.5 , V ); ( T_j = 25 , ^\circ C )</td>
<td>200</td>
<td>240</td>
<td>280</td>
<td>( \mu A )</td>
</tr>
<tr>
<td></td>
<td></td>
<td>normal operation; without gate charge; ( V_{\text{DRAIN}} ) step from 5.5 V to −250 mV; ( T_j = 25 , ^\circ C )</td>
<td>1.0</td>
<td>1.2</td>
<td>1.4</td>
<td>mA</td>
</tr>
<tr>
<td>( t_{\text{act(pwrsave)}} )</td>
<td>power-save activation time</td>
<td></td>
<td>70</td>
<td>100</td>
<td>130</td>
<td>( \mu s )</td>
</tr>
</tbody>
</table>

Synchronous rectification sense input (DRAIN and SOURCE pins)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Parameter</th>
<th>Conditions</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>( V_{\text{act(drv)}} )</td>
<td>driver activation voltage</td>
<td>( V_{\text{SOURCE}} = 0 , V ); ( T_j = 25 , ^\circ C )</td>
<td>−510</td>
<td>−470</td>
<td>−430</td>
<td>mV</td>
</tr>
<tr>
<td>( V_{\text{reg(drv)}} )</td>
<td>driver regulation voltage</td>
<td>( V_{\text{SOURCE}} = 0 , V ); ( T_j = 25 , ^\circ C )</td>
<td>−30</td>
<td>−25</td>
<td>−20</td>
<td>mV</td>
</tr>
<tr>
<td>( V_{\text{swoff}} )</td>
<td>switch-off voltage</td>
<td>( V_{\text{SOURCE}} = 0 , V )</td>
<td>180</td>
<td>250</td>
<td>320</td>
<td>mV</td>
</tr>
</tbody>
</table>
### Symbol | Parameter | Conditions | Min | Typ | Max | Unit
--- | --- | --- | --- | --- | --- | ---
\( t_{\text{d(act)(drv)}} \) | driver activation delay time | \( V_{\text{SOURCE}} = 0 \text{ V; normal operation; time for step-on } V_{\text{DRAIN}} (2 \text{ V to } -0.5 \text{ V}) \) to rising of \( V_{\text{G}} \) at 10 % of end value | - | 40 | - | ns
\( t_{\text{d(deact)(drv)}} \) | driver deactivation delay time | \( V_{\text{SOURCE}} = 0 \text{ V; normal operation; time for step-on } V_{\text{DRAIN}} (-50 \text{ mV to } 2 \text{ V}) \) to falling of \( V_{\text{G}} \) at 90 % of begin value | - | 40 | - | ns
\( t_{\text{act}(sr)(min)} \) | minimum synchronous rectification active time | \( V_{\text{SOURCE}} = 0 \text{ V; normal operation; time for step-on } V_{\text{DRAIN}} (-700 \text{ mV to } +100 \text{ mV}) \) to falling of \( V_{\text{G}} \) at 90 % of begin value; without gate charge | 1.1 | 1.4 | 1.8 | μs
\( V_{\text{G(max)}} \) | maximum gate voltage | \( V_{\text{XV}} = 0 \text{ V; } V_{\text{ds}} = 0.5 \text{ V; } V_{\text{G}} = 0 \text{ V} \) | 9.0 | 9.4 | 9.8 | V
\( V_{\text{G}} \) | maximum gate voltage | \( V_{\text{XV}} = 5 \text{ V; } V_{\text{ds}} = 0 \text{ V; } V_{\text{G}} = 3 \text{ V} \) | 4.45 | 4.60 | 4.75 | V
\( V_{\text{G}} \) | maximum gate voltage | \( V_{\text{XV}} = 5 \text{ V; } V_{\text{ds}} = 0 \text{ V; } V_{\text{G}} = 4 \text{ V} \) | 4.8 | 4.9 | 5.0 | V
\( V_{\text{G}} \) | maximum gate voltage | \( V_{\text{XV}} = 10 \text{ V; } V_{\text{ds}} = 0 \text{ V; } V_{\text{G}} = 4 \text{ V} \) | 9.8 | 9.9 | 10.0 | V
\( V_{\text{G}} \) | maximum gate voltage | \( V_{\text{XV}} = 26 \text{ V; } V_{\text{ds}} = 0 \text{ V; } V_{\text{G}} = 4 \text{ V} \) | 10.3 | 10.7 | 11.1 | V

### Gate driver (GATE pin)

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit
--- | --- | --- | --- | --- | --- | ---
\( I_{\text{source}} \) | source current | peak.current; \( V_{\text{XV}} = 5 \text{ V; } V_{\text{ds}} = -0.5 \text{ V; } V_{\text{G}} = 0 \text{ V} \) | - | -0.70 | - | A
\( I_{\text{sink}} \) | sink current | regulation current; \( V_{\text{XV}} = 5 \text{ V; } V_{\text{ds}} = 0 \text{ V; } V_{\text{G}} = 3 \text{ V} \); peak current; \( V_{\text{XV}} = 5 \text{ V; } V_{\text{ds}} = 0.5 \text{ V; } V_{\text{G}} = 4 \text{ V} \) | - | 100 | - | mA
\( R_{\text{pd(G)}} \) | gate pull-down resistance | \( V_{\text{DRAIN}} = 0.5 \text{ V; } I_{\text{G}} = 100 \text{ mA; } V_{\text{XV}} = 5 \text{ V; } T_{j} = 25 \degree \text{C} \) | 2.6 | 3.2 | 4.0 | Ω

### Enable function (ENABLE pin)

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit
--- | --- | --- | --- | --- | --- | ---
\( V_{\text{th(en)}} \) | enable threshold voltage | | 1.1 | 1.6 | 2.0 | V
\( V_{\text{th(dis)}} \) | disable threshold voltage | | 1.0 | 1.5 | 1.9 | V
\( t_{\text{d(en)}} \) | enable delay time | turn-on delay | - | 85 | - | ns
\( t_{\text{d(dis)}} \) | disable delay time | turn-off delay | - | 100 | - | ns

### Temperature protection

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit
--- | --- | --- | --- | --- | --- | ---
\( T_{\text{otp(act)}} \) | activation overtemperature protection temperature | | 155 | 165 | 175 | °C
\( T_{\text{otp(hys)}} \) | overtemperature protection trip hysteresis | | - | 20 | - | °C
12.1 Temperature curves

12.1.1 Charge current (CAP pin)

![Graph showing charge current (CAP pin) vs temperature]

- (1) $I_{\text{CH(CAP)}}$ at $V_{\text{CAP}} = 8 \text{ V}$; $V_{\text{XV}} = 0 \text{ V}$
- (2) $I_{\text{CH(CAP)}}$ at $V_{\text{CAP}} = 4 \text{ V}$; $V_{\text{XV}} = 2 \text{ V}$

Figure 5. $I_{\text{CH(CAP)}}$ as a function of temperature

12.1.2 Operating current (XV pin)

![Graph showing operating current (XV pin) vs temperature]

- (1) $I_{\text{XV}}$ - normal operation
- (2) $I_{\text{XV}}$ - power save operation

Figure 6. $I_{\text{XV}}$ as a function of temperature
12.1.3 Driver regulation voltage

![Diagram showing V_{reg(drv)} as a function of temperature.]

Figure 7. $V_{reg(drv)}$ as a function of temperature

12.1.4 Gate pull-down resistance

![Diagram showing $R_{pd(G)}$ as a function of temperature.]

Figure 8. $R_{pd(G)}$ as a function of temperature
12.1.5  Switch-off voltage

![Graph showing V\text{swoff} as a function of temperature](aaa-027368)

Figure 9. $V_{\text{swoff}}$ as a function of temperature

12.1.6  Minimum synchronous rectification active time

![Graph showing $t_{\text{act}(sr)(\text{min})}$ as a function of temperature](aaa-027369)

Figure 10. $t_{\text{act}(sr)(\text{min})}$ as a function of temperature

(1) Blanking connected to CAP
(2) $V_{\text{blanking}} = 0$ V
13 Application information

A flyback switched mode power supply with the TEA1999TK consists of a primary side controller with a primary switch, a transformer, and an output stage. To obtain low conduction loss rectification, an SR MOSFET is used in the output stage. The SR MOSFET can be placed low-side (see Figure 3) or can be placed high-side (see Figure 11). In the high-side application, the TEA1999TK is self-supplying. The capacitor on the CAP pin supplies the TEA1999TK. When the drain voltage is positive, it is charged via the DRAIN pin.

The gate drive voltage for the synchronous rectifier switch is derived from the voltage difference between the corresponding drain sense and source sense pins.

Special attention must be paid to the connection of the drain sense and source sense pins. The voltages measured on these pins are used for the gate drive voltage. Wrong measurement results in a less efficient gate drive because a gate voltage that is either too low or too high. The connections to these pins must not interfere with the power wiring.

The power wiring conducts currents with high $dI/dt$ values. It can easily cause measurement errors resulting from induced voltages due to parasitic inductances. The separate source sense pins make it possible to sense the source voltage of the external MOSFETs directly, without having to use the current carrying power ground tracks.

![Figure 11. TEA1999TK configuration with high-side rectification](image-url)
Some important guidelines for a good layout:

- Keep the trace from the DRAIN pin to the MOSFET drain as short as possible.
- Keep the trace from the SOURCE pin to the MOSFET source as short as possible.
- Keep the area of the loop from the DRAIN pin to the MOSFET drain, to the MOSFET source, and to the SOURCE pin as small as possible. Make sure that the overlap of this loop over the power drain track or the power source track is as small as possible.
- Keep the track from the GATE pin to the gate of the MOSFET as short as possible.
- Use separate clean tracks for the XV and GND pins. If possible, use a small ground plane underneath the IC, which improves the heat dispersion.
14 Package outline

HVSON8: plastic thermal enhanced very thin small outline package; no leads; 8 terminals; body 3 x 3 x 0.85 mm

<table>
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<th>A</th>
<th>A1</th>
<th>b</th>
<th>c</th>
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<td>0.05</td>
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Note
1. Plastic or metal protrusions of 0.075 maximum per side are not included.

Outline version | References | European projection | Issue date
---|---|---|---
SOT782-1 | IEC | MO-229 | 09-08-28

Figure 12. Package outline SOT782-1 (HVSON8)
15 Revision history

Table 8. Revision history

<table>
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<tr>
<th>Document ID</th>
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<th>Data sheet status</th>
<th>Change notice</th>
<th>Supersedes</th>
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<td>Product data sheet</td>
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Modifications:
- Text and graphics have been updated throughout this document.
16 Legal information

16.1 Data sheet status

<table>
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<td>This document contains data from the objective specification for product development.</td>
</tr>
<tr>
<td>Preliminary [short] data sheet</td>
<td>Qualification</td>
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<tr>
<td>Product [short] data sheet</td>
<td>Production</td>
<td>This document contains the product specification.</td>
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</tbody>
</table>

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